

NODAL AND LOOP ANALYSIS TECHNIQUES

LEARNING GOALS

NODAL ANALYSIS
LOOP ANALYSIS

Develop systematic techniques to determine all the voltages and currents in a circuit

CIRCUITS WITH OPERATIONAL AMPLIFIERS

Op-amps are very important devices, widely available, that permit the design of very useful circuit...

and they can be modeled by circuits with dependent sources



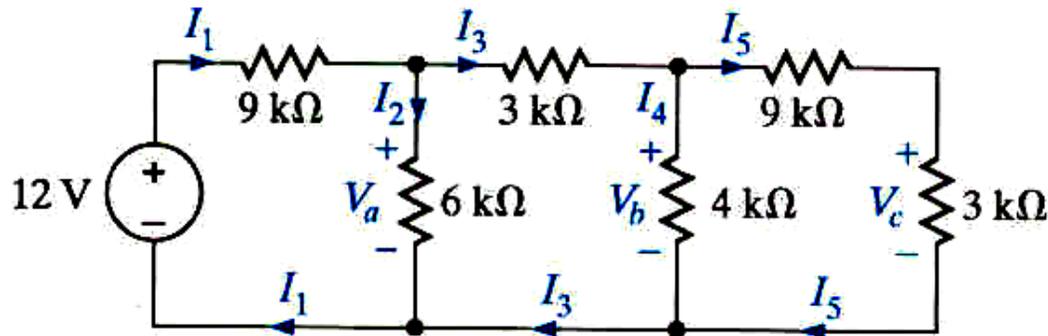
NODE ANALYSIS

- One of the systematic ways to determine every voltage and current in a circuit

The variables used to describe the circuit will be “Node Voltages”
-- The voltages of each node with respect to a pre-selected reference node



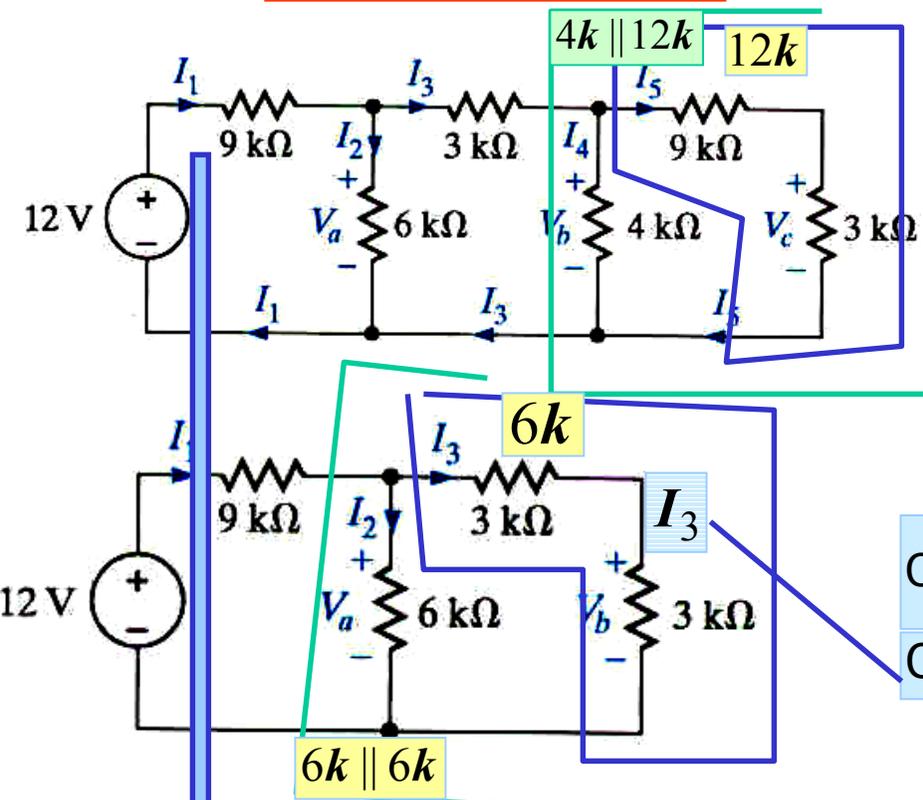
IT IS INSTRUCTIVE TO START THE PRESENTATION WITH A RECAP OF A PROBLEM SOLVED BEFORE USING SERIES/PARALLEL RESISTOR COMBINATIONS



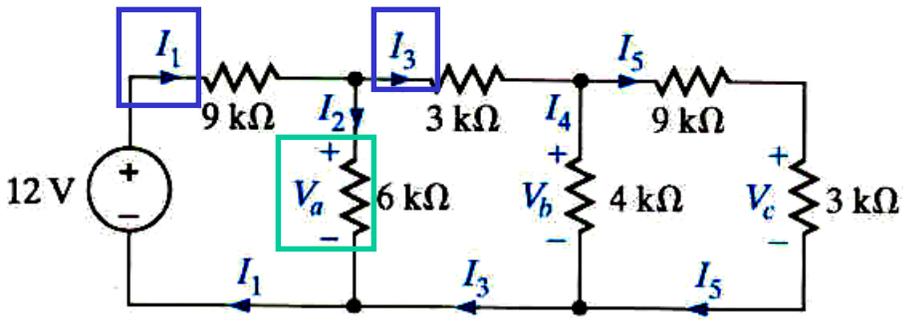
COMPUTE ALL THE VOLTAGES AND CURRENTS IN THIS CIRCUIT



We wish to find all the currents and voltages labeled in the ladder network shown



SECOND: "BACKTRACK" USING KVL, KCL OHM'S



OHM'S: $I_2 = \frac{V_a}{6k}$

KCL: $I_1 - I_2 - I_3 = 0$

OHM'S: $V_b = 3k * I_3$

...OTHER OPTIONS...

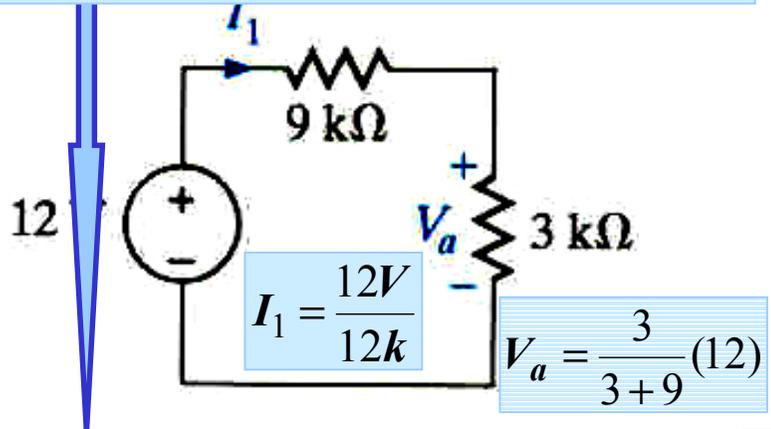
$I_4 = \frac{12}{4+12} I_3$

$V_b = 4k * I_4$

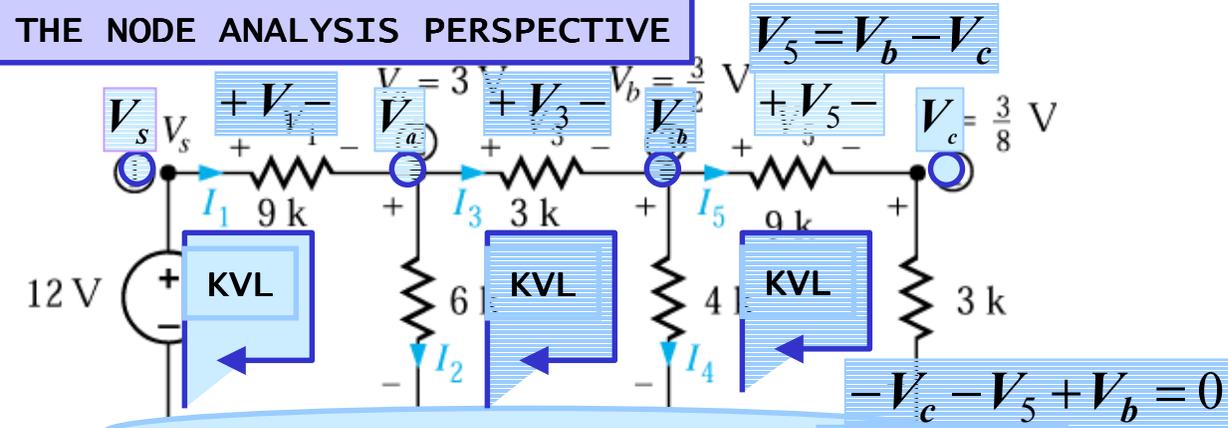
KCL: $I_5 + I_4 - I_3 = 0$

OHM'S: $V_c = 3k * I_5$

FIRST REDUCE TO A SINGLE LOOP CIRCUIT



THE NODE ANALYSIS PERSPECTIVE



THERE ARE FIVE NODES. IF ONE NODE IS SELECTED AS REFERENCE THEN THERE ARE FOUR VOLTAGES WITH RESPECT TO THEREERENCE NODE

$$-V_s + V_1 + V_a = 0$$

$$V_1 = V_s - V_a$$

$$-V_a + V_3 + V_b = 0$$

$$V_3 = V_a - V_b$$

$$-V_c - V_5 + V_b = 0$$

REFERENCE

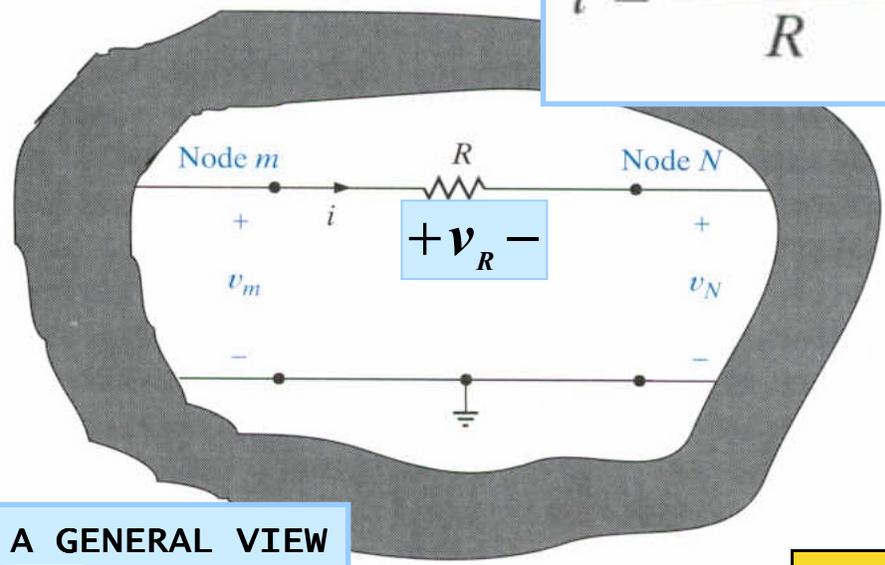
WHAT IS THE PATTERN???

