

# CAPACITANCE AND INDUCTANCE

Introduces two passive, energy storing devices: Capacitors and Inductors

## LEARNING GOALS

### CAPACITORS

Store energy in their electric field (electrostatic energy)  
Model as circuit element

### INDUCTORS

Store energy in their magnetic field  
Model as circuit element

### CAPACITOR AND INDUCTOR COMBINATIONS

Series/parallel combinations of elements

### RC OP-AMP CIRCUITS

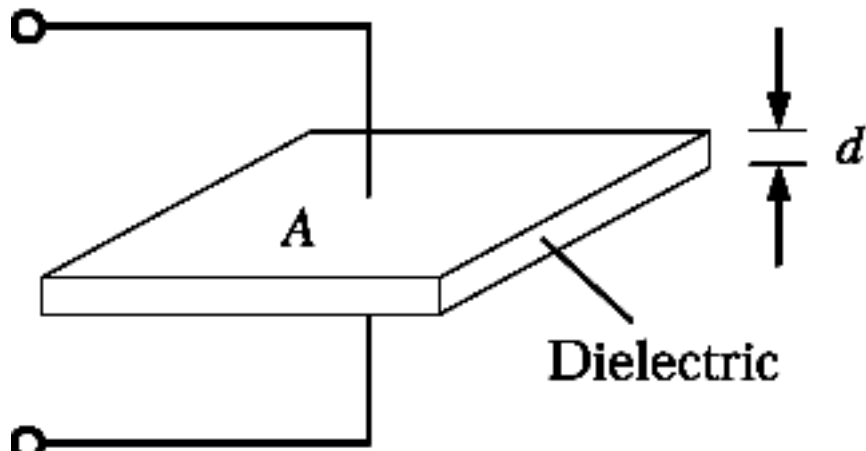
Integration and differentiation circuits



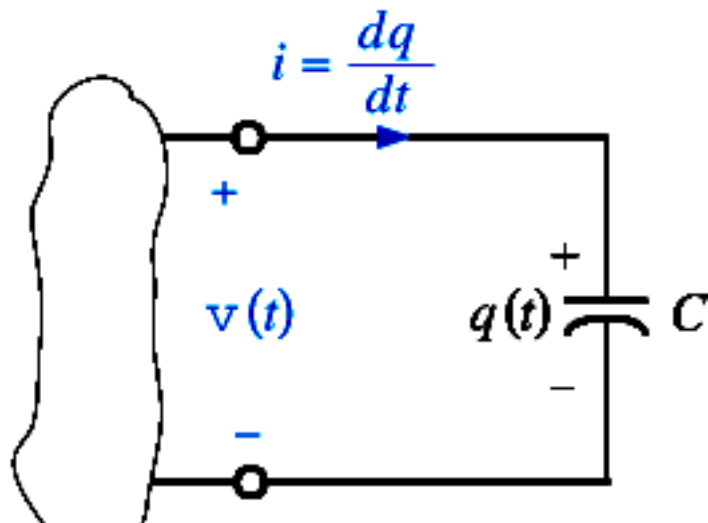
# CAPACITORS

First of the energy storage devices to be discussed

Typical Capacitors

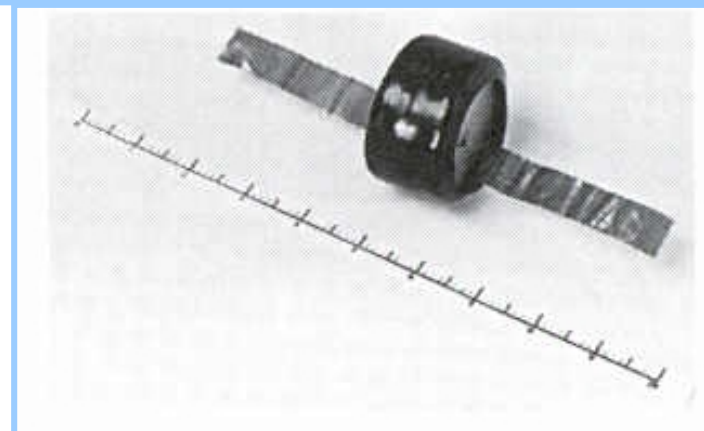


Basic parallel-plates capacitor



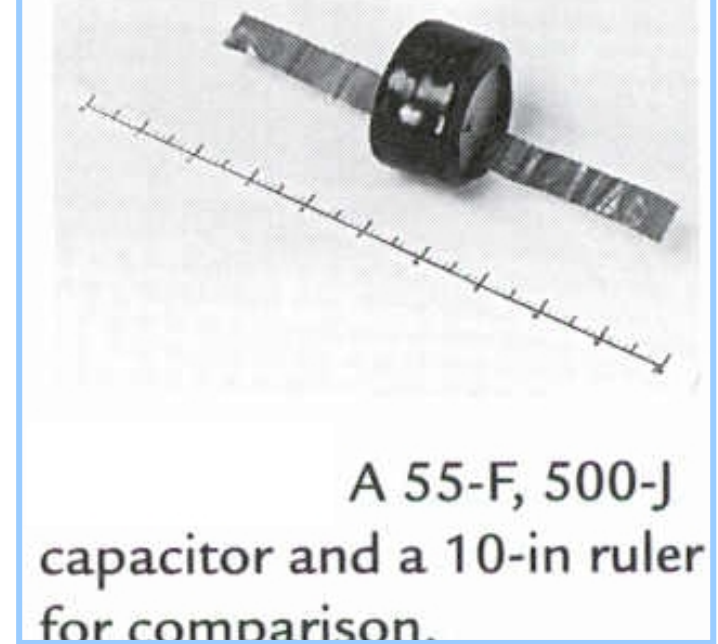
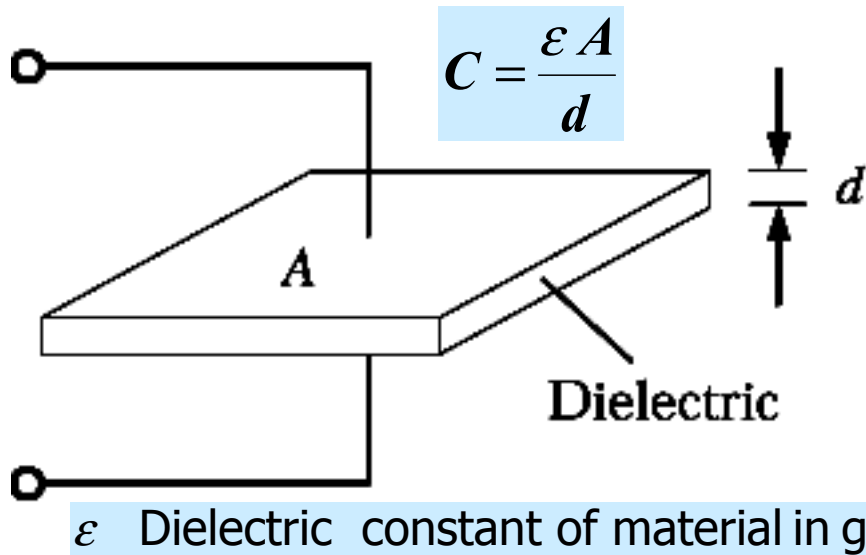
CIRCUIT REPRESENTATION

NOTICE USE OF PASSIVE SIGN CONVENTION



A 55-F, 500-J capacitor and a 10-in ruler for comparison.





### PLATE SIZE FOR EQUIVALENT AIR-GAP CAPACITOR

$$55F = \frac{8.85 \times 10^{-12} A}{1.016 \times 10^{-4}} \Rightarrow A = 6.3141 \times 10^8 m^2$$

Normal values of capacitance are small.  
 Microfarads is common.  
 For integrated circuits nano or pico farads are not unusual

Basic capacitance law  $Q = f(V_C)$

Linear capacitors obey Coulomb's law  $Q = CV_C$

C is called the CAPACITANCE of the device and has units of  $\frac{\text{charge}}{\text{voltage}}$

One Farad(F) is the capacitance of a device that can store one Coulomb of charge at one Volt.

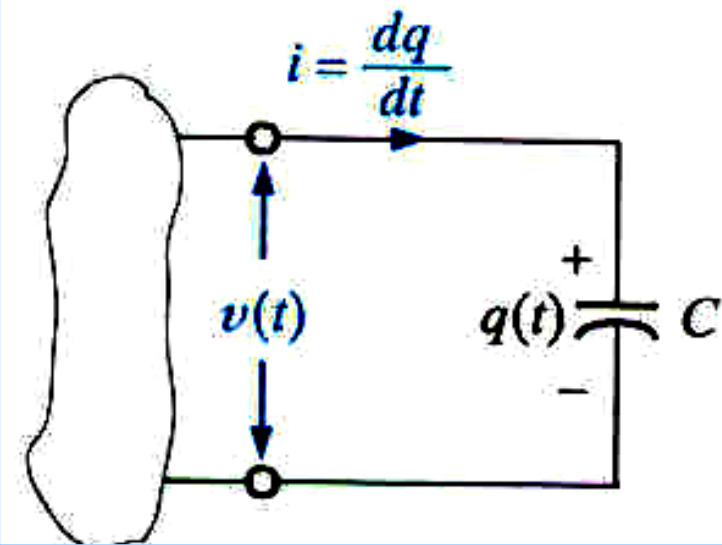
$$\text{Farad} = \frac{\text{Coulomb}}{\text{Volt}}$$

