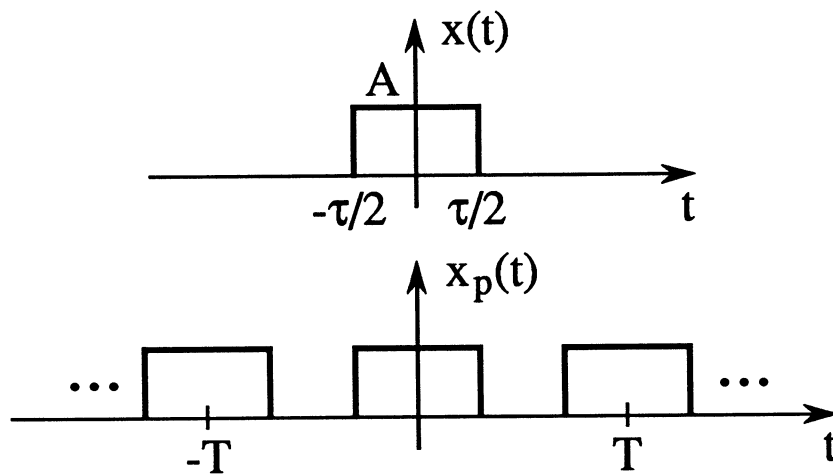


1.3.2 CONTINUOUS-TIME FOURIER TRANSFORM (CTFT)

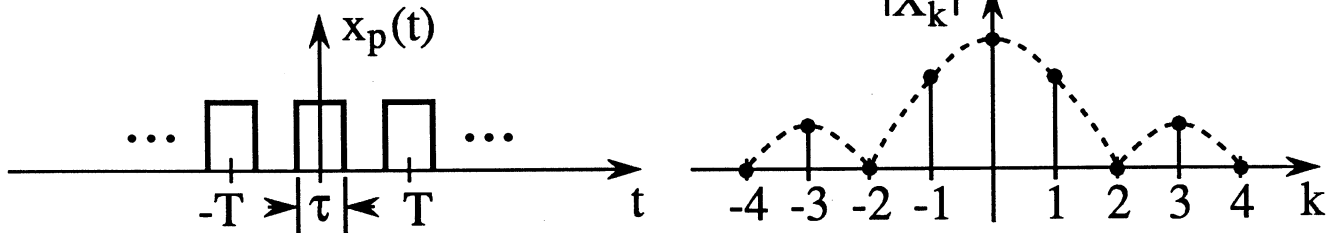
Spectral representation for *aperiodic* CT signals

Consider a fixed signal $x(t)$ and let $x_p(t) = \text{rep}_T[x(t)]$

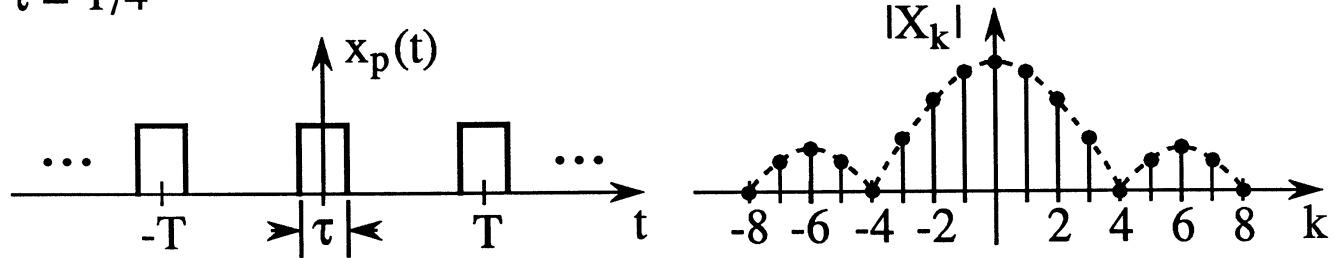


What happens to Fourier series as T increases?