ONCE THE VOLTAGES ARE KNOWN THE CURRENTS CAN BE COMPUTED USING OHM'S LAW

THEOREM: IF ALL NODE VOLTAGES WITH RESPECT TO A COMMON REFERENCE NODE ARE KNOWN THEN ONE CAN DETERMINE ANY OTHER ELECTRICAL VARIABLE FOR THE CIRCUIT

$$v_R = v_m - v_N$$

$$i = \frac{v_m - v_N}{R}$$



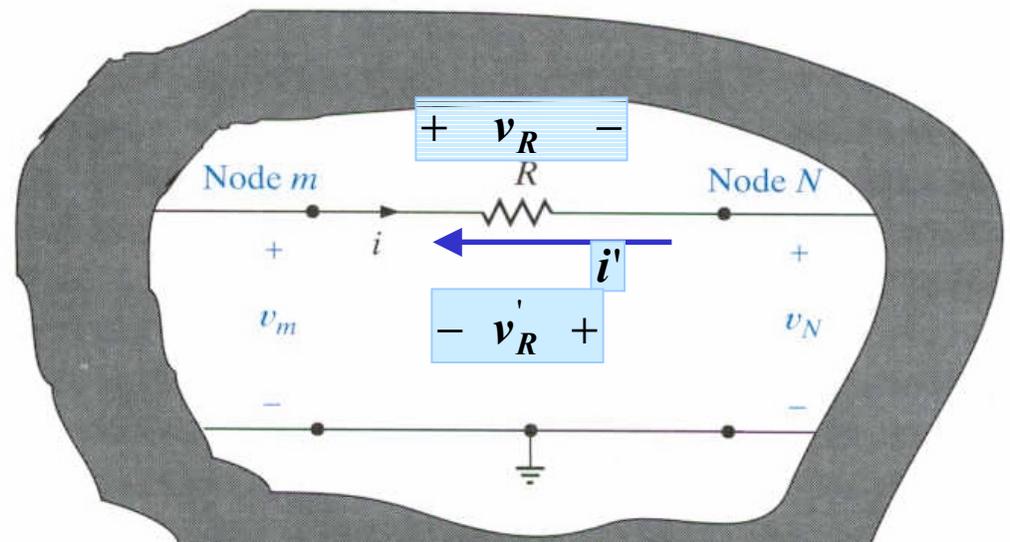
A GENERAL VIEW

DRILL QUESTION

$$V_{ca} = \underline{\hspace{2cm}}$$



THE REFERENCE DIRECTION FOR CURRENTS IS IRRELEVANT



USING THE LEFT-RIGHT REFERENCE DIRECTION THE VOLTAGE DROP ACROSS THE RESISTOR MUST HAVE THE POLARITY SHOWN

OHM'S LAW $i = \frac{v_m - v_N}{R}$

IF THE CURRENT REFERENCE DIRECTION IS REVERSED ...

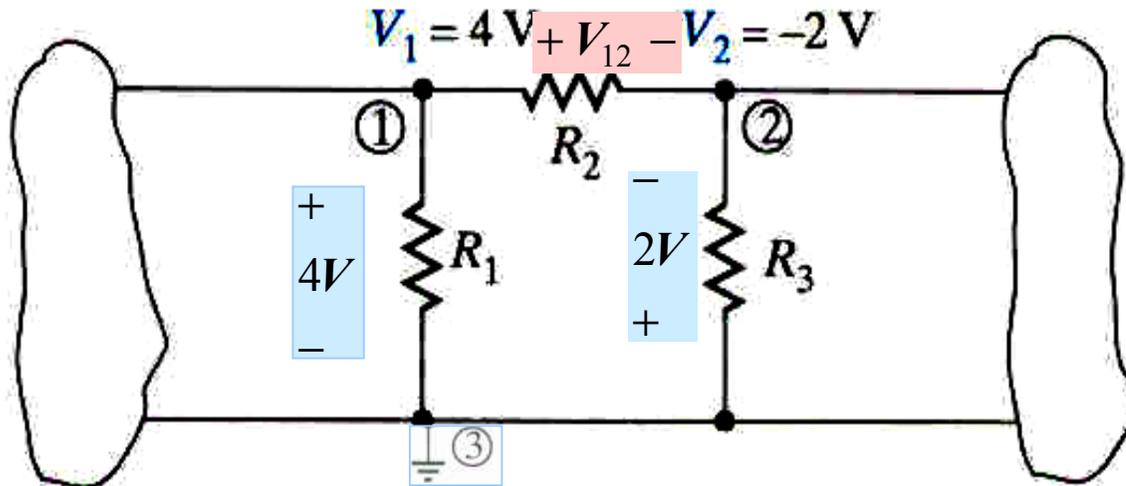
THE PASSIVE SIGN CONVENTION WILL ASSIGN THE REVERSE REFERENCE POLARITY TO THE VOLTAGE ACROSS THE RESISTOR

OHM'S LAW $i' = \frac{v_N - v_m}{R}$

$i = -i'$ PASSIVE SIGN CONVENTION RULES!



DEFINING THE REFERENCE NODE IS VITAL



THE STATEMENT $V_1 = 4V$ IS MEANINGLESS

UNTIL THE REFERENCE POINT IS DEFINED

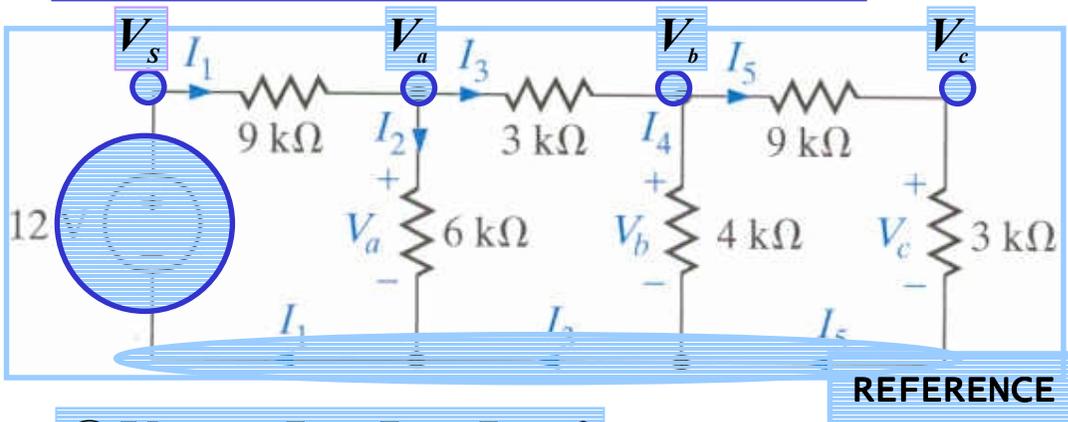
BY CONVENTION THE GROUND SYMBOL
SPECIFIES THE REFERENCE POINT.

ALL NODE VOLTAGES ARE MEASURED WITH
RESPECT TO THAT REFERENCE POINT

$$V_{12} = \underline{\hspace{2cm}} ?$$



THE STRATEGY FOR NODE ANALYSIS



1. IDENTIFY ALL NODES AND SELECT A REFERENCE NODE

2. IDENTIFY KNOWN NODE VOLTAGES

3. AT EACH NODE WITH UNKNOWN VOLTAGE WRITE A KCL EQUATION (e.g., SUM OF CURRENT LEAVING = 0)

4. REPLACE CURRENTS IN TERMS OF NODE VOLTAGES

AND GET ALGEBRAIC EQUATIONS IN THE NODE VOLTAGES ...

$$@V_a: -I_1 + I_2 + I_3 = 0$$

$$\frac{V_a - V_s}{9k} + \frac{V_a}{6k} + \frac{V_a - V_b}{3k} = 0$$

$$@V_b: -I_3 + I_4 + I_5 = 0$$

$$\frac{V_b - V_a}{3k} + \frac{V_b}{4k} + \frac{V_b - V_c}{9k} = 0$$

$$@V_c: -I_5 + I_6 = 0$$

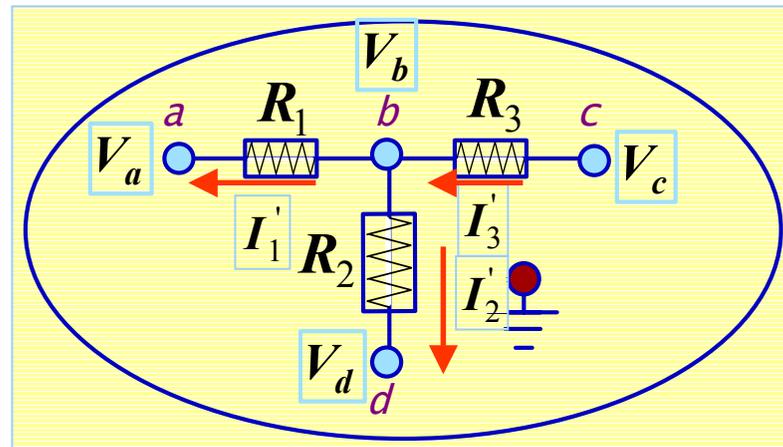
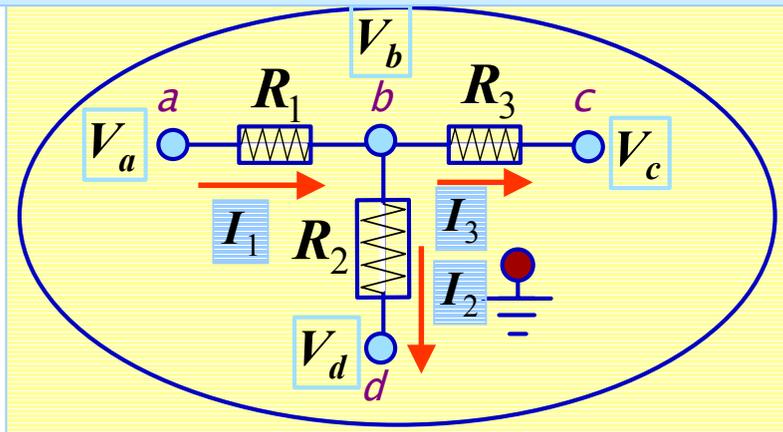
$$\frac{V_c - V_b}{9k} + \frac{V_c}{3k} = 0$$

SHORTCUT: SKIP WRITING THESE EQUATIONS...

AND PRACTICE WRITING THESE DIRECTLY



WHEN WRITING A NODE EQUATION...
AT EACH NODE ONE CAN CHOSE ARBITRARY
DIRECTIONS FOR THE CURRENTS



\sum CURRENTS LEAVING = 0

$$I'_1 + I'_2 - I'_3 = 0 \Rightarrow \frac{V_b - V_a}{R_1} + \frac{V_b - V_d}{R_2} - \frac{V_c - V_b}{R_3} = 0$$

AND SELECT ANY FORM OF KCL.
WHEN THE CURRENTS ARE REPLACED IN TERMS
OF THE NODE VOLTAGES THE NODE EQUATIONS
THAT RESULT ARE THE SAME OR EQUIVALENT

\sum CURRENTS LEAVING = 0

$$-I_1 + I_2 + I_3 = 0 \Rightarrow -\frac{V_a - V_b}{R_1} + \frac{V_b - V_d}{R_2} + \frac{V_b - V_c}{R_3} = 0$$

\sum CURRENTS INTO NODE = 0

$$I_1 - I_2 - I_3 = 0 \Rightarrow \frac{V_a - V_b}{R_1} - \frac{V_b - V_d}{R_2} - \frac{V_b - V_c}{R_3} = 0$$