**EXAMPLE** Voltage across a capacitor of 2 micro Farads holding 10mC of charge

$$V_C = \frac{1}{C} Q = \frac{1}{2 * 10^{-6}} 10 * 10^{-3} = 5000 \text{ v}$$

Capacitance in Farads, charge in Coulombs result in voltage in Volts

**Capacitors can be dangerous!!!**

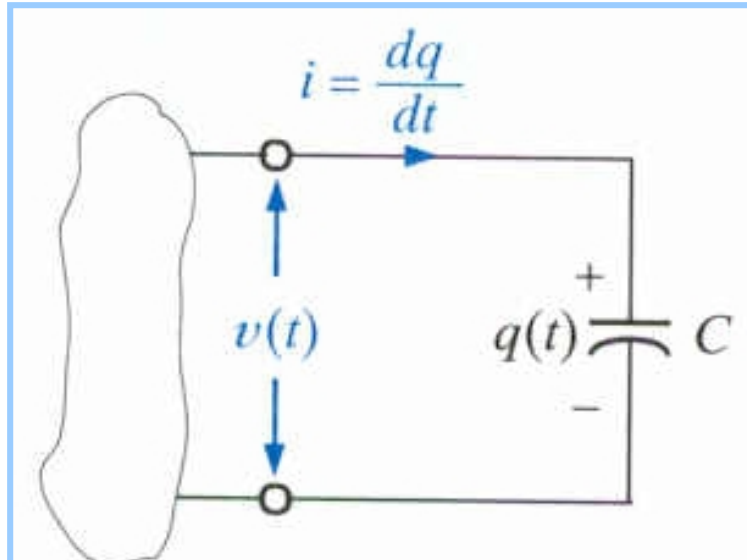


Linear capacitor circuit representation



Capacitors only store and release ELECTROSTATIC energy. They do not "create"

The capacitor is a passive element and follows the passive sign convention

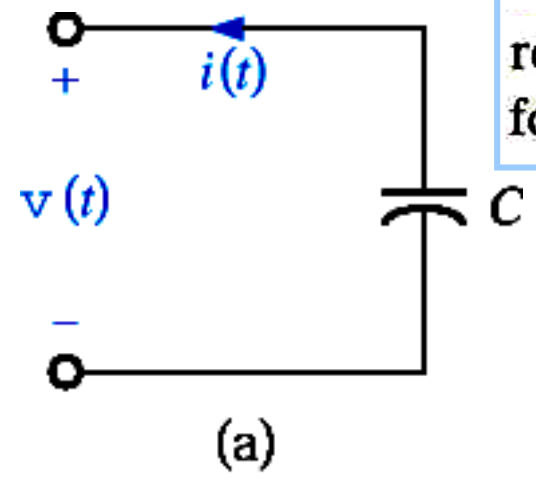


Linear capacitor circuit representation

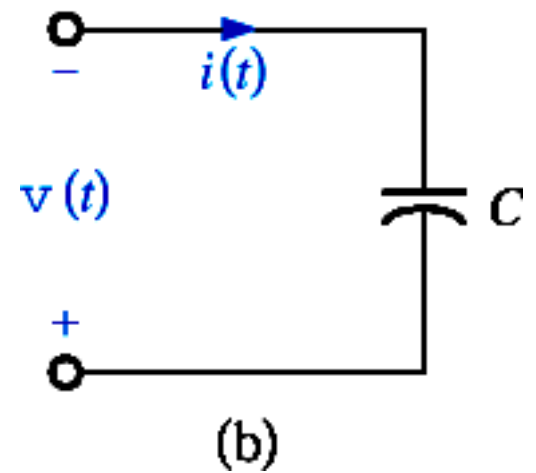
$$i(t) = C \frac{dv}{dt}(t)$$

### LEARNING BY DOING

Write the  $i-v$  relationship for the following capacitors.



$$i(t) = -C \frac{dv(t)}{dt}$$



$$i(t) = -C \frac{dv(t)}{dt}$$



$$Q_C = CV_C \quad \text{Capacitance Law}$$

If the voltage varies the charge varies and there is a displacement current

One can also express the voltage across in terms of the current

... Or one can express the current through in terms of the voltage across

$$V_C(t) = \frac{1}{C} Q = \frac{1}{C} \int_{-\infty}^t i_C(x) dx$$

$$i_C = \frac{dQ}{dt} = C \frac{dV_C}{dt}$$

Integral form of Capacitance law

Differential form of Capacitance law

The mathematical implication of the integral form is ...

Implications of differential form??

$$V_C(t-) = V_C(t+); \quad \forall t$$

$$V_C = \text{Const} \Rightarrow i_C = 0$$

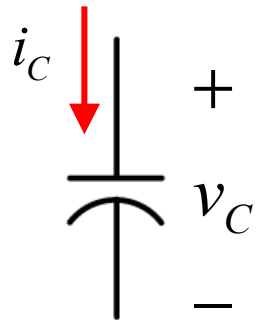
DC or steady state behavior

Voltage across a capacitor **MUST** be continuous

A capacitor in steady state acts as an **OPEN CIRCUIT**



# CAPACITOR AS CIRCUIT ELEMENT



$$i_C(t) = C \frac{dv_C(t)}{dt}$$

$$i_R = \frac{1}{R} v_R$$

$$v_C(t) = \frac{1}{C} \int_{-\infty}^t i_C(x) dx$$

$$v_R = R i_R$$

**Ohm's Law**

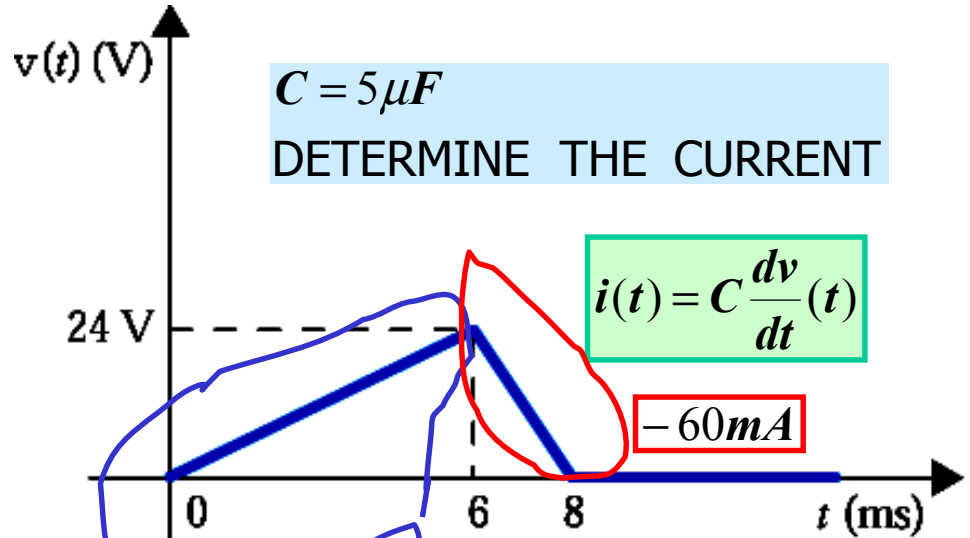
$$\int_{-\infty}^t = \int_{-\infty}^{t_0} + \int_{t_0}^t$$

$$v_C(t) = \frac{1}{C} \int_{-\infty}^{t_0} i_C(x) dx + \frac{1}{C} \int_{t_0}^t i_C(x) dx$$

$$v_C(t) = v_C(t_0) + \frac{1}{C} \int_{t_0}^t i_C(x) dx$$

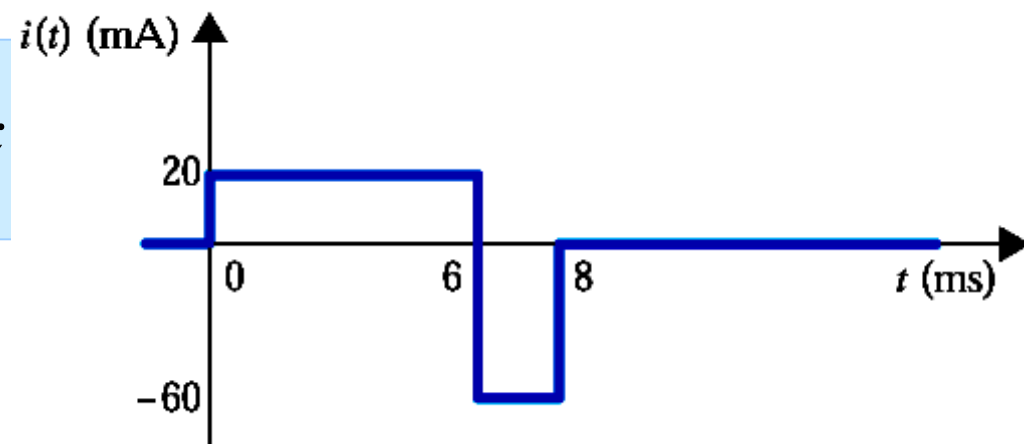
The fact that the voltage is defined through an integral has important implications...

# LEARNING EXAMPLE

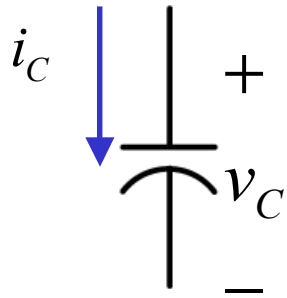


$$i = 5 \times 10^{-6} [F] \times \frac{24}{6 \times 10^{-3}} \left[ \frac{V}{s} \right] = 20mA$$

$i(t) = 0$  elsewhere



# CAPACITOR AS ENERGY STORAGE DEVICE



**Instantaneous power**

$$p_C(t) = v_C(t)i_C(t) \quad \mathbf{W}$$

$$i_C(t) = C \frac{dv_C}{dt}(t)$$

$$v_C(t) = \frac{1}{C} \int_{-\infty}^t i_C(x) dx = \frac{q_C(t)}{C}$$

$$p_C(t) = Cv_C(t) \frac{dv_C}{dt}$$

$$p_C(t) = \frac{1}{C} q_C(t) \frac{dq_C}{dt}(t)$$

**Energy is the integral of power**

$$p_C(t) = C \frac{d}{dt} \left( \frac{1}{2} v_C^2(t) \right)$$

$$w_C(t_2, t_1) = \int_{t_1}^{t_2} p_C(x) dx$$

$$p_C(t) = \frac{1}{C} \frac{d}{dt} \left( \frac{1}{2} q_C^2(t) \right)$$

If  $t_1$  is minus infinity we talk about “energy stored at time  $t_2$ .”