$\tau = T/2$



$\tau = T/4$



Fourier Coefficients

$$X_k = \int_{-T/2}^{T/2} x(t) e^{-j2\pi kt/T} dt$$

Let $T \rightarrow \infty$

$k/T \rightarrow f$

$X_k \rightarrow X(f)$

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

Fourier Series Expansion

$$x_p(t) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_k e^{j2\pi kt/T}$$

Let $T \rightarrow \infty$

$$x_p(t) \rightarrow x(t)$$

$$k/T \rightarrow f \quad X_k \rightarrow X(f)$$

$$\frac{1}{T} \sum_{k=-\infty}^{\infty} \rightarrow \int_{-\infty}^{\infty} df$$

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df$$

Fourier Transform Pair

Forward transform

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt \quad (1)$$

Inverse transform

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df \quad (2)$$

Sufficient Conditions for Existence of CTFT

1. $x(t)$ has finite energy

$$\int_{-\infty}^{\infty} |x(t)|^2 dt < \infty$$

2. $x(t)$ is absolutely integrable

$$\int_{-\infty}^{\infty} |x(t)| dt < \infty$$

and it satisfies the Dirichlet conditions

Transform Relations

1. linearity

$$a_1 x_1(t) + a_2 x_2(t) \stackrel{\text{CTFT}}{\leftrightarrow} a_1 X_1(f) + a_2 X_2(f)$$

2. scaling and shifting

$$x\left(\frac{t-t_0}{a}\right) \stackrel{\text{CTFT}}{\leftrightarrow} |a| X(af) e^{-j2\pi ft_0}$$

3. modulation

$$x(t) e^{j2\pi f_0 t} \stackrel{\text{CTFT}}{\leftrightarrow} X(f - f_0)$$

4. reciprocity

$$\text{CTFT} \\ X(f) \leftrightarrow x(t)$$

5. Parseval's relation

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(f)|^2 df$$

6. Initial value

$$\int_{-\infty}^{\infty} x(t) dt = X(0)$$

Comments

1. Reflection is a special case of scaling and shifting with $a = -1$ and $t_0 = 0$, *i.e.*

$$\text{CTFT} \\ x(-t) \leftrightarrow X(-f)$$

2. The scaling relation exhibits reciprocal spreading.
3. Uniqueness of the CTFT follows from Parseval's relation.

CTFT for Real Signals

If $x(t)$ is real, $X(f) = [X(-f)]^*$

$$\Rightarrow |X(f)| = |X(-f)| \quad \text{and} \quad \underline{\angle X(f)} = - \underline{\angle X(-f)}$$

In this case, the inverse transform may be written as

$$x(t) = 2 \int_0^{\infty} |X(f)| \cos[2\pi ft + \underline{\angle X(f)}] df$$

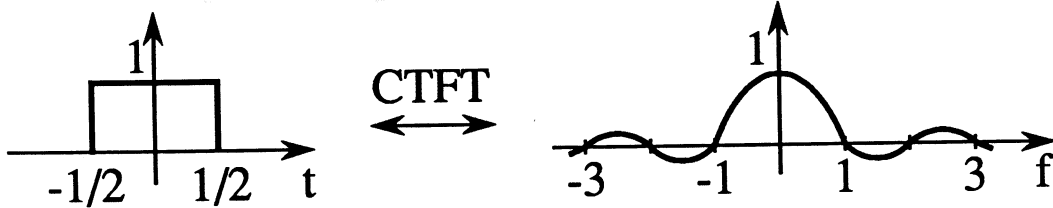
Additional symmetry relations:

$x(t)$ is real and even $\Leftrightarrow X(f)$ is real and even

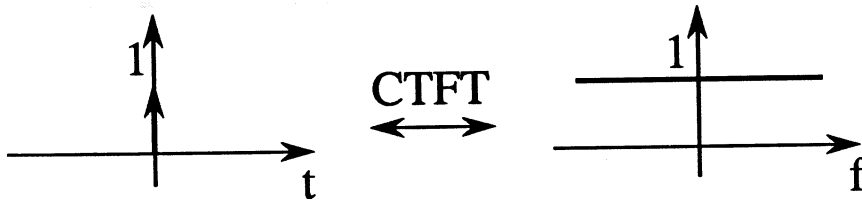
$x(t)$ is real and odd $\Leftrightarrow X(f)$ is imaginary and odd