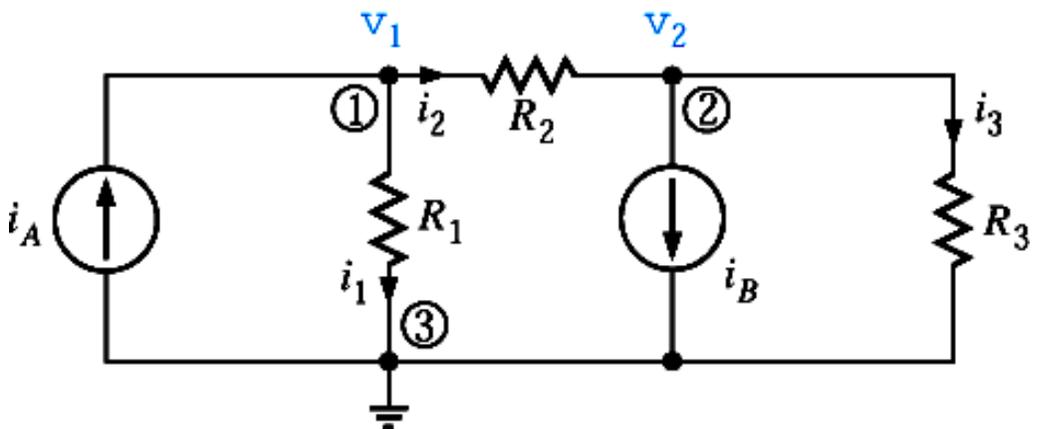
\sum CURRENTS INTO NODE = 0

$$-I'_1 - I'_2 + I'_3 = 0 \Rightarrow -\frac{V_b - V_a}{R_1} - \frac{V_b - V_d}{R_2} + \frac{V_c - V_b}{R_3} = 0$$

WHEN WRITING THE NODE EQUATIONS
WRITE THE EQUATION DIRECTLY IN TERMS
OF THE NODE VOLTAGES.
BY DEFAULT USE KCL IN THE FORM
SUM-OF-CURRENTS-LEAVING = 0

THE REFERENCE DIRECTION FOR THE
CURRENTS DOES NOT AFFECT THE NODE
EQUATION

CIRCUITS WITH ONLY INDEPENDENT SOURCES



HINT: THE FORMAL MANIPULATION OF EQUATIONS MAY BE SIMPLER IF ONE USES CONDUCTANCES INSTEAD OF RESISTANCES.

@ NODE 1 $-i_A + i_1 + i_2 = 0$

USING RESISTANCES $-i_A + \frac{v_1}{R_1} + \frac{v_1 - v_2}{R_2} = 0$

WITH CONDUCTANCES $-i_A + G_1 v_1 + G_2 (v_1 - v_2) = 0$

REORDERING TERMS $(G_1 + G_2)v_1 - G_2 v_2 = i_A$

@ NODE 2

$-G_2(v_1 - v_2) + i_B + G_3(v_2 - 0) = 0$

REORDERING TERMS $-G_2 v_1 + (G_2 + G_3)v_2 = -i_B$

THE MODEL FOR THE CIRCUIT IS A SYSTEM OF ALGEBRAIC EQUATIONS

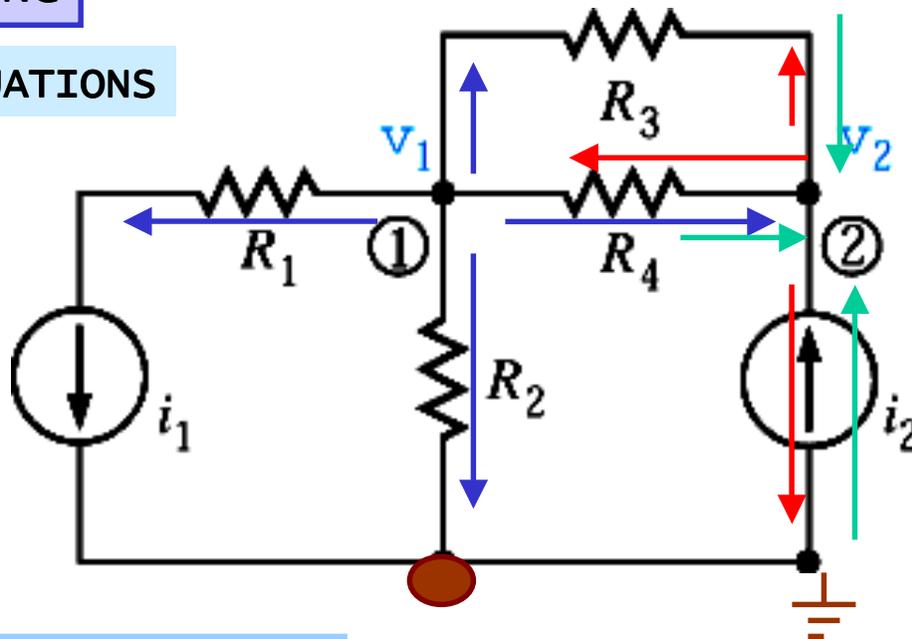
$$\begin{aligned} (G_1 + G_2)v_1 - G_2 v_2 &= i_A \\ -G_2 v_1 + (G_2 + G_3)v_2 &= -i_B \end{aligned}$$

THE MANIPULATION OF SYSTEMS OF ALGEBRAIC EQUATIONS CAN BE EFFICIENTLY DONE USING MATRIX ANALYSIS



LEARNING BY DOING

WRITE THE KCL EQUATIONS



@ NODE 1 WE VISUALIZE THE CURRENTS LEAVING AND WRITE THE KCL EQUATION

$$i_1 + \frac{v_1}{R_2} + \frac{v_1 - v_2}{R_3} + \frac{v_1 - v_2}{R_4} = 0$$

REPEAT THE PROCESS AT NODE 2

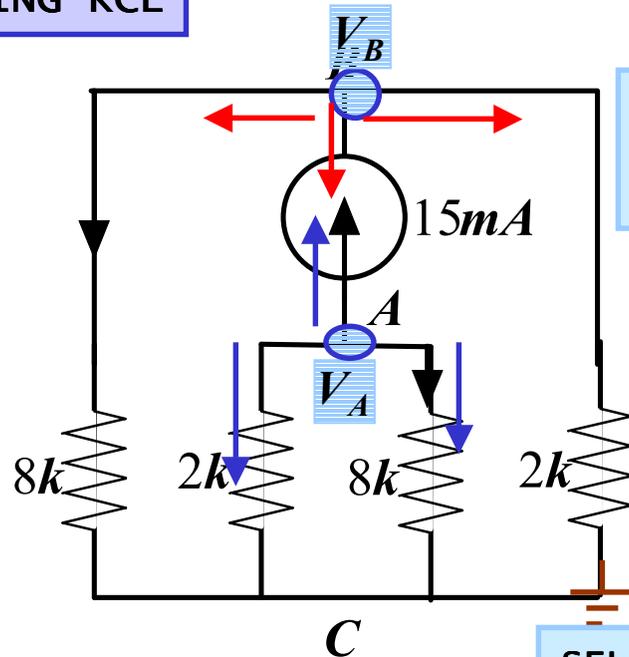
$$-i_2 + \frac{v_2 - v_1}{R_4} + \frac{v_2 - v_1}{R_3} = 0$$

OR VISUALIZE CURRENTS GOING INTO NODE

$$i_2 + \frac{v_1 - v_2}{R_3} + \frac{v_1 - v_2}{R_4} = 0$$



ANOTHER EXAMPLE OF WRITING KCL



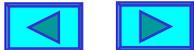
MARK THE NODES
(TO INSURE THAT
NONE IS MISSING)

SELECT AS
REFERENCE

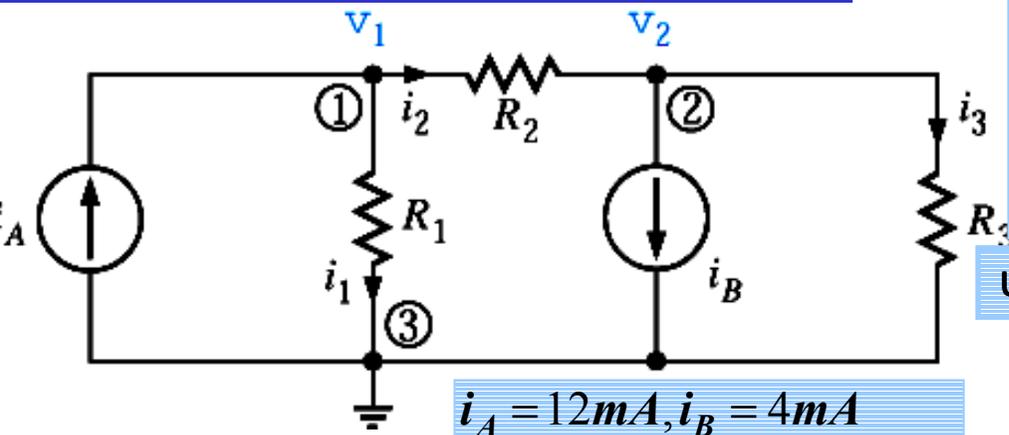
WRITE KCL AT EACH NODE IN TERMS OF
NODE VOLTAGES

$$\textcircled{A} \frac{V_A}{2k} + \frac{V_A}{8k} + 15mA = 0$$

$$\textcircled{B} \frac{V_B}{8k} + \frac{V_B}{2k} - 15mA = 0$$



A MODEL IS SOLVED BY MANIPULATION OF EQUATIONS AND USING MATRIX ANALYSIS



$i_A = 12mA, i_B = 4mA$
 $R_1 = 12k\Omega, R_2 = R_3 = 6k\Omega$

THE NODE EQUATIONS

$$-i_A + G_1(v_1 - 0) + G_2(v_1 - v_2) = 0$$

$$-G_2(v_1 - v_2) + i_B + G_3(v_2 - 0) = 0$$

THE MODEL

$$(G_1 + G_2)v_1 - G_2v_2 = i_A$$

$$-G_2v_1 + (G_2 + G_3)v_2 = -i_B$$

REPLACE VALUES AND SWITCH NOTATION TO UPPER CASE

$$V_1 \left[\frac{1}{12k} + \frac{1}{6k} \right] - V_2 \left[\frac{1}{6k} \right] = 1 \times 10^{-3}$$

$$-V_1 \left[\frac{1}{6k} \right] + V_2 \left[\frac{1}{6k} + \frac{1}{6k} \right] = -4 \times 10^{-3}$$

NUMERICAL MODEL

$$\frac{V_1}{4k} - \frac{V_2}{6k} = 1 \times 10^{-3}$$

$$-\frac{V_1}{6k} + \frac{V_2}{3k} = -4 \times 10^{-3}$$

LEARNING EXAMPLE

$$V_1 = V_2 \left(\frac{2}{3} \right) + 4$$

USE GAUSSIAN ELIMINATION

$$-\frac{1}{6k} \left(\frac{2}{3} V_2 + 4 \right) + \frac{V_2}{3k} = -4 \times 10^{-3}$$

$$V_2 = -15V \quad V_1 = \frac{2}{3} V_2 + 4 = -6V$$

ALTERNATIVE MANIPULATION

$$\frac{V_1}{4k} - \frac{V_2}{6k} = 1 \times 10^{-3} \quad */12k$$

$$-\frac{V_1}{6k} + \frac{V_2}{3k} = -4 \times 10^{-3} \quad */6k$$

RIGHT HAND SIDE IS VOLTS. COEFFS ARE NUMBERS

$$3V_1 - 2V_2 = 12$$

$$-V_1 + 2V_2 = -24 \quad */3 \text{ (and add equations)}$$

ADD EQS $2V_1 = -12[V] \quad 4V_2 = -60[V]$



SOLUTION USING MATRIX ALGEBRA

$$\begin{aligned} \frac{V_1}{4k} - \frac{V_2}{6k} &= 1 \times 10^{-3} \\ -\frac{V_1}{6k} + \frac{V_2}{3k} &= -4 \times 10^{-3} \end{aligned}$$

$$\text{Adj } \mathbf{A} = \begin{bmatrix} \frac{1}{3k} & \frac{1}{6k} \\ \frac{1}{6k} & \frac{1}{4k} \end{bmatrix} \quad |\mathbf{A}| = \left(\frac{1}{3k}\right)\left(\frac{1}{4k}\right) - \left(\frac{-1}{6k}\right)\left(\frac{-1}{6k}\right) = \frac{1}{18k^2}$$

PLACE IN MATRIX FORM

$$\begin{bmatrix} \frac{1}{4k} & \frac{-1}{6k} \\ \frac{-1}{6k} & \frac{1}{3k} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 1 \times 10^{-3} \\ -4 \times 10^{-3} \end{bmatrix}$$

USE MATRIX ANALYSIS TO SHOW SOLUTION

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} \frac{1}{4k} & \frac{-1}{6k} \\ \frac{-1}{6k} & \frac{1}{3k} \end{bmatrix}^{-1} \begin{bmatrix} 1 \times 10^{-3} \\ -4 \times 10^{-3} \end{bmatrix}$$

PERFORM THE MATRIX MANIPULATIONS

$$\mathbf{A}^{-1} = \frac{\text{Adj}(\mathbf{A})}{|\mathbf{A}|}$$

FOR THE ADJOINT REPLACE EACH ELEMENT BY ITS COFACTOR

AND DO THE MATRIX ALGEBRA ...