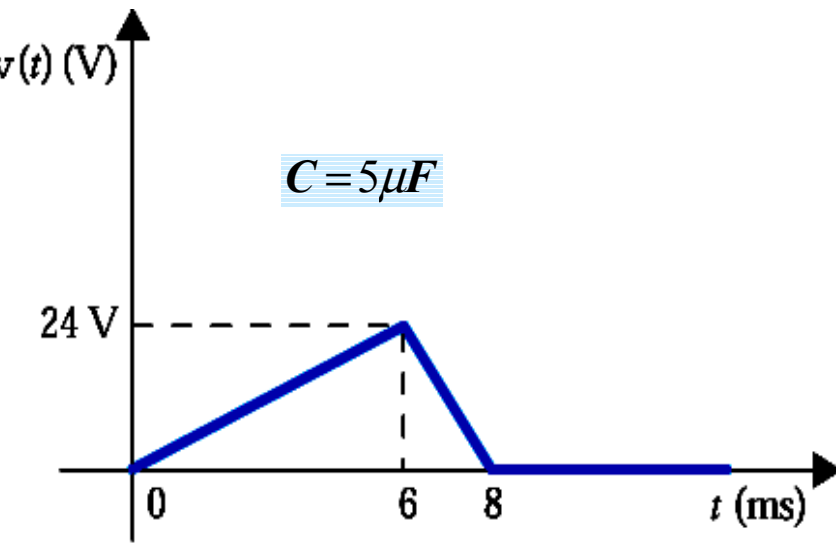
If both limits are infinity then we talk about the “total energy stored.”

$$w_C(t_2, t_1) = \frac{1}{2} C v_C^2(t_2) - \frac{1}{2} C v_C^2(t_1)$$

$$w_C(t_2, t_1) = \frac{1}{C} q_C^2(t_2) - \frac{1}{C} q_C^2(t_1)$$







## LEARNING EXAMPLE

Energy stored in 0 - 6 msec

$$w_C(0,6) = \frac{1}{2} C v_C^2(6) - \frac{1}{2} C v_C^2(0)$$

$$w_C(0,6) = \frac{1}{2} 5 * 10^{-6} [F] * (24)^2 [V^2]$$

Charge stored at 3msec

$$q_C(3) = C v_C(3)$$

$$q_C(3) = 5 * 10^{-6} [F] * 12 [V] = 60 \mu C$$

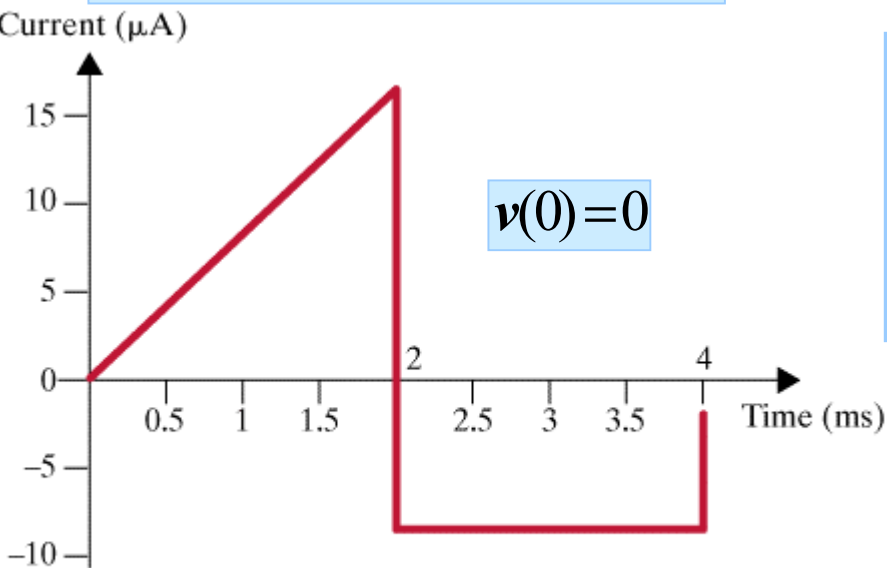
“total energy stored?” ....

“total charge stored?” ...

If charge is in Coulombs  
and capacitance in Farads  
then the energy is in ....



$C = 4\mu F$ . FIND THE VOLTAGE



$$i(t) = \begin{cases} \frac{16 \times 10^{-6} t}{2 \times 10^{-3}} & 0 \leq t \leq 2 \text{ ms} \\ -8 \times 10^{-6} & 2 \text{ ms} \leq t \leq 4 \text{ ms} \\ 0 & 4 \text{ ms} < t \end{cases}$$

$$v(t) = v(0) + \frac{1}{C} \int_0^t i(x) dx; \quad t > 0$$

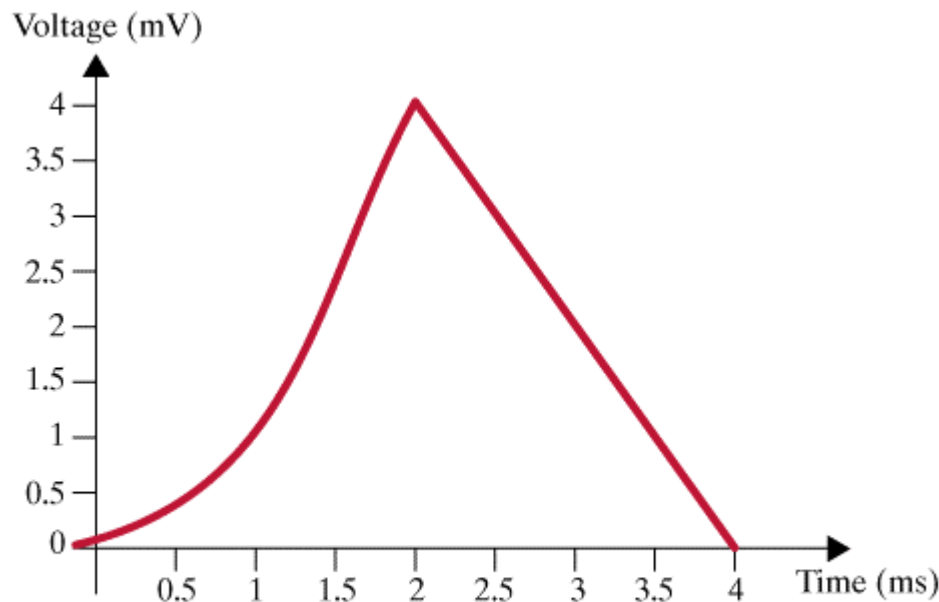
$$v(t) = \frac{1}{(4)(10^{-6})} \int_0^t 8(10^{-3})x dx = 10^3 t^2 \quad 0 \leq t \leq 2$$

$$v(t) = v(2) + \frac{1}{C} \int_2^t i(x) dx; \quad t > 2$$

$$v(2 \text{ ms}) = 10^3 (2 \times 10^{-3})^2 = 4 \text{ mV}$$

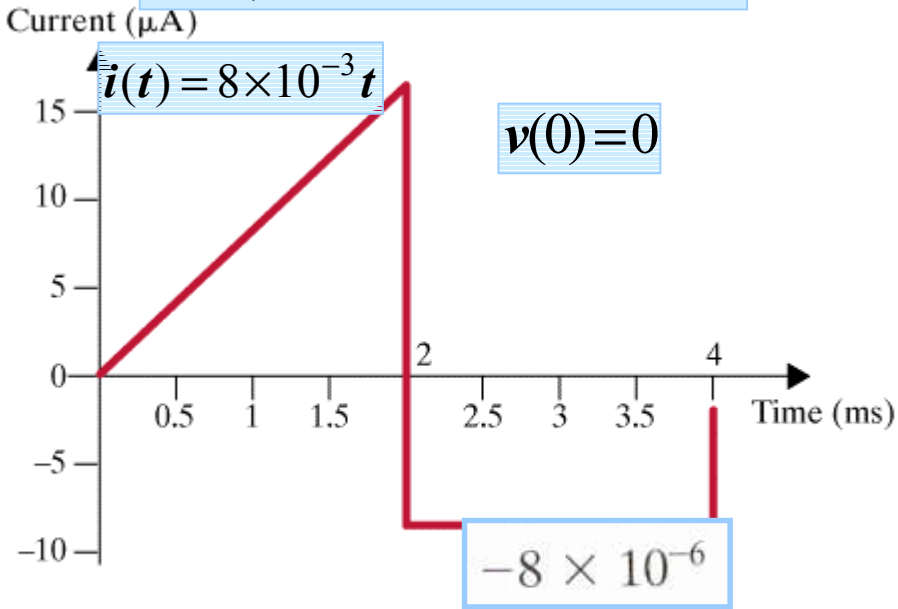
$$v(t) = \frac{1}{(4)(10^{-6})} \int_{2(10^{-3})}^t - (8)(10^{-6}) dx + (4)(10^{-3}) \quad 2 < t \leq 4 \text{ ms}$$

$$v(t) = -2t + 8 \times 10^{-3} [V]$$



(b)

$C = 4\mu F$ . FIND THE POWER



(a)