Important Transform Pairs

CTFT
1. $\text{rect}(t) \leftrightarrow \text{sinc}(f)$



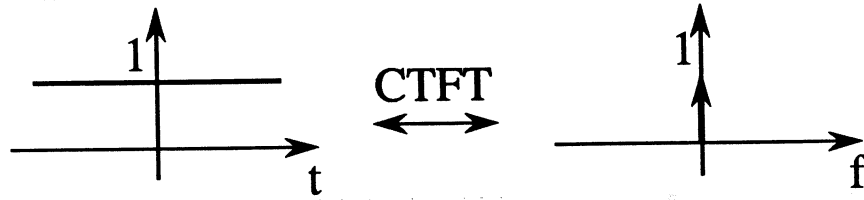
CTFT
2. $\delta(t) \leftrightarrow 1$ (by sifting property)



Proof:

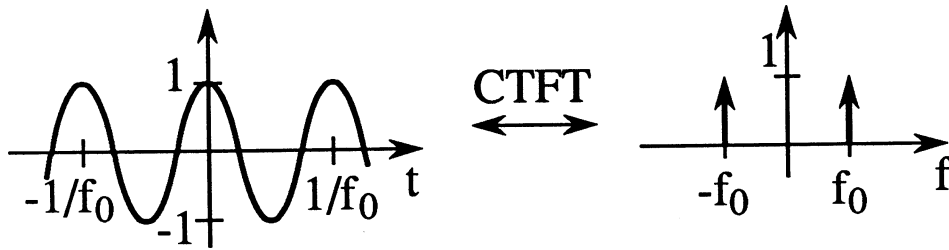
$$\mathcal{F}\{\delta(t)\} = \int_{-\infty}^{\infty} \delta(t) e^{-j2\pi ft} dt = 1$$

3. $1 \xleftrightarrow{\text{CTFT}} \delta(f)$ (by reciprocity)



4. $e^{j2\pi f_0 t} \xleftrightarrow{\text{CTFT}} \delta(f - f_0)$ (by modulation property)

5. $\cos(2\pi f_0 t) \xleftrightarrow{\text{CTFT}} \frac{1}{2}[\delta(f - f_0) + \delta(f + f_0)]$



Generalized Fourier Transform

Note that $\delta(t)$ is absolutely integrable but not square integrable.

$$\text{Consider } \delta_{\Delta}(t) = \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta}\right)$$

$$\int_{-\infty}^{\infty} |\delta_{\Delta}(t)| dt = 1$$

$$\int_{-\infty}^{\infty} |\delta_{\Delta}(t)|^2 dt = \frac{1}{\Delta}$$

$$\therefore \lim_{\Delta \rightarrow 0} \int_{-\infty}^{\infty} |\delta_{\Delta}(t)|^2 dt = \infty$$

The function $x(t) \equiv 1$ is neither absolutely nor square integrable; and the integral

$$\int_{-\infty}^{\infty} 1 e^{-j2\pi ft} dt$$

is undefined.

Even when neither condition for existence of the CTFT is satisfied, we may still be able to define a Fourier transform through a limiting process.

Let $x_n(t)$, $n = 0, 1, 2, \dots$ denote a sequence of functions each of which has a valid CTFT $X_n(f)$

Suppose that $\lim_{n \rightarrow \infty} x_n(t) = x(t)$, a function that

does not have a valid transform

If $X(f) = \lim_{n \rightarrow \infty} X_n(f)$ exists, we call it the *generalized* Fourier transform of $x(t)$ i.e.

$$x_0(t) \xleftrightarrow{\text{CTFT}} X_0(f)$$

$$x_1(t) \xleftrightarrow{\text{CTFT}} X_1(f)$$

$$x_2(t) \xleftrightarrow{\text{CTFT}} X_2(f)$$

\vdots \vdots

$$x(t) \xleftrightarrow{\text{GCTFT}} X(f)$$

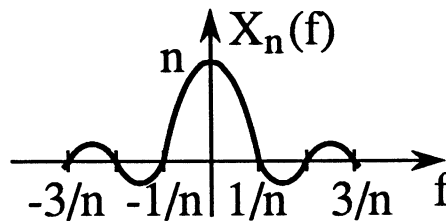
Example

$$\text{Let } x_n(t) = \text{rect}(t/n)$$

$$X_n(f) = n \text{ sinc}(nf) \quad (\text{by scaling})$$

$$\lim_{n \rightarrow \infty} x_n(t) = 1$$

What is $\lim_{n \rightarrow \infty} X_n(f)$?



$$X_n(f) \rightarrow 0, \quad f \neq 0$$

$$X_n(0) \rightarrow \infty$$

What is $\int_{-\infty}^{\infty} X_n(f)df$?