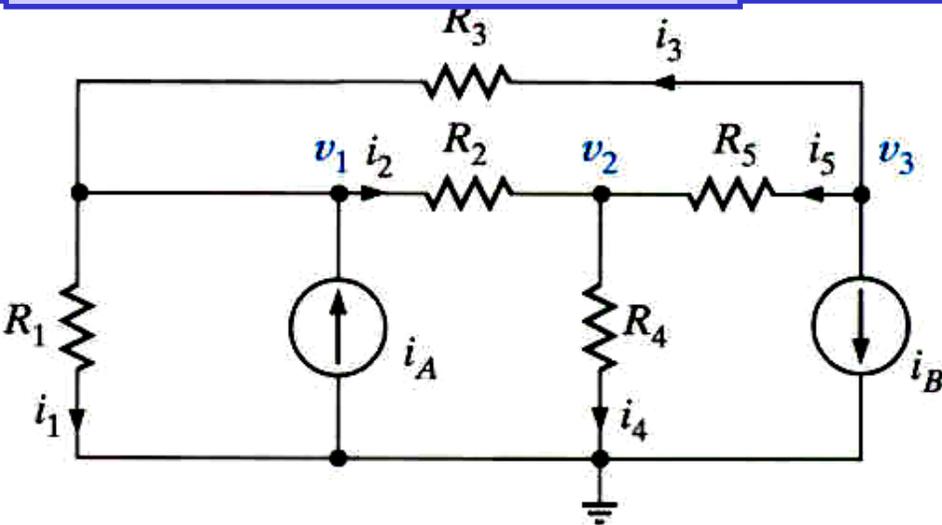
$$\begin{aligned} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} &= 18k^2 \begin{bmatrix} \frac{1}{3k} & \frac{1}{6k} \\ \frac{1}{6k} & \frac{1}{4k} \end{bmatrix} \begin{bmatrix} 1 \times 10^{-3} \\ -4 \times 10^{-3} \end{bmatrix} \\ &= 18k^2 \begin{bmatrix} \frac{1}{3k^2} - \frac{4}{6k^2} \\ \frac{1}{6k^2} - \frac{1}{k^2} \end{bmatrix} \\ &= \begin{bmatrix} -6 \\ -15 \end{bmatrix} \end{aligned}$$

SAMPLE

$$V_1 = 18k^2 \left(\frac{1 \times 10^{-3}}{3k} + \frac{-4 \times 10^{-3}}{6k} \right)$$

GEAUX 

AN EXAMPLE OF NODE ANALYSIS



Rearranging terms ...

$$v_1 \left(\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right) - v_2 \frac{1}{R_2} - v_3 \frac{1}{R_3} = i_A$$

$$-v_1 \frac{1}{R_2} + v_2 \left(\frac{1}{R_2} + \frac{1}{R_4} + \frac{1}{R_5} \right) - v_3 \frac{1}{R_5} = 0$$

$$-v_1 \frac{1}{R_3} - v_2 \frac{1}{R_5} + v_3 \left(\frac{1}{R_3} + \frac{1}{R_5} \right) = -i_B$$

@ v₁

$$\frac{v_1}{R_1} - i_A + \frac{v_1 - v_2}{R_2} - \frac{v_3 - v_1}{R_3} = 0$$

$$v_1 \left(\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right) - v_2 \frac{1}{R_2} - v_3 \frac{1}{R_3} = i_A$$

@ v₂

$$-\frac{v_1 - v_2}{R_2} + \frac{v_2}{R_4} - \frac{v_3 - v_2}{R_5} = 0$$

$$-v_1 \frac{1}{R_2} + v_2 \left(\frac{1}{R_2} + \frac{1}{R_4} + \frac{1}{R_5} \right) - v_3 \frac{1}{R_5} = 0$$

@ v₃

$$\frac{v_3 - v_1}{R_3} + \frac{v_3 - v_2}{R_5} + i_B = 0$$

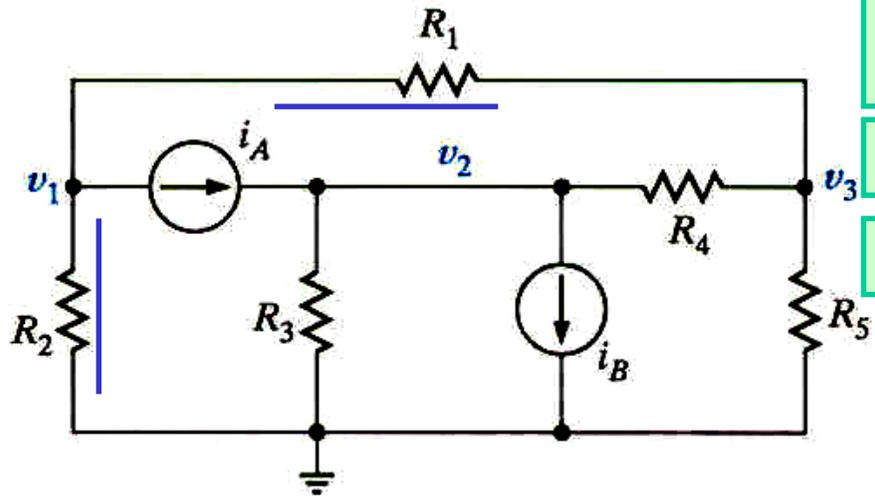
$$-v_1 \frac{1}{R_3} - v_2 \frac{1}{R_5} + v_3 \left(\frac{1}{R_3} + \frac{1}{R_5} \right) = -i_B$$

$$\begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} & -\frac{1}{R_2} & -\frac{1}{R_3} \\ -\frac{1}{R_2} & \frac{1}{R_2} + \frac{1}{R_4} + \frac{1}{R_5} & -\frac{1}{R_5} \\ -\frac{1}{R_3} & -\frac{1}{R_5} & \frac{1}{R_3} + \frac{1}{R_5} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} i_A \\ 0 \\ -i_B \end{bmatrix}$$

COULD WRITE EQUATIONS BY INSPECTION

- = \sum CONDUCTANCES CONNECTED TO NODE
- = \sum CONDUCTANCES BETWEEN 1 & 2
- = \sum CONDUCTANCES BETWEEN 1 & 3
- = \sum CONDUCTANCES BETWEEN 2 & 3

WRITING EQUATIONS "BY INSPECTION"



FOR CIRCUITS WITH ONLY INDEPENDENT SOURCES THE MATRIX IS ALWAYS SYMMETRIC

THE DIAGONAL ELEMENTS ARE POSITIVE

THE OFF-DIAGONAL ELEMENTS ARE NEGATIVE

Conductances connected to node 1

Conductances between 1 and 2

Conductances between 1 and 3

Conductances between 2 and 3

$$v_1 \left(\frac{1}{R_1} + \frac{1}{R_2} \right) - v_2 (0) - v_3 \left(\frac{1}{R_1} \right) = -i_A$$

$$-v_1 (0) + v_2 \left(\frac{1}{R_3} + \frac{1}{R_4} \right) - v_3 \left(\frac{1}{R_4} \right) = i_A - i_B$$

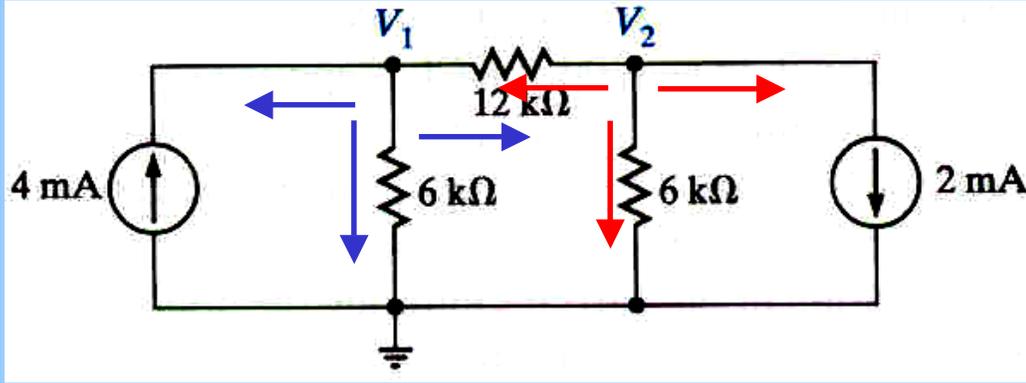
$$-v_1 \left(\frac{1}{R_1} \right) - v_2 \left(\frac{1}{R_4} \right) + v_3 \left(\frac{1}{R_1} + \frac{1}{R_4} + \frac{1}{R_5} \right) = 0$$

VALID ONLY FOR CIRCUITS WITHOUT DEPENDENT SOURCES

$$\begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} & 0 & -\frac{1}{R_1} \\ 0 & \frac{1}{R_3} + \frac{1}{R_4} & -\frac{1}{R_4} \\ -\frac{1}{R_1} & -\frac{1}{R_4} & \frac{1}{R_1} + \frac{1}{R_4} + \frac{1}{R_5} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} -i_A \\ i_A - i_B \\ 0 \end{bmatrix}$$

LEARNING EXTENSION

Write the node equations



$$\text{@ } V_1: -4mA + \frac{V_1}{6k} + \frac{V_1 - V_2}{12k} \quad \text{USING KCL}$$

$$\text{@ } V_2: 2mA + \frac{V_2}{6k} + \frac{V_2 - V_1}{12k} = 0$$

BY "INSPECTION"