$$p(t) = v(t)i(t)$$

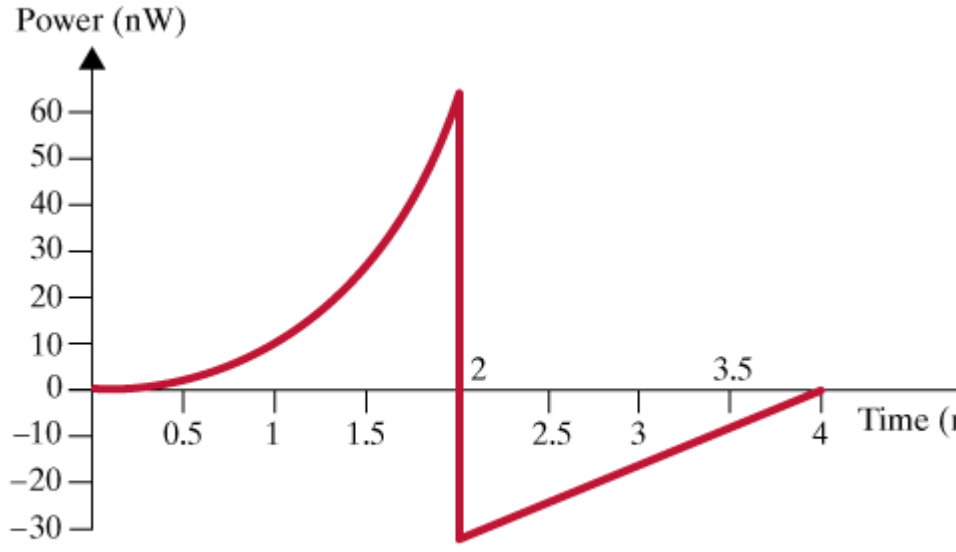
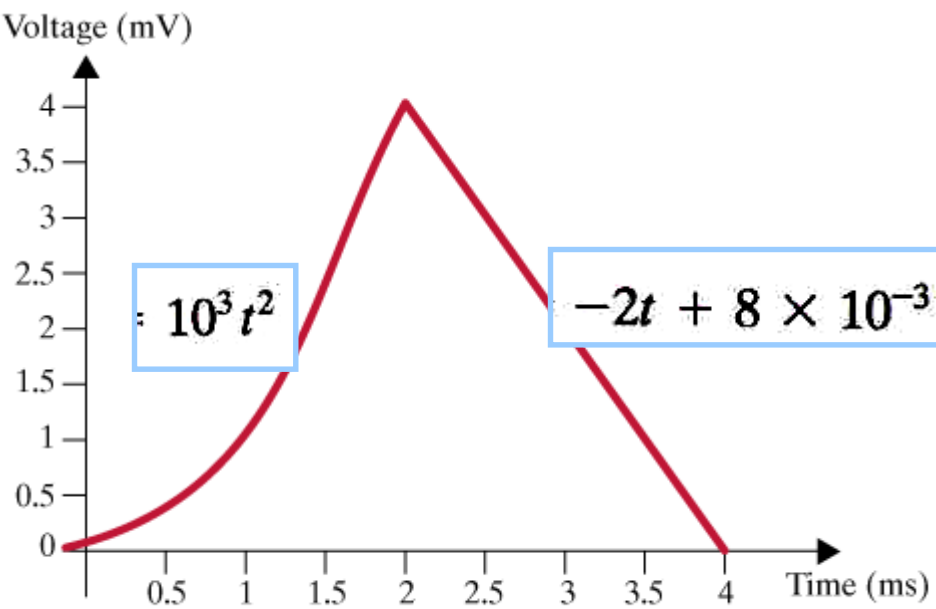
$$p(t) = 8t^3, \quad 0 \leq t \leq 2ms$$

$$p(t) = -(8)(10^{-6})(-2t + 8 \times 10^{-3})$$

$$= 16(10^{-6})t - 64(10^{-9})$$

$$2 < t \leq 4ms$$

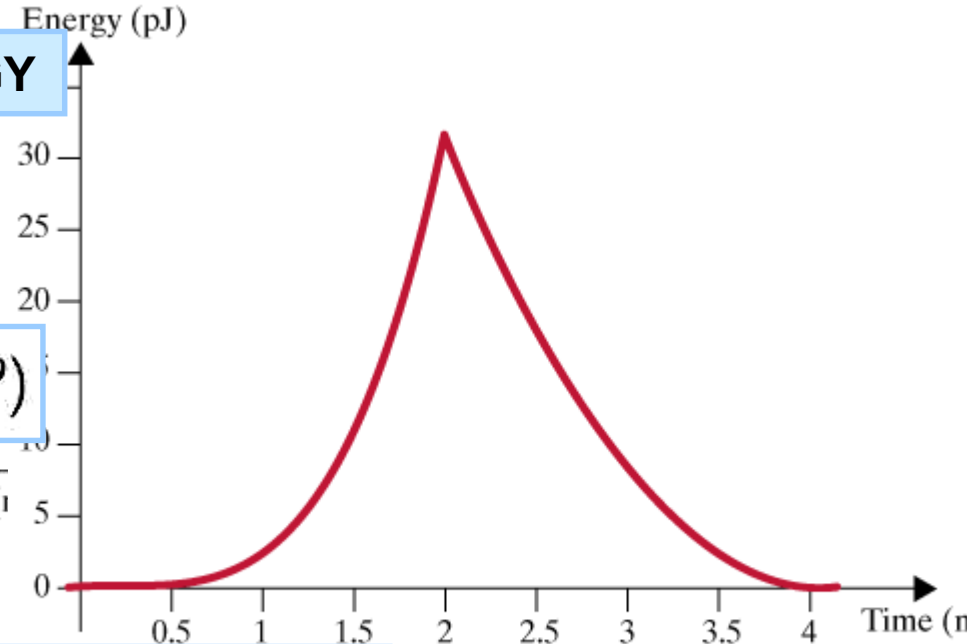
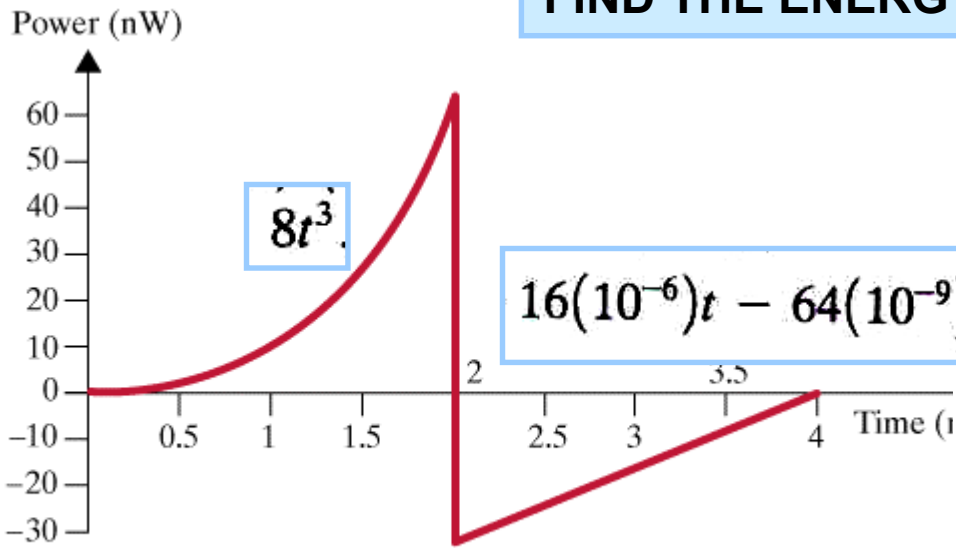
$$p(t) = 0, \text{ elsewhere}$$



(c)



# FIND THE ENERGY



$$p(t) = 8t^3, \quad 0 \leq t \leq 2 \text{ ms}$$

$$w(t) = \int_0^t 8x^3 dx = 2t^4$$

$$w(2 \text{ ms}) = 32 \text{ pJ}$$

$$p(t) = -(8)(10^{-6})(-2t + 8 \times 10^{-3})$$

$$= 16(10^{-6})t - 64(10^{-9})$$

$$p(t) = 0, \text{ elsewhere}$$

$$2 < t \leq 4 \text{ ms}$$

$$w(t) = \int_{t_0}^t p(x) dx + w(t_0)$$

$$w(t) = \int_{2 \times 10^{-3}}^t [(16 \times 10^{-6})x - (64 \times 10^{-9})] dx + 32 \times 10^{-12}$$