By the initial value relation

$$\int_{-\infty}^{\infty} X_n(f)df = x_n(0) = 1$$

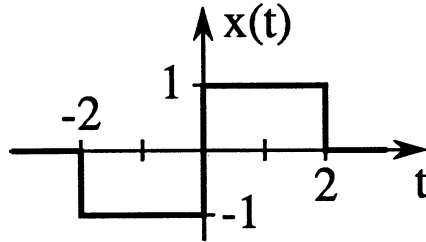
$$\therefore \lim_{n \rightarrow \infty} X_n(f) = \delta(f)$$

and we have

$$1 \stackrel{\text{GCTFT}}{\leftrightarrow} \delta(f)$$

Efficient Calculation of Fourier Transforms

Suppose we wish to determine the CTFT of the following signal



Brute force approach:

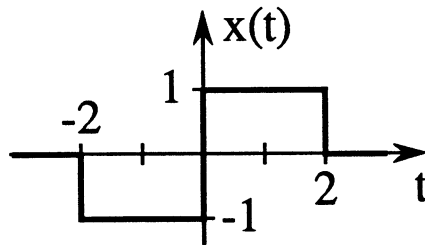
1. evaluate transform integral directly

$$X(f) = \int_{-2}^0 (-1) e^{-j2\pi ft} dt + \int_0^2 (1) e^{-j2\pi ft} dt$$

2. collect terms, simplify, etc...

Faster approach:

1. write $x(t)$ in terms of functions whose transforms are known

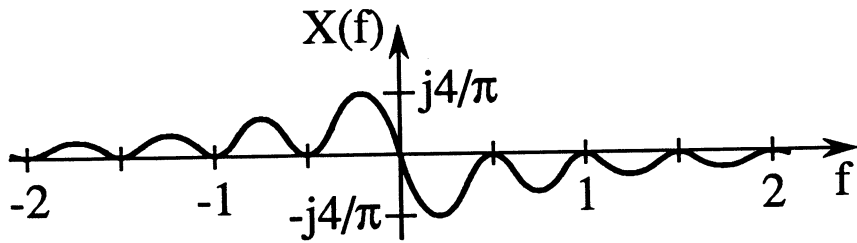
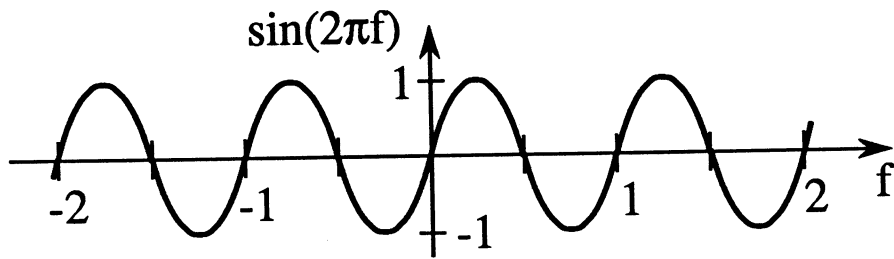
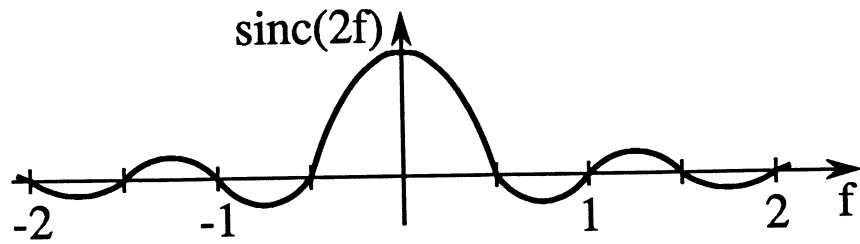


$$x(t) = -\text{rect}\left(\frac{t+1}{2}\right) + \text{rect}\left(\frac{t-1}{2}\right)$$

2. Use transform relations to determine $X(f)$

$$X(f) = 2 \text{sinc}(2f) [e^{-j2\pi f} - e^{j2\pi f}]$$

$$X(f) = -j4 \operatorname{sinc}(2f) \sin(2\pi f)$$



Comments

1. $A_x = 0$ and $X(0) = 0$
2. $x(t)$ is real and odd and $X(f)$ is imaginary and odd

CTFT and CT LTI Systems

The key factor that made it possible to express the response $y(t)$ of an LTI system to an *arbitrary* input $x(t)$ in terms of the impulse response $h(t)$ was the fact that we could write $x(t)$ as a superposition of impulses $\delta(t)$.