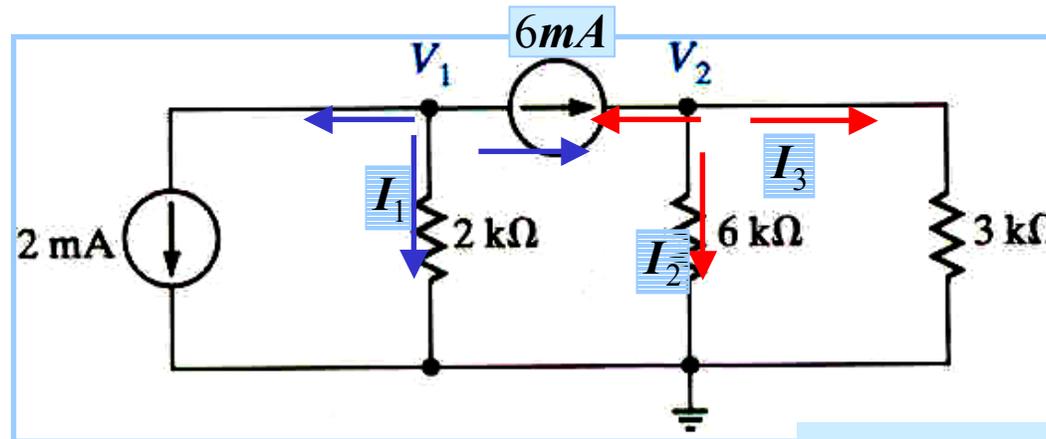
$$\left(\frac{1}{6k} + \frac{1}{12k}\right)V_1 - \frac{1}{12k}V_2 = 4mA$$

$$-\frac{1}{12k} + \left(\frac{1}{6k} + \frac{1}{12k}\right)V_2 = -2mA$$



LEARNING EXTENSION

Find all the **branch** currents



$$KCL @ V_1: I_1 + 2\text{ mA} + 6\text{ mA} = 0$$

$$V_2 = 12\text{ V}$$

$$@V_2: -6\text{ mA} + \frac{V_2}{6\text{ k}} + \frac{V_2}{3\text{ k}} = 0 \Rightarrow I_2 = 2\text{ mA}$$

$$I_3 = 4\text{ mA}$$

NODE EQS. BY INSPECTION

$$\frac{1}{2\text{ k}}V_1 + (0)V_2 = -(2 + 6)\text{ mA}$$

$$(0)V_1 + \left(\frac{1}{6\text{ k}} + \frac{1}{3\text{ k}}\right)V_2 = 6\text{ mA}$$

$$I_1 = \frac{V_1}{2\text{ k}} \quad I_2 = \frac{V_2}{6\text{ k}} \quad I_3 = \frac{V_2}{3\text{ k}}$$

IN MOST CASES THERE ARE SEVERAL DIFFERENT WAYS OF SOLVING A PROBLEM

$$I_2 = \frac{3\text{ k}}{3\text{ k} + 6\text{ k}}(6\text{ mA}) = 2\text{ mA}$$

$$I_3 = \frac{6\text{ k}}{3\text{ k} + 6\text{ k}}(6\text{ mA}) = 4\text{ mA}$$

CURRENTS COULD BE COMPUTED DIRECTLY USING CURRENT DIVIDER!!



CIRCUITS WITH DEPENDENT SOURCES

NUMERICAL EXAMPLE

CIRCUITS WITH DEPENDENT SOURCES CANNOT BE MODELED BY INSPECTION. THE SYMMETRY IS LOST.

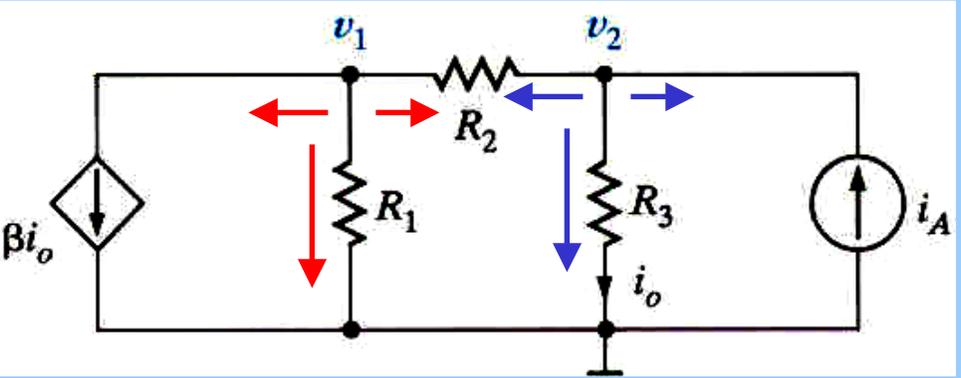
$\beta = 2$ $R_2 = 6 \text{ k}\Omega$ $i_A = 2 \text{ mA}$
 $R_1 = 12 \text{ k}\Omega$ $R_3 = 3 \text{ k}\Omega$

A PROCEDURE FOR MODELING

- WRITE THE NODE EQUATIONS USING DEPENDENT SOURCES AS REGULAR SOURCES.
- FOR EACH DEPENDENT SOURCE WE ADD ONE EQUATION EXPRESSING THE CONTROLLING VARIABLE IN TERMS OF THE NODE VOLTAGES

$$\left(\frac{1}{12k} + \frac{1}{6k}\right)v_1 + \left(\frac{2}{3k} - \frac{1}{6k}\right)v_2 = 0$$

$$-\frac{1}{6k}v_1 + \left(\frac{1}{12k} + \frac{1}{3k}\right)v_2 = 2mA$$



$$\frac{1}{4k}V_1 + \frac{1}{2k}V_2 = 0 \quad */4k$$

$$-\frac{1}{6k}V_1 + \frac{1}{2k}V_2 = 2 \times 10^{-3} \quad */6k$$

$$V_1 + 2V_2 = 0$$

$$-V_1 + 3V_2 = 12[V]$$

ADDING THE EQUATIONS $5V_2 = 12[V]$

$$V_1 = -\frac{24}{5}V$$

$$\beta i_o + \frac{v_1}{R_1} + \frac{v_1 - v_2}{R_2} = 0$$

$$-i_A + \frac{v_2}{R_3} + \frac{v_2 - v_1}{R_2} = 0$$

REPLACE AND REARRANGE

$$\left(\frac{1}{R_1} + \frac{1}{R_2}\right)v_1 + \left(\frac{\beta}{R_3} - \frac{1}{R_2}\right)v_2 = 0$$

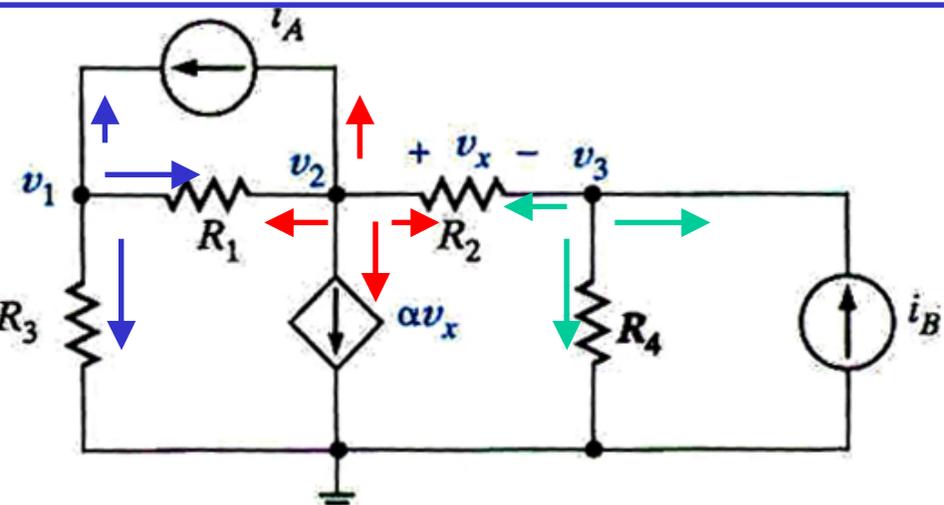
$$-\frac{1}{R_2}v_1 + \left(\frac{1}{R_2} + \frac{1}{R_3}\right)v_2 = i_A$$

MODEL FOR CONTROLLING VARIABLE

$$i_o = \frac{v_2}{R_3}$$



LEARNING EXAMPLE: CIRCUIT WITH VOLTAGE-CONTROLLED CURRENT



REPLACE AND REARRANGE

$$\begin{aligned} (G_1 + G_3)v_1 - G_1v_2 &= i_A \\ -G_1v_1 + (G_1 + \alpha + G_2)v_2 - (\alpha + G_2)v_3 &= -i_A \\ -G_2v_2 + (G_2 + G_4)v_3 &= i_B \end{aligned}$$

CONTINUE WITH GAUSSIAN ELIMINATION...

WRITE NODE EQUATIONS. TREAT DEPENDENT SOURCE AS REGULAR SOURCE

$$G_3v_1 + G_1(v_1 - v_2) - i_A = 0$$

$$i_A + G_1(v_2 - v_1) + \alpha v_x + G_2(v_2 - v_3) = 0$$

$$G_2(v_3 - v_2) + G_4v_3 - i_B = 0$$

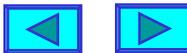
OR USE MATRIX ALGEBRA

$$\begin{bmatrix} (G_1 + G_3) & -G_1 & 0 \\ -G_1 & (G_1 + \alpha + G_2) & -(\alpha + G_2) \\ 0 & -G_2 & (G_2 + G_4) \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} i_A \\ -i_A \\ i_B \end{bmatrix}$$

EXPRESS CONTROLLING VARIABLE IN TERMS OF NODE VOLTAGES

$$v_x = v_2 - v_3$$

FOUR EQUATIONS IN OUR UNKNOWN. SOLVE USING FAVORITE TECHNIQUE



USING MATLAB TO SOLVE THE NODE EQUATIONS

$$\begin{bmatrix} (G_1 + G_3) & -G_1 & 0 \\ -G_1 & (G_1 + \alpha + G_2) & -(\alpha + G_2) \\ 0 & -G_2 & (G_2 + G_4) \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \end{bmatrix} = \begin{bmatrix} i_A \\ -i_A \\ i_B \end{bmatrix}$$

$$R_1 = 1k\Omega, R_2 = R_3 = 2k\Omega, \\ R_4 = 4k\Omega, i_A = 2mA, i_B = 4mA, \\ \alpha = 2[A/V]$$

DEFINE THE COMPONENTS OF THE CIRCUIT

```
» R1=1000;R2=2000;R3=2000;
R4=4000; %resistances in Ohm
» iA=0.002;iB=0.004; %sources in Amps
» alpha=2; %gain of dependent source
```

DEFINE THE MATRIX G

Entries in a row are separated by commas (or plain spaces). Rows are separated by semi colon

```
» G=[(1/R1+1/R2), -1/R1, 0; %first row of the matrix
-1/R1, (1/R1+alpha+1/R2), -(alpha+1/R2); %second row
0, -1/R2, (1/R2+1/R4)]; %third row. End in comma to have the echo
```

G =

```
0.0015 -0.0010 0
-0.0010 2.0015 -2.0005
0 -0.0005 0.0008
```

DEFINE RIGHT HAND SIDE VECTOR

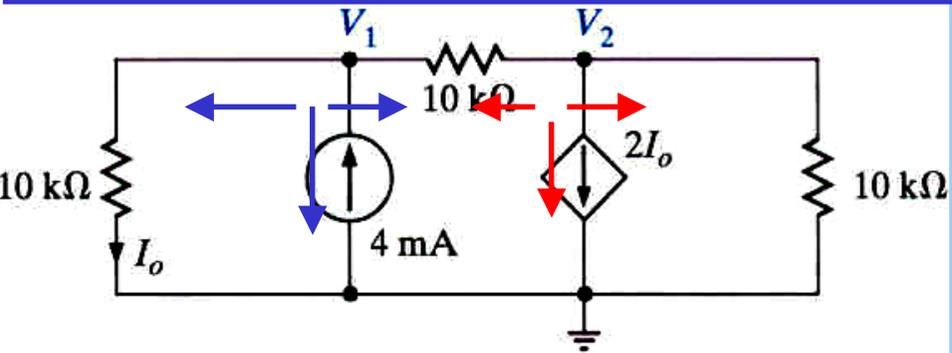
```
» I=[iA;-iA;iB]; %end in ";" to skip echo
```

SOLVE LINEAR EQUATION

```
» V=G\I % end with carriage return and get the echo
```

```
V =
11.9940
15.9910
15.9940
```