$$= (8 \times 10^{-6})t^2 - (64 \times 10^{-9})t + 128 \times 10^{-12}$$

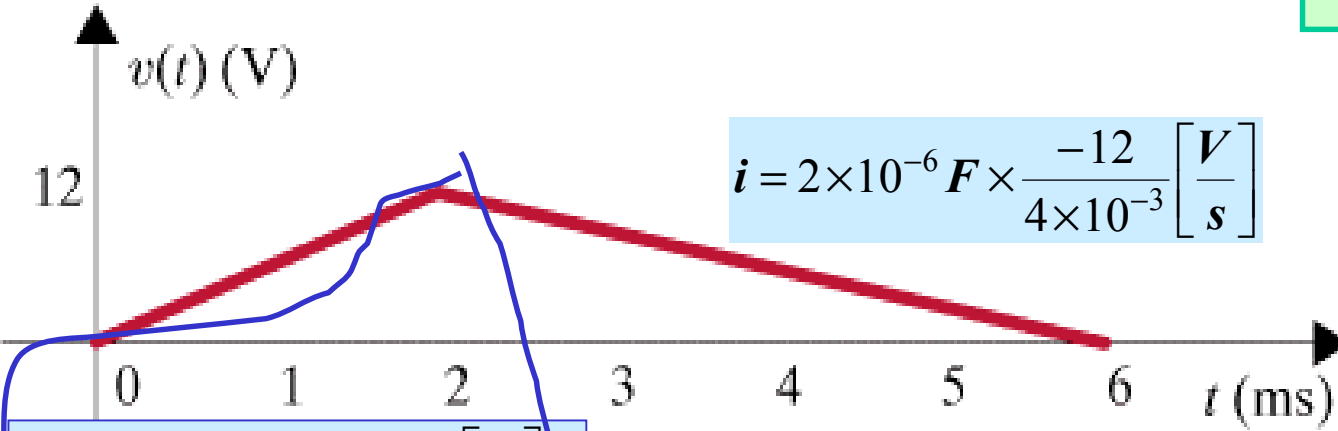


LEARNING EXTENSION

$$C = 2\mu F$$

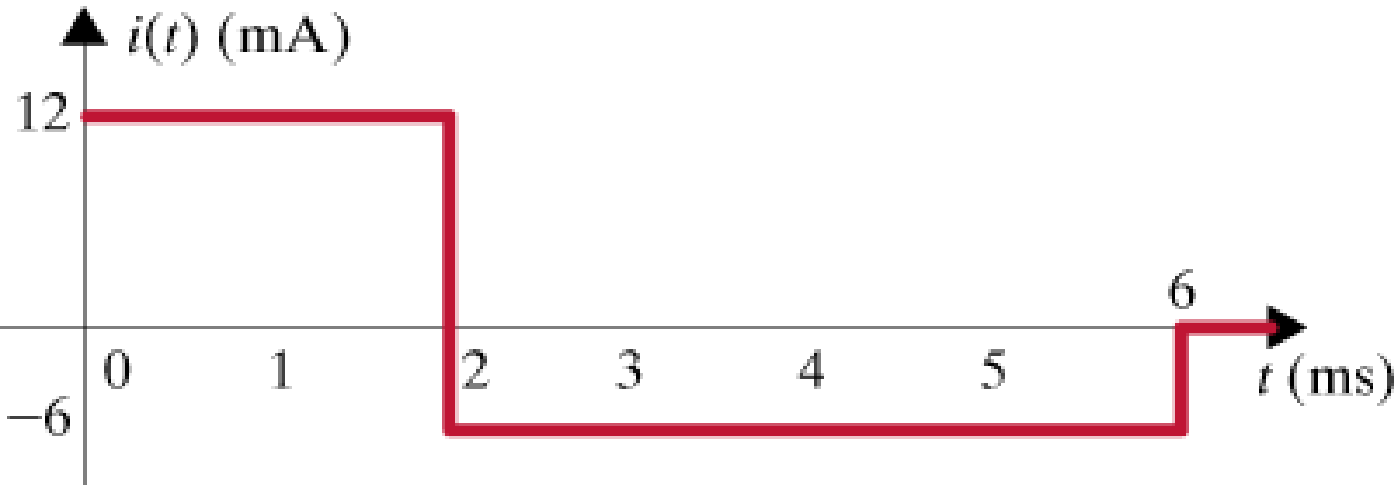
DETERMINE THE CURRENT

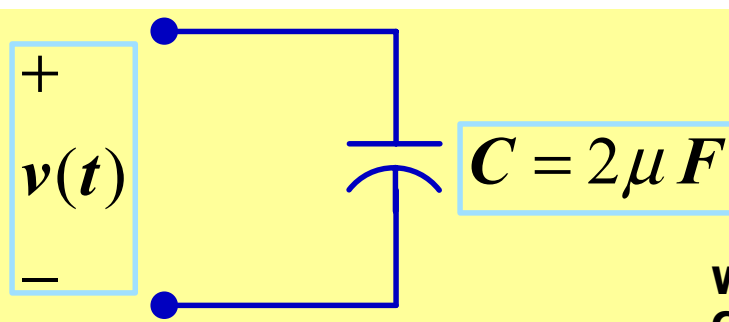
$$i(t) = C \frac{dv}{dt}(t)$$



$$i = 2 \times 10^{-6} F \times \frac{-12}{4 \times 10^{-3}} \left[ \frac{V}{s} \right]$$

$$i = 2 \times 10^{-6} F \times \frac{12}{2 \times 10^{-3}} \left[ \frac{V}{s} \right] =$$



**SAMPLE PROBLEM**

$$v(t) = 130 \sin(120\pi t)$$

**WHAT VARIABLES CAN BE COMPUTED?**
**Energy stored at a given time t**

$$E(t) = \frac{1}{2} C v_C^2(t)$$

$$E(1/240) = \frac{1}{2} 2 * 10^{-6} [F] * 130^2 \sin^2\left(\frac{\pi}{2}\right) \quad \mathbf{J}$$

**Charge stored at a given time**

$$q_C(t) = C v_C(t)$$

$$q_C(1/120) = 2 * 10^{-6} [C] * \sin(\pi) [V] = 0 \quad \mathbf{C}$$

**Current through the capacitor**

$$i_C = C \frac{dv_C}{dt}(t)$$

$$i_C(1/120) = 2 * 10^{-6} * 130 * 120\pi \cos(\pi) \quad \mathbf{A}$$

**Electric power supplied to capacitor at a given time**

$$p_C(t) = v_C(t) i_C(t) \quad \mathbf{W}$$

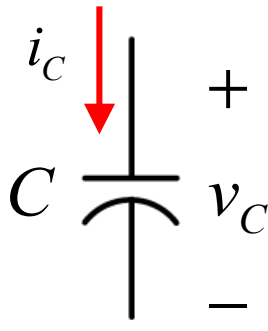
**Energy stored over a given time interval**

$$w(t_2, t_1) = \frac{1}{2} C v_C^2(t_2) - \frac{1}{2} C v_C^2(t_1) \quad \mathbf{J}$$



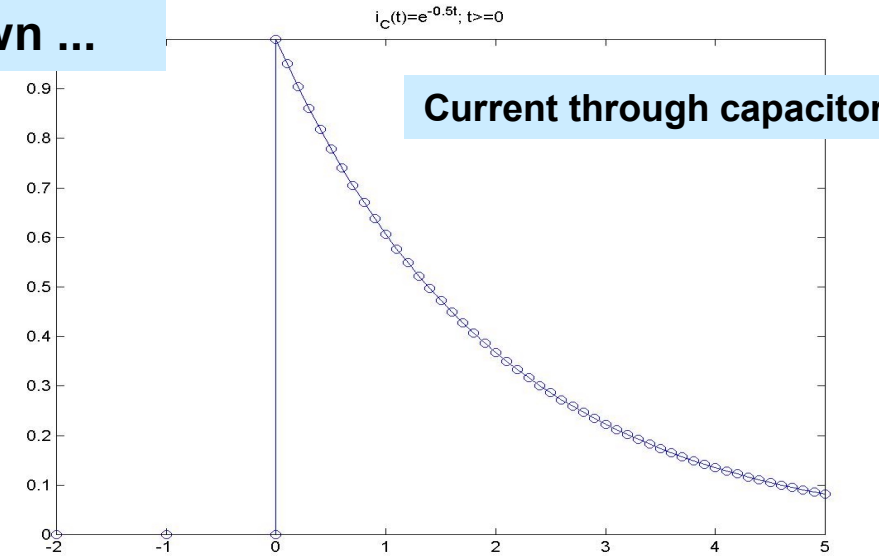
# SAMPLE PROBLEM

If the current is known ...



$$i_c(t) = \begin{cases} e^{-0.5t}; & t \geq 0 \\ 0; & t < 0 \end{cases} [mA]$$

$$C = 2\mu F$$



Voltage at a given time  $t$

$$v_C(t) = \frac{1}{C} \int_{-\infty}^t i_C(x) dx$$

$$v_C(0) = 0[V]$$

$$v_C(t) = v_C(t_0) + \frac{1}{C} \int_{t_0}^t i_C(x) dx$$

Voltage at a given time  $t$  when voltage at time  $t_0 < t$  is also known

$$v_C(2) = v_C(0) + \frac{1}{C} \int_0^2 e^{-0.5x} dx = \frac{1}{2 \cdot 10^{-6}} \left[ -\frac{1}{0.5} e^{-0.5x} \right]_0^2 = \frac{1}{2 \cdot 10^{-6}} \frac{1}{0.5} (1 - e^{-1}) = 0.6321 \cdot 10^6 \text{ V}$$

Charge at a given time

$$q_C(t) = C v_C(t)$$

$$q_C(2) = 2 \cdot 0.6321 \text{ C}$$

Voltage as a function of time

$$v_C(t) = \frac{1}{C} \int_{-\infty}^t i_C(x) dx$$

$$v_C(t) = 0; t \leq 0$$

$$v_C(t) = v_C(0) + \frac{1}{C} \int_0^t e^{-0.5x} dx$$

Electric power supplied to capacitor

$$p_C(t) = v_C(t) i_C(t) \text{ W}$$

$$v_C(t) = \begin{cases} 10^6 (1 - e^{-0.5t}); & t \geq 0 \\ 0; & t < 0 \end{cases} \text{ V}$$

Energy stored in capacitor at a given time

$$w(t) = \frac{1}{2} C v_C^2(t) \text{ J}$$

"Total" energy stored in the capacitor

$$w_T = \frac{1}{2} C v_C^2(\infty)$$

$$w_T = \frac{1}{2} 2 \cdot 10^{-6} \cdot (10^6)^2 = 10^6 \text{ J}$$



# SAMPLE PROBLEM

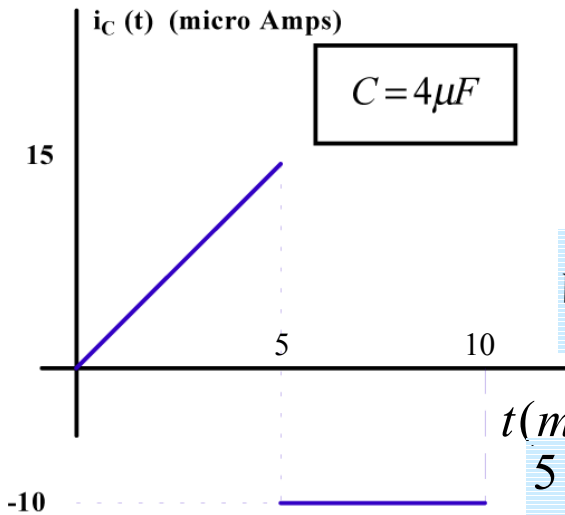
## Given current and capacitance

$$V_c(t) = \frac{1}{C} \int_{t_0}^t i_c(t) dt + V_c(t_0)$$

### Compute voltage as a function of time

At minus infinity everything is zero. Since current is zero for  $t < 0$  we have

$$V_c(t) = 0; t \leq 0$$



$$0 < t < 5 \text{ m sec} \Rightarrow$$

$$i_c(t) = \frac{15 \mu A}{5 \text{ ms}} t = 3 \frac{10^{-6} A}{10^{-3} s} t = 3 * 10^{-3} [A/s] t$$

$$V_c(0) = 0 \Rightarrow V_c(t) = \frac{3 * 10^{-3}}{4 * 10^{-6}} \int_0^t x dx [V] = \frac{3 * 10^3}{8} t^2 [V]; 0 < t < 5 * 10^{-3} [s]$$

### In particular

$$V_c(5 \text{ ms}) = \frac{3 * 10^3 * (5 * 10^{-3})^2}{8} [V] = \frac{75}{8} [mV]$$

$$5 < t < 10 \text{ ms} \Rightarrow i_c(t) = -10 [\mu A]$$

### Charge stored at 5ms

$$q_c(t) = CV_c(t)$$

$$q(5 \text{ ms}) = 4 * 10^{-6} [F] * \frac{75 * 10^{-3}}{8} [V]$$

$$q(5 \text{ ms}) = (75/2) [nC]$$

$$V_c(5 \text{ ms}) = \frac{75}{8} [mV] \Rightarrow V_c(t) = \frac{75 * 10^{-3}}{8} + \frac{1}{4 * 10^{-6}} \int_{5 * 10^{-3}}^t (-10 * 10^{-6}) [A/s] dx$$

$$V_c(t) = \frac{75 * 10^{-3}}{8} - \frac{10}{4} (t - 5 * 10^{-3}) [V]; 5 * 10^{-3} < t < 10 * 10^{-3} [s]$$

### Total energy stored

$$E = \frac{1}{2} CV_c^2$$

Total means at infinity. Hence

$$E_T = 0.5 * 4 * 10^{-6} \left( \frac{25 * 10^{-3}}{8} \right)^2 [J]$$

Before looking into a formal way to describe the current we will look at additional questions that can be answered.