We can similarly express $x(t)$ as a superposition of complex exponential signals:

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df$$

Let $\tilde{H}(f)$ denote the frequency response of the system, *i.e.* for a fixed frequency f

$$e^{j2\pi ft} \rightarrow \boxed{\text{System}} \rightarrow \tilde{H}(f) e^{j2\pi ft}$$

then by homogeneity

$$X(f) e^{j2\pi ft} \rightarrow \boxed{\text{System}} \rightarrow \tilde{H}(f) X(f) e^{j2\pi ft}$$

and by superposition

$$\int_{-\infty}^{\infty} X(f) e^{j2\pi ft} dt \rightarrow \boxed{\text{System}} \rightarrow \int_{-\infty}^{\infty} \tilde{H}(f)X(f) e^{j2\pi ft} dt$$

Thus, the response to $x(t)$ is

$$y(t) = \int_{-\infty}^{\infty} \tilde{H}(f)X(f) e^{j2\pi ft} dt$$

But also,

$$y(t) = \int_{-\infty}^{\infty} Y(f) e^{j2\pi ft} dt$$

$$\therefore Y(f) = \tilde{H}(f) X(f) \quad (1)$$

We also know that

$$y(t) = \int_{-\infty}^{\infty} h(t - \tau)x(\tau)d\tau$$

What is relation between $h(t)$ and $\tilde{H}(f)$?

$$\begin{aligned} \text{Let } x(t) &= \delta(t) \Rightarrow y(t) = h(t) \\ \text{then } X(f) &= 1 \quad \text{and} \quad Y(f) = H(f) \end{aligned}$$

From Eq. (1), conclude that $\tilde{H}(f) = H(f)$

Since the frequency response is the CTFT of the impulse response, we will drop the tilde.

Summarizing, we have two equivalent characterizations for CT LTI systems

$$y(t) = \int_{-\infty}^{\infty} h(t - \tau) x(\tau) d\tau$$

$$Y(f) = H(f) X(f)$$

Convolution Theorem

Since $x(t)$ and $h(t)$ are arbitrary signals, we also have the following Fourier transform relation

$$\int x_1(\tau) x_2(t - \tau) d\tau \stackrel{\text{CTFT}}{\leftrightarrow} X_1(f) X_2(f)$$

or

$$x_1(t) * x_2(t) \stackrel{\text{CTFT}}{\leftrightarrow} X_1(f) X_2(f)$$

Product Theorem

By reciprocity, we also have the following result

$$x_1(t) x_2(t) \stackrel{\text{CTFT}}{\leftrightarrow} X_1(f) * X_2(f)$$

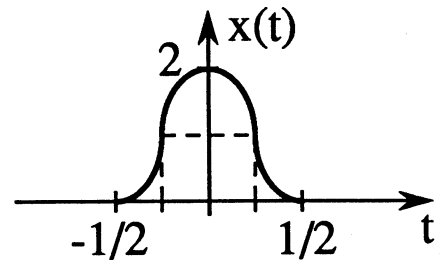
This can be very useful for calculating transforms of certain functions.

Example

$$x(t) = \begin{cases} \frac{1}{2} [1 + \cos(2\pi t)] & , \quad |t| \leq 1/2 \\ 0 & , \quad |t| > 1/2 \end{cases}$$

Find $X(f)$

$$x(t) = \frac{1}{2} [1 + \cos(2\pi t)] \text{rect}(t)$$



$$\therefore X(f) = \frac{1}{2} \left\{ \delta(f) + \frac{1}{2} [\delta(f - 1) + \delta(f + 1)] \right\} * \text{sinc}(f)$$

Since convolution obeys linearity, we can write this as

$$X(f) = \frac{1}{2} \left\{ \delta(f) * \text{sinc}(f) + \frac{1}{2} [\delta(f - 1) * \text{sinc}(f) + \delta(f + 1) * \text{sinc}(f)] \right\}$$

All three convolutions here are of the same general form.