LEARNING EXTENSION: FIND NODE VOLTAGES



NODE EQUATIONS

$$\text{@}V_1: \frac{V_1}{10k} - 4mA + \frac{V_1 - V_2}{10k} = 0$$

$$\text{@}V_2: \frac{V_2 - V_1}{10k} + 2I_o + \frac{V_2}{10k} = 0$$

CONTROLLING VARIABLE (IN TERMS OF NODE VOLTAGES)

$$I_o = \frac{V_1}{10k}$$

REPLACE

$$\frac{V_1}{10k} - 4mA + \frac{V_1 - V_2}{10k} = 0$$

$$\frac{V_2 - V_1}{10k} + 2 \frac{V_1}{10k} + \frac{V_2}{10k} = 0$$

REARRANGE AND MULTIPLY BY 10k

$$2V_1 - V_2 = 40[V] \quad */2 \text{ and add eqs.}$$

$$V_1 + 2V_2 = 0$$

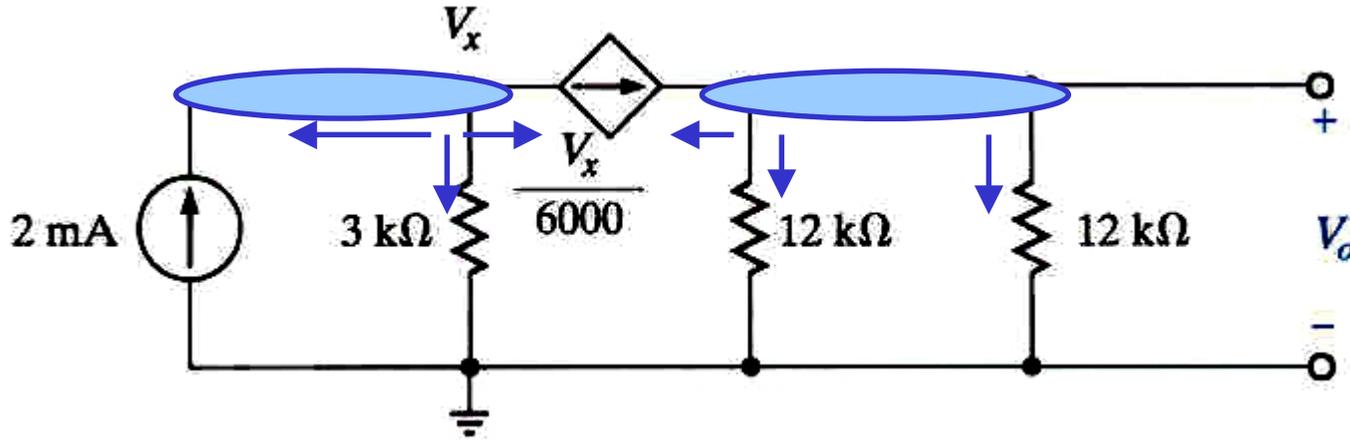
$$5V_1 = 80V \Rightarrow V_1 = 16V$$

$$V_2 = -\frac{V_1}{2} \Rightarrow V_2 = -8V$$



FIND THE VOLTAGE V_o

LEARNING EXTENSION



NODE EQUATIONS

NOTICE REPLACEMENT OF DEPENDENT SOURCE
IN TERMS OF NODE VOLTAGE

$$-2mA + \frac{V_x}{3k} + \frac{V_x}{6k} = 0 \quad */6k$$

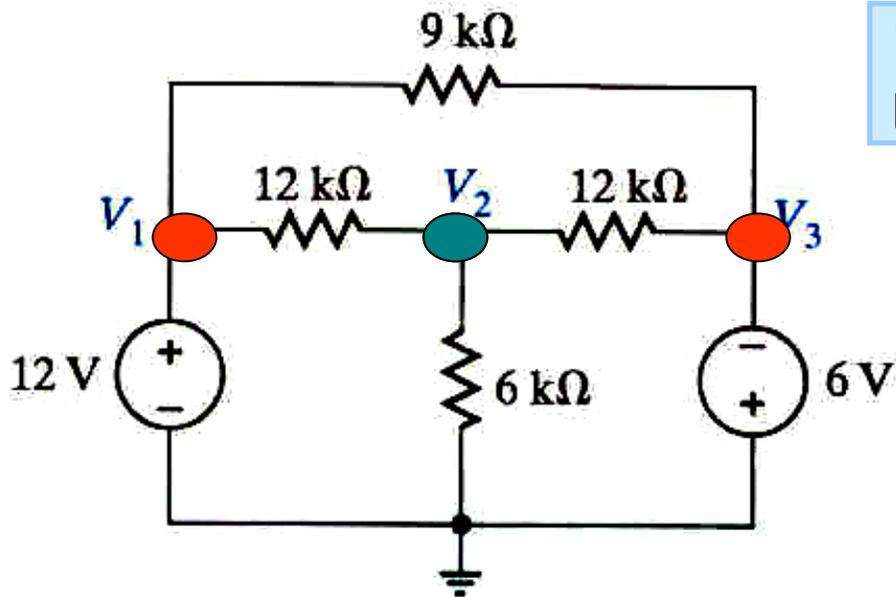
$$-\frac{V_x}{6k} + \frac{V_o}{12k} + \frac{V_o}{12k} = 0 \quad */12k$$

$$3V_x = 12[V] \Rightarrow V_x = 4[V]$$

$$2V_o - 2V_x = 0 \Rightarrow V_o = 4[V]$$



CIRCUITS WITH INDEPENDENT VOLTAGE SOURCES



Hint: Each voltage source connected to the reference node saves one node equation

One more example

3 nodes plus the reference. In principle one needs 3 equations...

...but two nodes are connected to the reference through voltage sources. Hence those node voltages are known!!!

...only one KCL is necessary

$$\frac{V_2}{6k} + \frac{V_2 - V_3}{12k} + \frac{V_2 - V_1}{12k} = 0$$

$$V_1 = 12[V]$$

$$V_3 = -6[V]$$

THESE ARE THE REMAINING TWO NODE EQUATIONS

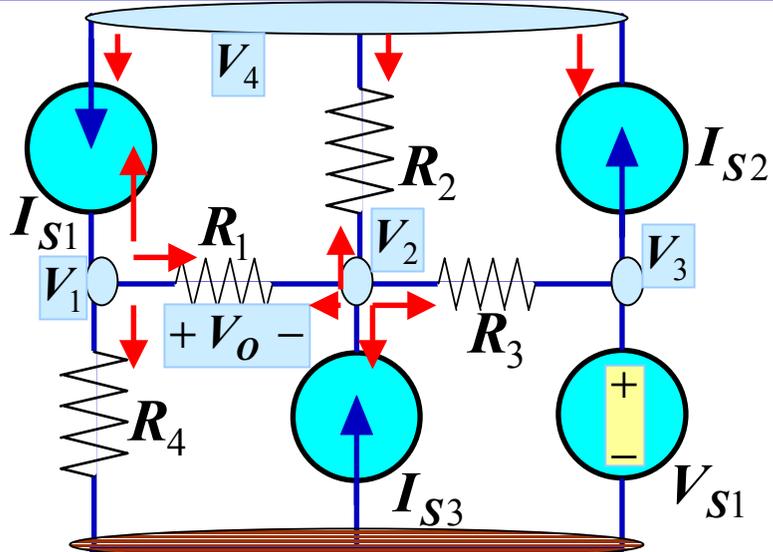
SOLVING THE EQUATIONS

$$2V_2 + (V_2 - V_3) + (V_2 - V_1) = 0$$

$$4V_2 = 6[V] \Rightarrow V_2 = 1.5[V]$$



Problem 3.67 (6th Ed) Find V_0



$R_1 = 1k$; $R_2 = 2k$, $R_3 = 1k$, $R_4 = 2k$
 $I_{S1} = 2mA$, $I_{S2} = 4mA$, $I_{S3} = 4mA$,
 $V_S = 12V$

KNOWN NODE VOLTAGE

@ V_3 : $V_3 = V_{VS} = 12[V]$

@ V_1 : $-I_{S1} + \frac{V_1 - V_2}{R_1} + \frac{V_1}{R_4} = 0$
 $-2[mA] + \frac{V_1 - V_2}{1k} + \frac{V_1}{2k} = 0$

@ V_2 : $-I_{S3} + \frac{V_2 - V_1}{R_1} + \frac{V_2 - V_3}{R_3} + \frac{V_2 - V_4}{R_2} = 0$
 $-4[mA] + \frac{V_2 - V_1}{1k} + \frac{V_2 - 12}{1k} + \frac{V_2 - V_4}{2k} = 0$

@ V_4 : $I_{S1} - I_{S2} + \frac{V_4 - V_2}{R_2} = 0$
 $2[mA] - 4[mA] + \frac{V_4 - V_2}{2k} = 0$

IDENTIFY AND LABEL ALL NODES

WRITE THE NODE EQUATIONS

NOW WE LOOK WHAT IS BEING ASKED TO DECIDE THE SOLUTION STRATEGY.

$V_0 = V_1 - V_2$

ONLY V_1, V_2 ARE NEEDED FOR V_0



TO SOLVE BY HAND ELIMINATE DENOMINATORS

$$-2[mA] + \frac{V_1 - V_2}{1k} + \frac{V_1}{2k} = 0$$

$\xrightarrow{*/2k}$

$$3V_1 - 2V_2 = 4[V] \quad (1)$$

$$-4[mA] + \frac{V_2 - V_1}{1k} + \frac{V_2 - 12}{1k} + \frac{V_2 - V_4}{2k} = 0$$

$\xrightarrow{*/2k}$

$$-2V_1 + 5V_2 - V_4 = 32[V] \quad (2)$$

$$2[mA] - 4[mA] + \frac{V_4 - V_2}{2k} = 0$$

$\xrightarrow{*/2k}$

$$-V_2 + V_4 = 4[V] \quad (3)$$

Add 2+3 $\underline{-2V_1 + 4V_2 = 36[V]}$

ALTERNATIVE: USE LINEAR ALGEBRA

$$\begin{bmatrix} 3 & -2 & 0 \\ -2 & 5 & -1 \\ 0 & -1 & 1 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} 4 \\ 32 \\ 4 \end{bmatrix}$$

$$3V_1 - 2V_2 = 4[V] \quad */2 \text{ and add}$$

$$4V_1 = 40[V] \Rightarrow V_1 = 10[V]$$

$$4V_2 = 56[V] \Rightarrow V_2 = 14[V]$$

FINALLY!! $V_0 = V_1 - V_2 = -4[V]$

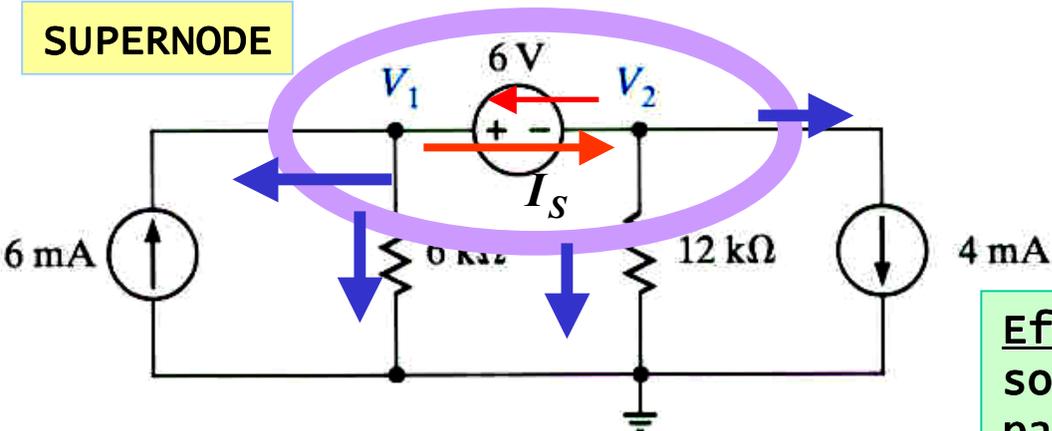
So. what happens when sources are connected between two non reference nodes?