Now, for a formal way to represent piecewise functions....

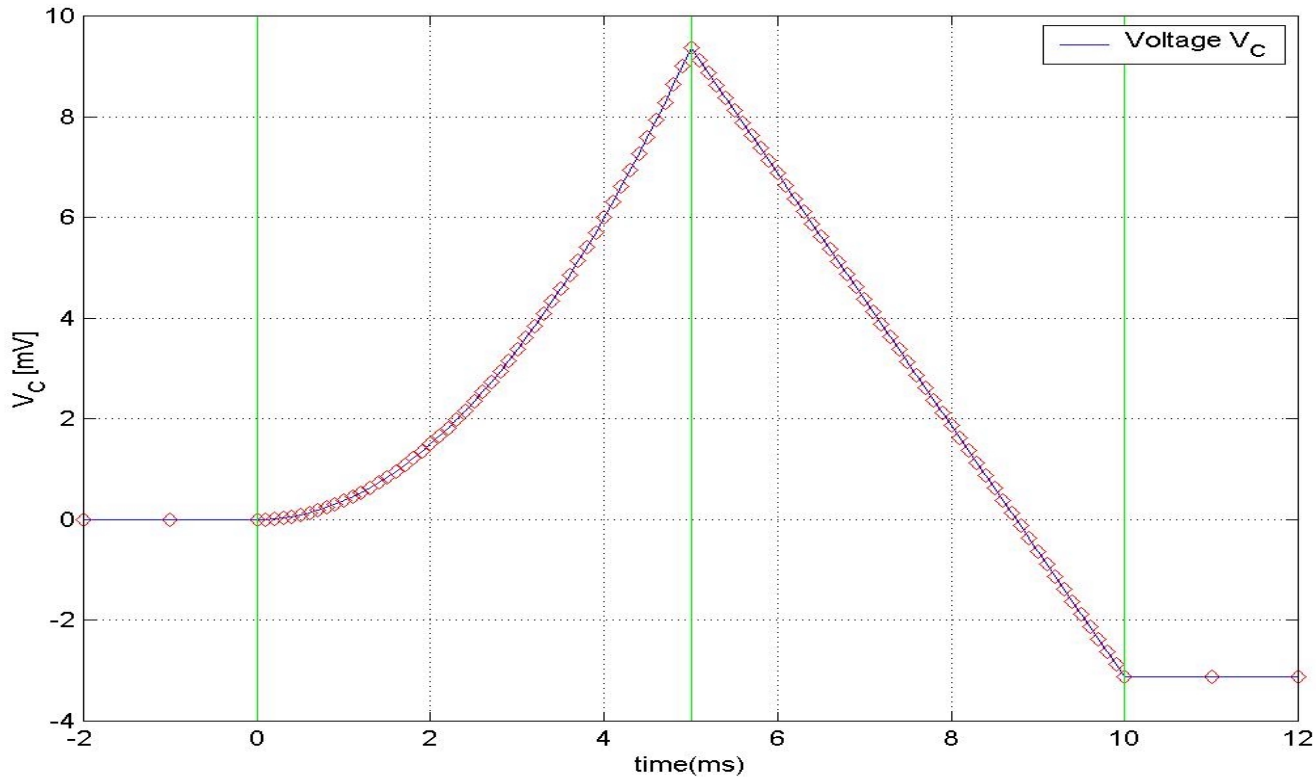




# Formal description of a piecewise analytical signal

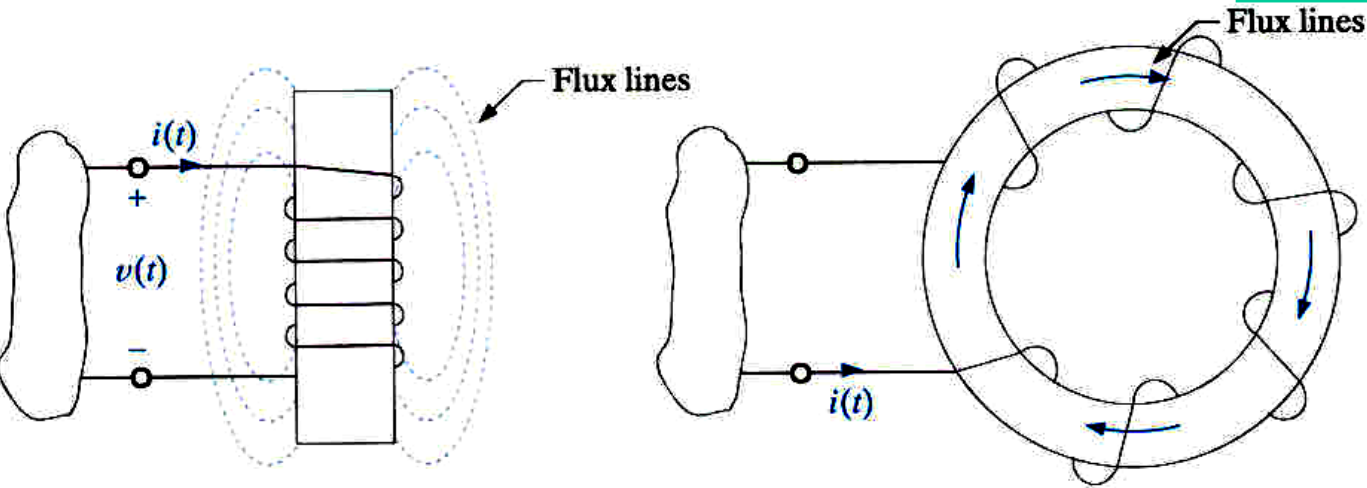
$$V_c(t) = \begin{cases} 0; & t \leq 0 \\ \frac{3}{8}t^2; & 0 < t < 5\text{ms} \\ \frac{75}{8} - \frac{10}{4}(t-5); & 5 < t \leq 10 \text{ [ms]} \\ -\frac{25}{8}; & t > 10 \text{ [ms]} \end{cases}$$

[mV]



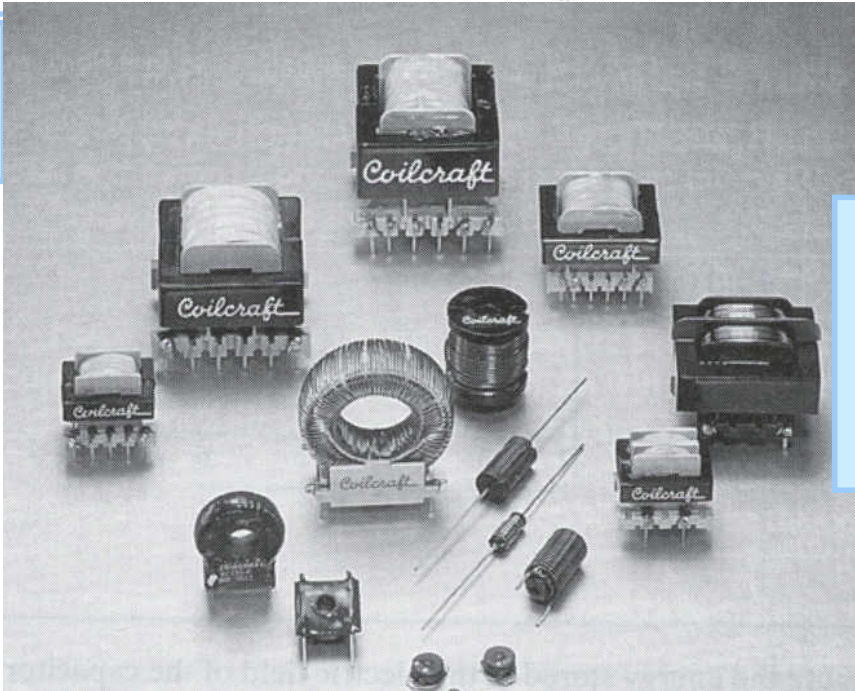
# INDUCTORS

NOTICE USE OF PASSIVE SIGN CONVENTION



Circuit representation for an inductor

Flux lines may extend beyond inductor creating stray inductance effects



A TIME VARYING FLUX CREATES A COUNTER EMF AND CAUSES A VOLTAGE TO APPEAR AT THE TERMINALS OF THE DEVICE



A TIME VARYING MAGNETIC FLUX INDUCES A VOLTAGE

$$v_L = \frac{d\phi}{dt}$$

Induction law

FOR A LINEAR INDUCTOR THE FLUX IS PROPORTIONAL TO THE CURRENT

$$\phi = Li_L \Rightarrow$$

$$v_L = L \frac{di_L}{dt}$$

DIFFERENTIAL FORM OF INDUCTION LAW

THE PROPORTIONALITY CONSTANT, L, IS CALLED THE INDUCTANCE OF THE COMPONENT

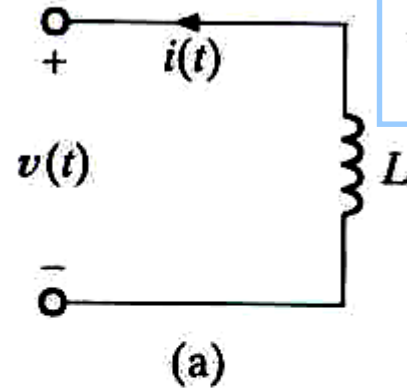
INDUCTANCE IS MEASURED IN UNITS OF henry (H). DIMENSIONALLY

$$\text{HENRY} = \frac{\text{Volt}}{\text{Amp}/\text{sec}}$$

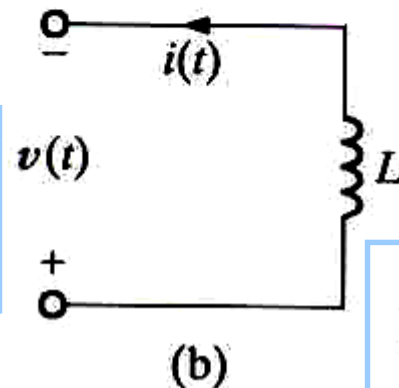
INDUCTORS STORE ELECTROMAGNETIC ENERGY. THEY MAY SUPPLY STORED ENERGY BACK TO THE CIRCUIT BUT THEY CANNOT CREATE ENERGY. THEY MUST ABIDE BY THE PASSIVE SIGN CONVENTION

LEARNING by Doing

Write the  $i$ - $v$  relationship for the following inductors.