Identity

For any signal $w(t)$,

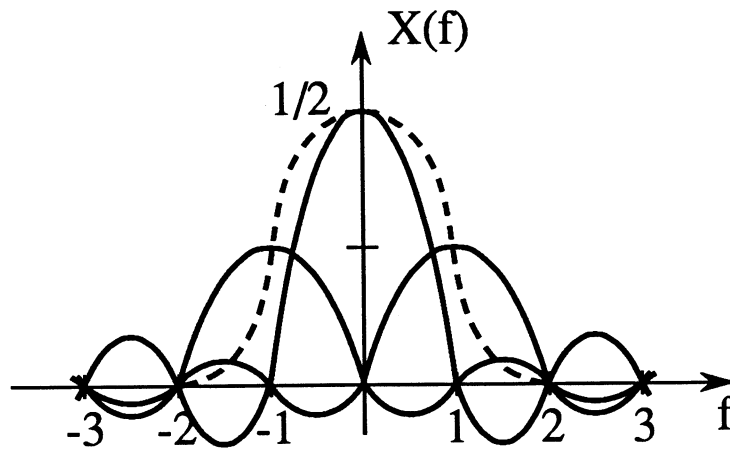
$$w(t) * \delta(t - t_0) = w(t - t_0)$$

Proof:

$$\begin{aligned} w(t) * \delta(t - t_0) &= \int w(\tau) \delta(t - \tau - t_0) d\tau \\ &= w(t - t_0) \quad (\text{by sifting property}) \end{aligned}$$

Using the identity,

$$\begin{aligned} X(f) &= \frac{1}{2} \left\{ \delta(f) * \text{sinc}(f) + \frac{1}{2} [\delta(f - 1) * \text{sinc}(f) \right. \\ &\quad \left. + \delta(f + 1) * \text{sinc}(f)] \right\} \\ &= \frac{1}{2} \left\{ \text{sinc}(f) + \frac{1}{2} [\text{sinc}(f - 1) + \text{sinc}(f + 1)] \right\} \end{aligned}$$



Fourier Transform of Periodic Signals

- We previously developed the Fourier series as a spectral representation for periodic CT signals.
- Such signals are neither square integrable nor absolutely integrable, and hence do not satisfy the conditions for existence of the CTFT.
- However, by applying the concept of the generalized Fourier transform, we can obtain a Fourier transform for periodic signals.
- This allows us to treat the spectral analysis of *all* CT signals within a single framework.

- We can also obtain the same result directly from the Fourier series.

Let $x_0(t)$ denote one period of a signal that is periodic with period T , *i.e.* $x_0(t) = 0$, $|t| > T/2$.

Define $x(t) = \text{rep}_T[x_0(t)]$

The Fourier series representation for $x(t)$ is

$$x(t) = \frac{1}{T} \sum_k X_k e^{j2\pi kt/T}$$

Taking the CTFT directly, we obtain

$$\begin{aligned} X(f) &= \mathcal{F}\left\{\frac{1}{T} \sum_k X_k e^{j2\pi kt/T}\right\} \\ &= \frac{1}{T} \sum_k X_k \mathcal{F}\left\{e^{j2\pi kt/T}\right\} \quad (\text{by linearity}) \\ &= \frac{1}{T} \sum_k X_k \delta(f - k/T) \end{aligned}$$

Also

$$\begin{aligned} X_k &= \int_{-T/2}^{T/2} x(t) e^{-j2\pi kt/T} dt \\ &= \int_{-\infty}^{\infty} x_0(t) e^{-j2\pi kt/T} dt \\ &= X_0(k/T) \end{aligned}$$

Thus

$$\begin{aligned} X(f) &= \frac{1}{T} \sum_k X_0(k/T) \delta(f - k/T) \\ &= \frac{1}{T} \text{comb} \frac{1}{T} [X_0(f)] \end{aligned}$$

Dropping the subscript 0, we may state this result in the form of a transform relation:

$$\text{rep}_T[x(t)] \stackrel{\text{CTFT}}{\leftrightarrow} \frac{1}{T} \text{comb} \frac{1}{T} [X(f)]$$

For our derivation, we required that $x(t) = 0$, $|t| > T/2$. However, when the generalized transform is used to derive the result, this restriction is not needed.

Example