THE SUPERNODE TECHNIQUE

We will use this example to introduce the concept of a SUPERNODE

SUPERNODE



Conventional node analysis requires all currents at a node

@V₁

$$-6mA + \frac{V_1}{6k} + I_S = 0$$

@V₂

$$-I_S + 4mA + \frac{V_2}{12k} = 0$$

Efficient solution: enclose the source, and all elements in parallel, inside a surface.

Apply KCL to the surface!!!

$$-6mA + \frac{V_1}{6k} + \frac{V_2}{12k} + 4mA = 0$$

The source current is interior to the surface and is not required

We STILL need one more equation

$$V_1 - V_2 = 6[V]$$

2 eqs, 3 unknowns...Panic!!

The current through the source is not related to the voltage of the source

Math solution: add one equation

$$V_1 - V_2 = 6[V]$$

Only 2 eqs in two unknowns!!!



The Equations

$$(1) \quad \frac{V_1}{6k} + \frac{V_2}{12k} - 6mA + 4mA = 0 \quad * / 12k$$

$$(2) \quad V_1 - V_2 = 6[V]$$

Solution

1. Eliminate denominators in Eq(1). Multiply by ...

$$2V_1 + V_2 = 24[V]$$

$$V_1 - V_2 = 6[V]$$

2. Add equations to eliminate V_2

$$3V_1 = 30[V] \Rightarrow V_1 = 10[V]$$

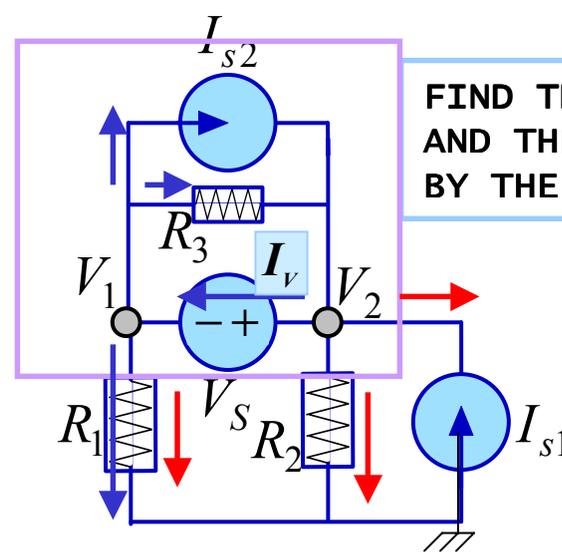
3. Use Eq(2) to compute V_2

$$V_2 = V_1 - 6[V] = 4[V]$$



The supernode technique

- Used when a branch between two nonreference nodes contains a voltage source.
- First encircle the voltage source and the two connecting nodes to form the supernode.
- Write the equation that defines the voltage relationship between the two nonreference nodes as a result of the presence of the voltage source.
- Write the KCL equation for the supernode.
- If the voltage source is dependent, then the controlling equation for the dependent source is also needed.



FIND THE NODE VOLTAGES AND THE POWER SUPPLIED BY THE VOLTAGE SOURCE

$$R_1 = R_2 = 10k\Omega, R_3 = 4k\Omega$$

$$V_S = 20[V], I_{s1} = 10[mA], I_{s2} = 6[mA]$$

$$V_2 - V_1 = 20$$

$$\Rightarrow -V_1 + V_2 = 20[V]$$

$$\frac{V_1}{10k} + \frac{V_2}{10k} - 10mA = 0$$

$$*/10k \Rightarrow V_1 + V_2 = 100[V]$$

$$\text{adding: } V_2 = 60[V]$$

$$V_1 = 100 - V_2 = 40[V]$$

TO COMPUTE THE POWER SUPPLIED BY VOLTAGE SOURCE WE MUST KNOW THE CURRENT THROUGH IT

$$I_v = \frac{V_1}{10k} + 6mA + \frac{V_1 - V_2}{10k} = 8mA$$

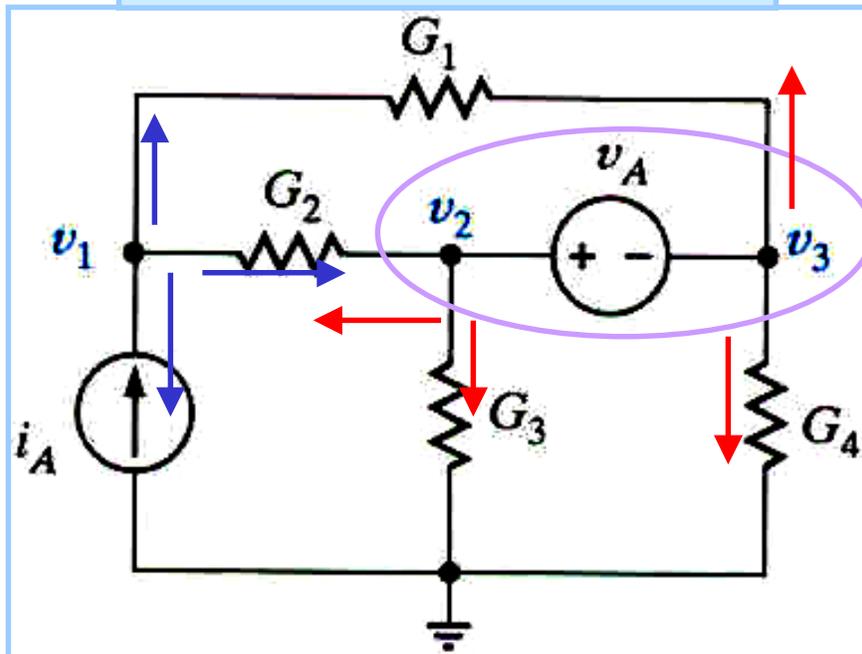
$$P = 20[V] \times 8[mA] = 160mW$$

BASED ON PASSIVE SIGN CONVENTION THE POWER IS RECEIVED BY THE SOURCE!!



LEARNING EXAMPLE

WRITE THE NODE EQUATIONS



$$\text{@ } v_1 \quad (v_1 - v_3)G_1 + (v_1 - v_2)G_2 - i_A = 0$$

@ SUPERNODE

$$\text{CONSTRAINT: } v_2 - v_3 = v_A$$

KCL (leaving supernode):

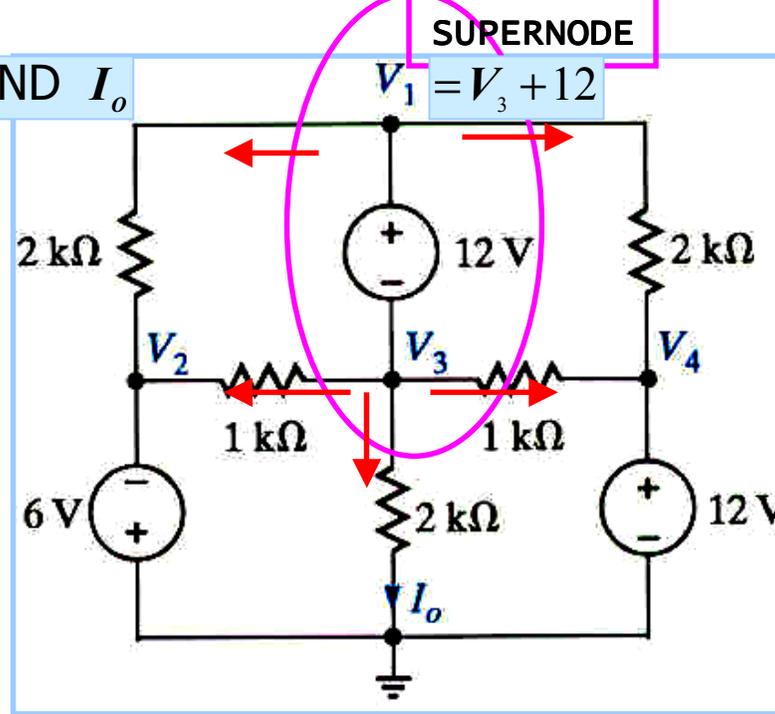
$$(v_2 - v_1)G_2 + v_2G_3 + (v_3 - v_1)G_1 + v_3G_4 = 0$$

THREE EQUATIONS IN THREE UNKNOWNNS



LEARNING EXAMPLE

FIND I_o



$V_2 = -6V, V_4 = 12V$ KNOWN NODE VOLTAGES

SUPERNODE CONSTRAINT $\Rightarrow V_1 - V_3 = 12$

$$\frac{V_3 + 12 - (-6)}{2k} + \frac{V_3 + 12 - 12}{2k} + \frac{V_3 - (-6)}{1k} + \frac{V_3 - 12}{1k} + \frac{V_3}{2k} = 0$$

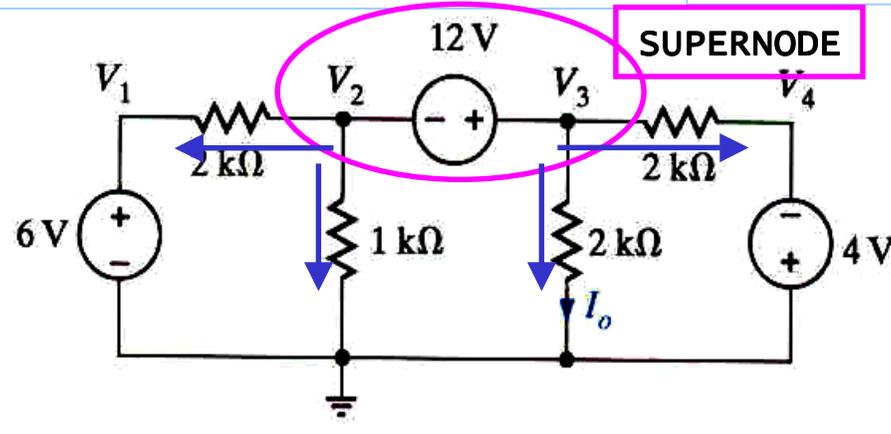
KCL @ SUPERNODE

$$V_3 = -\frac{6}{7}V \Rightarrow I_o = \frac{-\frac{6}{7}}{2k} = -\frac{3}{7}mA$$



LEARNING EXTENSION

Use nodal analysis to find I_o



$$V_1 = 6V$$

$$V_4 = -4V$$

SOURCES CONNECTED TO THE REFERENCE

CONSTRAINT EQUATION $V_3 - V_2 = 12V$

KCL @ SUPERNODE

$$\frac{V_2 - 6}{2k} + \frac{V_2}{1k} + \frac{V_3}{2k} + \frac{V_3 - (-4)}{2k} = 0 \quad */2k$$