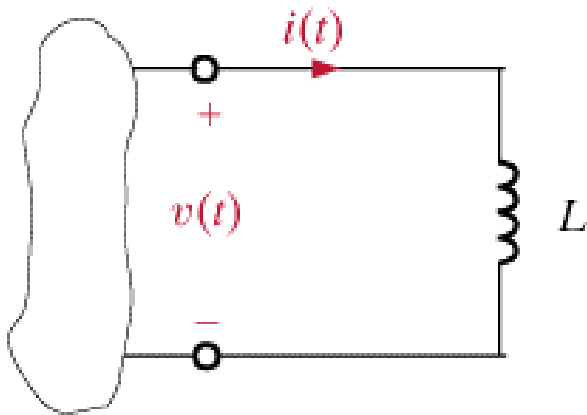
$$v(t) = -L \frac{di(t)}{dt}$$



$$v(t) = L \frac{di(t)}{dt}$$

Follow passive sign convention





$$v_L = L \frac{di_L}{dt}$$

Differential form of induction law

$$i_L(t) = \frac{1}{L} \int_{-\infty}^t v_L(x) dx$$

Integral form of induction law

$$i_L(t) = i_L(t_0) + \frac{1}{L} \int_{t_0}^t v_L(x) dx; \quad t \geq t_0$$

A direct consequence of integral form

$$i_L(t^-) = i_L(t^+); \quad \forall t$$

Current MUST be continuous

A direct consequence of differential form

$$i_L = \text{Const.} \Rightarrow v_L = 0$$

DC (steady state) behavior

### Power and Energy stored

$$p_L(t) = v_L(t) i_L(t) \quad \mathbf{w}$$

$$p_L(t) = L \frac{di_L}{dt}(t) i_L(t) = \frac{d}{dt} \left( \frac{1}{2} L i_L^2(t) \right)$$

$$w_L(t_2, t_1) = \int_{t_1}^{t_2} \frac{d}{dt} \left( \frac{1}{2} L i_L^2(x) \right) dx \quad \mathbf{J}$$

Current in Amps, Inductance in Henrys  
yield energy in Joules

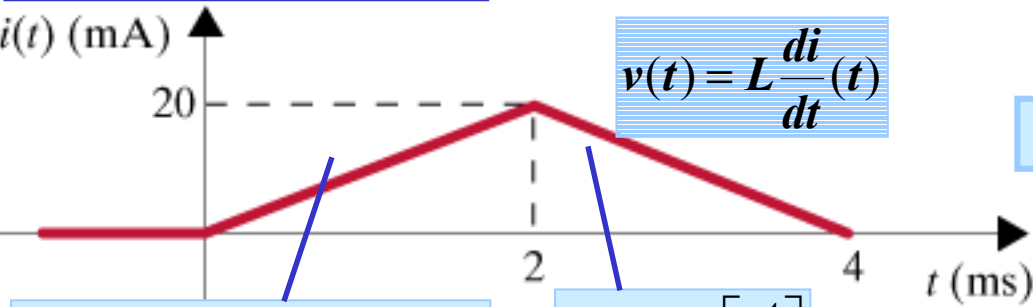
$$w(t_2, t_1) = \frac{1}{2} L i_L^2(t_2) - \frac{1}{2} L i_L^2(t_1)$$

Energy stored on the interval  
Can be positive or negative

$$w_L(t) = \frac{1}{2} L i_L^2(t)$$

“Energy stored at time t”  
Must be non-negative.  
Passive element!!!



**LEARNING EXAMPLE****L=10mH. FIND THE VOLTAGE****ENERGY STORED BETWEEN 2 AND 4 ms**

$$w(4,2) = \frac{1}{2} Li_L^2(4) - \frac{1}{2} Li_L^2(2)$$

$$w(4,2) = 0 - 0.5 * 10 * 10^{-3} (20 * 10^{-3})^2 \text{ J}$$

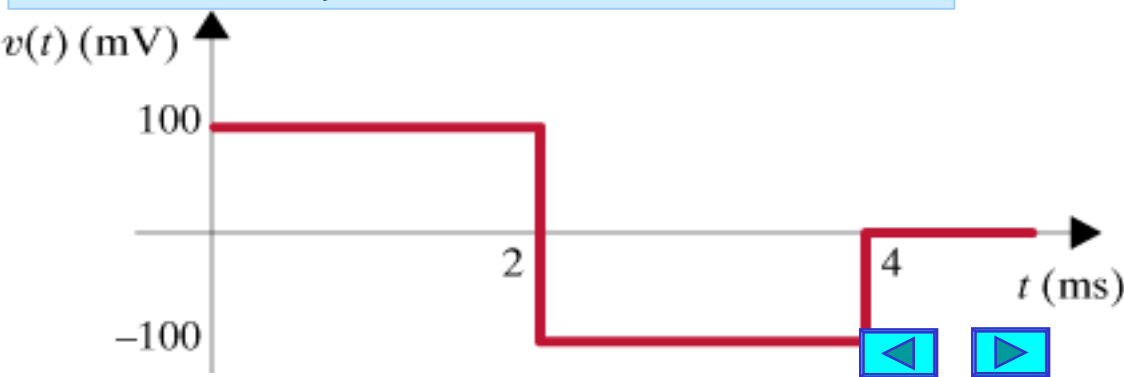
$$m = \frac{20 \times 10^{-3} \text{ A}}{2 \times 10^{-3} \text{ s}} = 10 \left[ \frac{\text{A}}{\text{s}} \right]$$

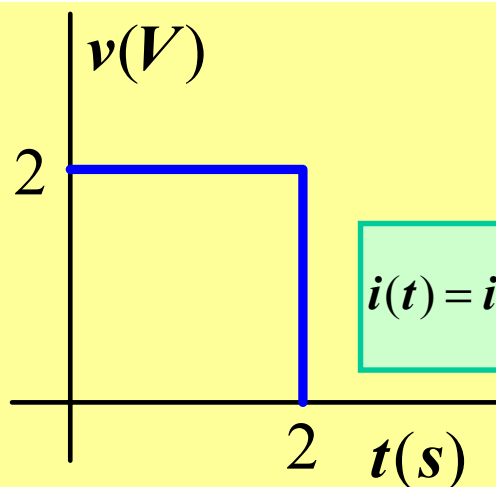
$$m = -10 \left[ \frac{\text{A}}{\text{s}} \right]$$

**THE DERIVATIVE OF A STRAIGHT LINE IS ITS SLOPE**

$$\frac{di}{dt} = \begin{cases} 10(\text{A/s}) & 0 \leq t \leq 2\text{ms} \\ -10(\text{A/s}) & 2 < t \leq 4\text{ms} \\ 0 & \text{elsewhere} \end{cases}$$

$$\left. \begin{array}{l} \frac{di}{dt}(t) = 10(\text{A/s}) \\ L = 10 \times 10^{-3} \text{ H} \end{array} \right\} \Rightarrow v(t) = 100 \times 10^{-3} \text{ V} = 100 \text{ mV}$$

**THE VALUE IS NEGATIVE BECAUSE THE INDUCTOR IS SUPPLYING ENERGY PREVIOUSLY STORED**

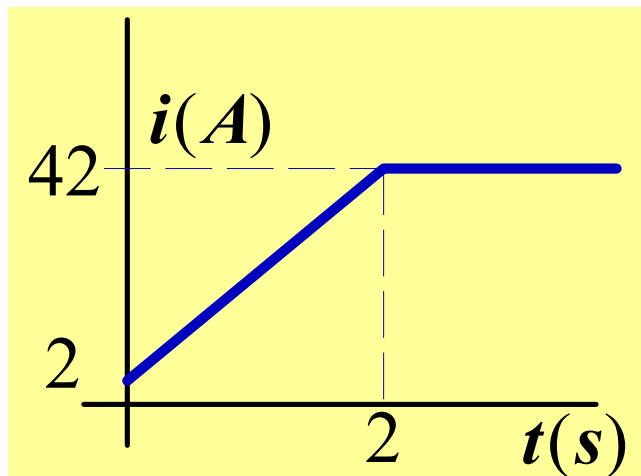
**SAMPLE PROBLEM** $L=0.1H, i(0)=2A$ . Find  $i(t), t>0$ 