V_2 IS NOT NEEDED FOR I_o

$$3V_2 + 2V_3 = 2V$$

$$-V_2 + V_3 = 12V \quad */3 \text{ and add}$$

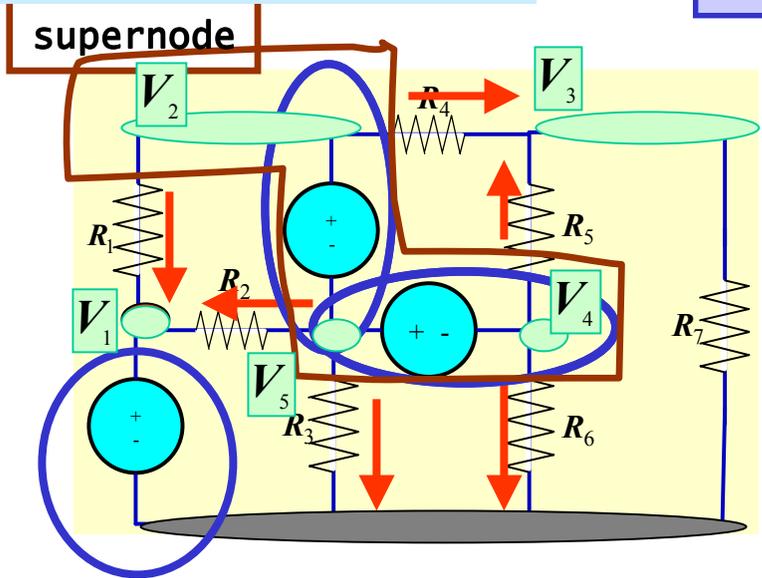
$$5V_3 = 38V$$

OHM'S LAW $I_o = \frac{V_3}{2k} = 3.8mA$



Supernodes can be more complex

WRITE THE NODE EQUATIONS



KCL@V_3

$$\frac{V_3 - V_2}{R_4} + \frac{V_3 - V_4}{R_5} + \frac{V_3}{R_7} = 0$$

KCL @SUPERNODE
(Careful not to omit any current)

$$\frac{V_2 - V_1}{R_1} + \frac{V_5 - V_1}{R_2} + \frac{V_5}{R_3} + \frac{V_4}{R_6} + \frac{V_4 - V_3}{R_5} + \frac{V_2 - V_3}{R_4} = 0$$

CONSTRAINTS DUE TO VOLTAGE SOURCES

$$V_1 = V_{S1}$$

$$V_2 - V_5 = V_{S2}$$

$$V_5 - V_4 = V_{S3}$$

5 EQUATIONS IN FIVE UNKNOWNNS.

Identify all nodes, select a reference and label nodes

Nodes connected to reference through a voltage source

Voltage sources in between nodes and possible supernodes

EQUATION BOOKKEEPING:
KCL@ V_3, KCL@ supernode,
2 constraints equations
and one known node



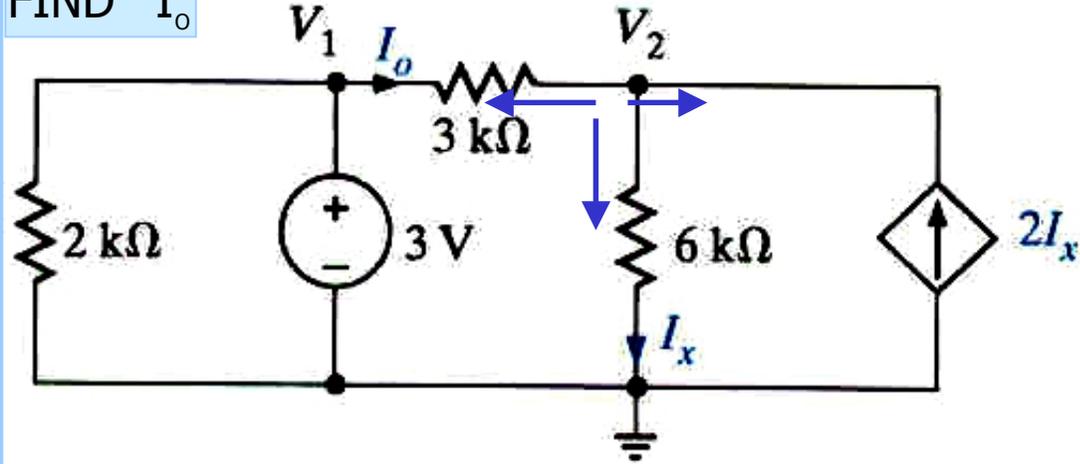
**CIRCUITS WITH DEPENDENT SOURCES
PRESENT NO SIGNIFICANT ADDITIONAL
COMPLEXITY. THE DEPENDENT SOURCES
ARE TREATED AS REGULAR SOURCES**

**WE MUST ADD ONE EQUATION FOR EACH
CONTROLLING VARIABLE**



LEARNING EXAMPLE

FIND I_o



VOLTAGE SOURCE CONNECTED TO REFERENCE

$$V_1 = 3V$$

$$\text{KCL@ } V_2: \frac{V_2 - V_1}{3k} + \frac{V_2}{6k} - 2I_x = 0$$

REPLACE

CONTROLLING VARIABLE IN TERMS OF NODE VOLTAGES

$$I_x = \frac{V_2}{6k}$$

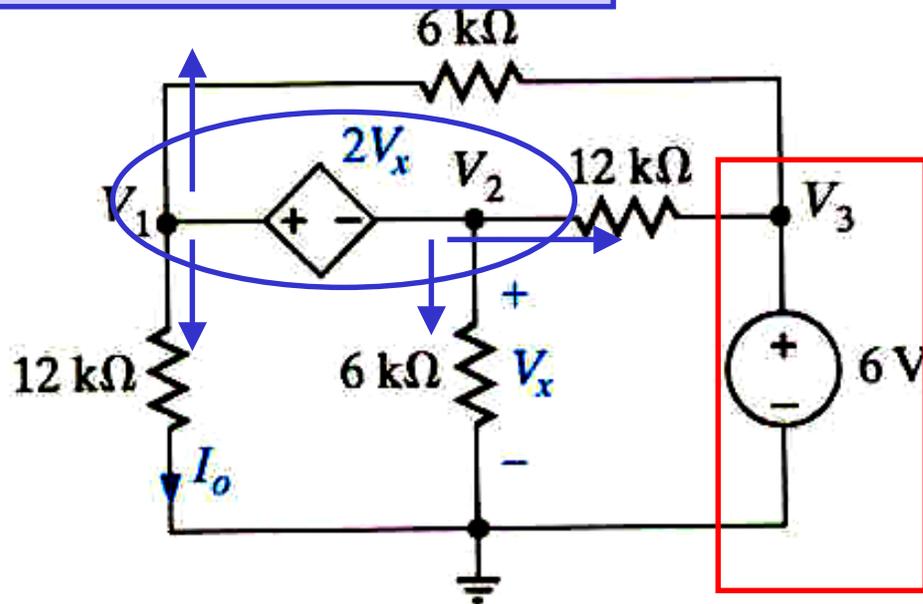
$$\frac{V_2 - V_1}{3k} + \frac{V_2}{6k} - 2 \frac{V_2}{6k} = 0 \quad */6k$$

$$V_2 - 2V_1 = 0 \Rightarrow V_2 = 6V$$

$$I_o = \frac{V_1 - V_2}{3k} = -1mA$$



SUPER NODE WITH DEPENDENT SOURCE



VOLTAGE SOURCE CONNECTED TO REFERENCE

$$V_3 = 6V$$

SUPERNODE CONSTRAINT $V_1 - V_2 = 2V_x$

CONTROLLING VARIABLE IN TERMS OF NODES

KCL AT SUPERNODE $V_x = V_2 \Rightarrow V_1 = 3V_2$

$$\frac{V_1 - V_3}{6k} + \frac{V_1}{12k} + \frac{V_2}{6k} + \frac{V_2 - V_3}{12k} = 0 \quad */12k$$

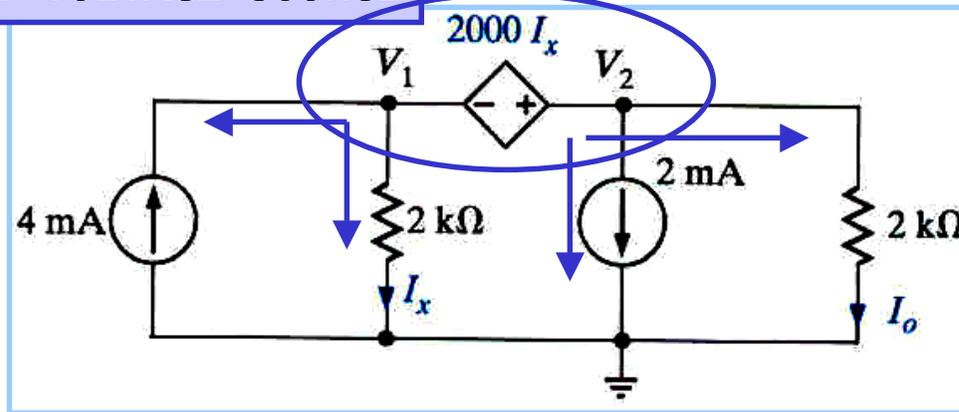
$$2(V_1 - 6) + V_1 + 2V_2 + V_2 - 6 = 0$$

$$3V_1 + 3V_2 = 18 \Rightarrow 4V_1 = 18$$

$$I_o = \frac{V_1}{12k} = \frac{3}{8} \text{ mA}$$



CURRENT CONTROLLED VOLTAGE SOURCE



CONSTRAINT DUE TO SOURCE $V_2 - V_1 = 2kI_x$

CONTROLLING VARIABLE IN TERMS OF NODES

$$\Rightarrow V_1 = 2kI_x \Rightarrow V_2 = 2V_1 \quad I_x = \frac{V_1}{2k}$$

KCL AT SUPERNODE $-4mA + \frac{V_1}{2k} + 2mA + \frac{V_2}{2k} = 0$

$$V_1 + V_2 = 4[V] \quad */2 \text{ and add}$$

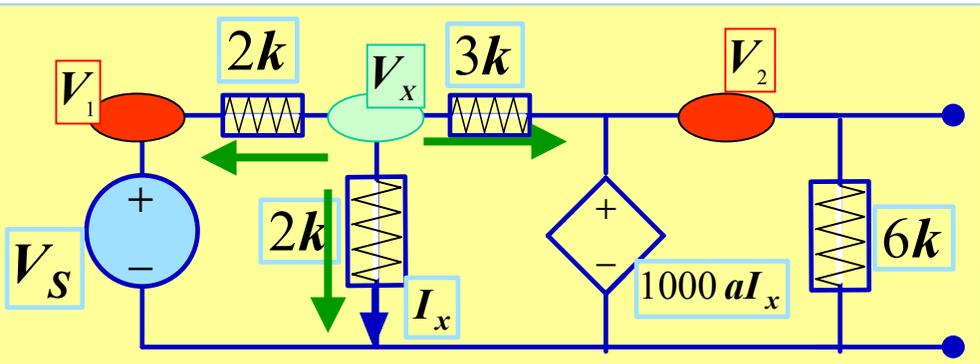
$$-2V_1 + V_2 = 0$$

$$3V_2 = 8[V]$$

$$I_o = \frac{V_2}{2k} = \frac{4}{3} mA$$



An example with dependent sources



'a' has units of [volt/Amp]

IDENTIFY AND LABEL NODES

2 nodes are connected to the reference through voltage sources

$$V_1 = V_S$$

$$V_2 = 1000aI_x$$

KCL @ v_x

$$\frac{V_x - V_S}{2k} + \frac{V_x}{2k} + \frac{V_x - v_2}{3k} = 0$$

EXPRESS CONTROLLING VARIABLE IN TERMS OF NODE VOLTAGES

$$I_x = \frac{V_x}{2k}$$

REPLACE I_x IN V2

$$V_2 = \frac{1k * aV_x}{2k}$$

$$V_2 = \frac{aV_x}{2}$$

REPLACE V2 IN KCL

$$3(V_x - V_S) + 3V_x + 2(V_x - aV_x/2) = 0$$

$$(8 - a)V_x = 3V_S$$

what happens when a=8?