$$i(t) = i(0) + \frac{1}{L} \int_0^t v(x) dx$$

$$v(x) = 2 \Rightarrow \int_0^t v(x) dx = 2t; \quad 0 < t \leq 2$$

$$L = 0.1H \Rightarrow i(t) = 2 + 20t; \quad 0 \leq t \leq 2s$$

$$v(x) = 0; \quad t > 2 \Rightarrow i(t) = i(2); \quad t > 2s$$

**ENERGY COMPUTATIONS**

$$w(t_2, t_1) = \frac{1}{2} Li_L^2(t_2) - \frac{1}{2} Li_L^2(t_1)$$

Energy stored on the interval  
Can be positive or negative

**Initial energy stored in inductor**

$$w(0) = 0.5 * 0.1[H](2A)^2 = 0.2[J]$$

**“Total energy stored in the inductor”**

$$w(\infty) = 0.5 * 0.1[H] * (42A)^2 = 88.2J$$

**Energy stored between 0 and 2 sec**

$$w(2,0) = \frac{1}{2} Li_L^2(2) - \frac{1}{2} Li_L^2(0)$$

$$w(2,0) = 0.5 * 0.1 * (42)^2 - 0.5 * 0.1 * (2)^2$$

$$w(2,0) = 88[J]$$



**LEARNING EXAMPLE**

The current in a 2-mH inductor is

$$i(t) = 2 \sin 377t \text{ A}$$

**FIND THE VOLTAGE ACROSS AND THE ENERGY STORED (AS FUNCTION OF TIME)**

$$v(t) = L \frac{di(t)}{dt} = (2 \times 10^{-3}) \frac{d}{dt} (2 \sin 377t)$$

$$v(t) = 1.508 \cos 377t \text{ V}$$

**FOR ENERGY STORED IN THE INDUCTOR**

$$w_L(t) = \frac{1}{2} Li^2(t)$$

$$w_L(t) = \frac{1}{2} (2 \times 10^{-3}) (2 \sin 377t)^2$$

$$= 0.004 \sin^2 377t,$$

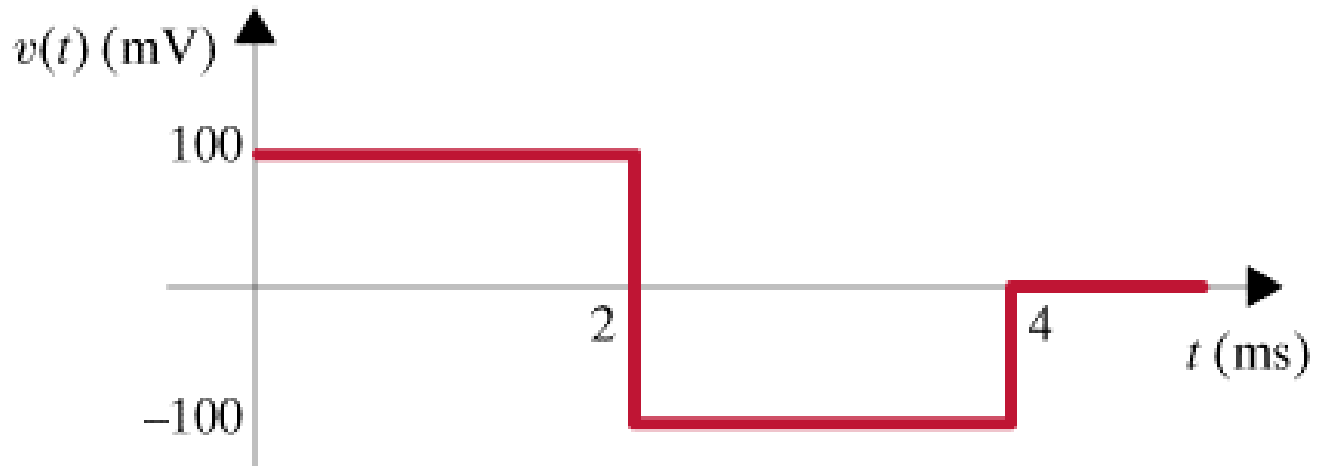
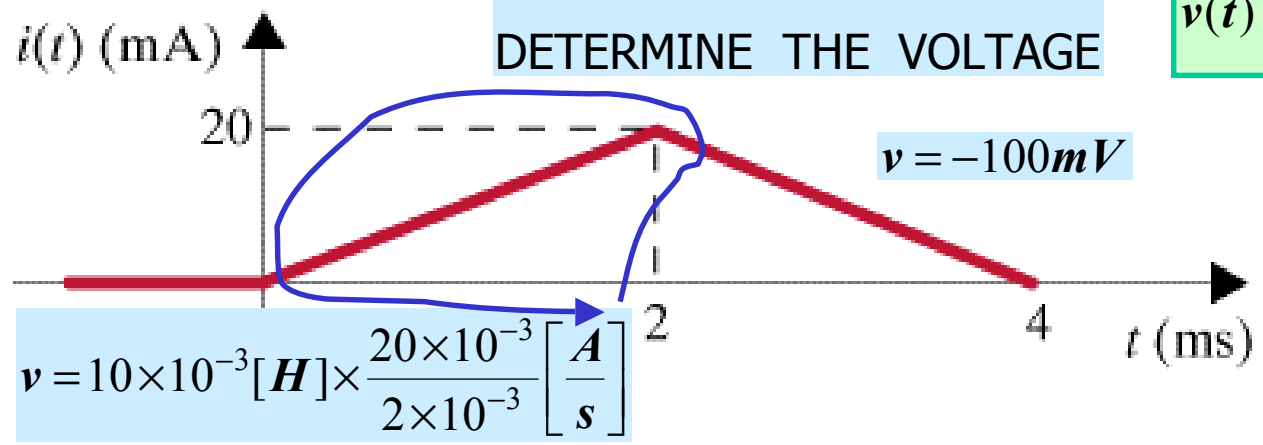
**NOTICE THAT ENERGY STORED AT ANY GIVEN TIME IS NON NEGATIVE  
-THIS IS A PASSIVE ELEMENT-**



# LEARNING EXAMPLE

$L = 10\text{mH}$   
DETERMINE THE VOLTAGE

$$v(t) = L \frac{di}{dt}(t)$$





**LEARNING EXAMPLE**

**FIND THE CURRENT**

Voltage (mV)

$$v(t) = (1 - 3t)e^{-3t} \text{ mV} \quad t \geq 0$$

$$= 0 \quad t < 0$$

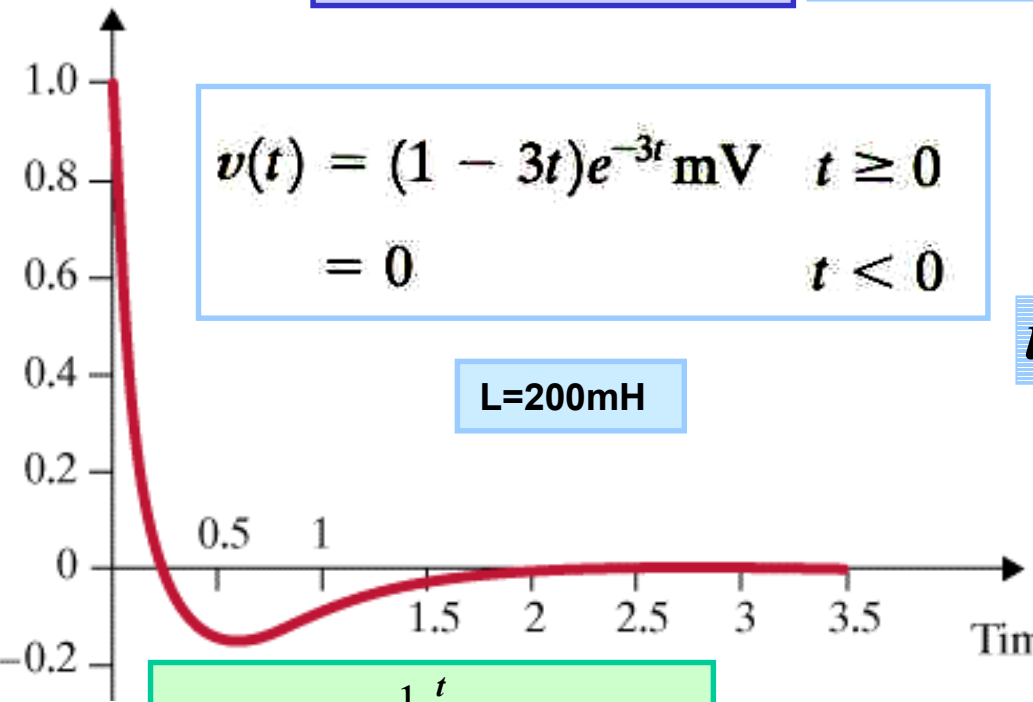
**L=200mH**

$$i(t) = 5 \left\{ \frac{e^{-3x}}{-3} \Big|_0^t - 3 \left[ -\frac{e^{-3x}}{9} (3x + 1) \right] \Big|_0^t \right\}$$

**i(t)**

$$= 5te^{-3t} \text{ mA} \quad t \geq 0$$

$$= 0 \quad t < 0$$



$$i(t) = i(0) + \frac{1}{L} \int_0^t v(x) dx; \quad t > 0$$

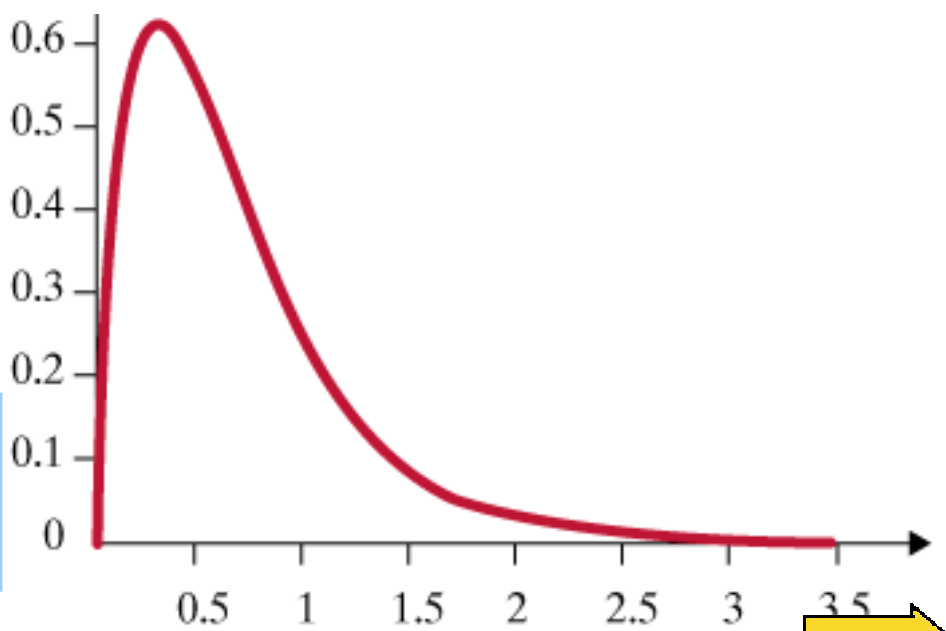
$$v(t) = 0; \quad t < 0 \Rightarrow i(0) = 0$$

$$i(t) = \frac{10^3}{200} \int_0^t (1 - 3x)e^{-3x} dx$$

$$= 5 \left\{ \int_0^t e^{-3x} dx - 3 \int_0^t x e^{-3x} dx \right\}$$

(mA)

Time (s)



$L=200\text{mH}$

$$v(t) = (1 - 3t)e^{-3t} \text{ mV} \quad t \geq 0$$
$$= 0 \quad t < 0$$

$$i(t) = 5te^{-3t} \text{ mA} \quad t \geq 0$$
$$= 0 \quad t < 0$$

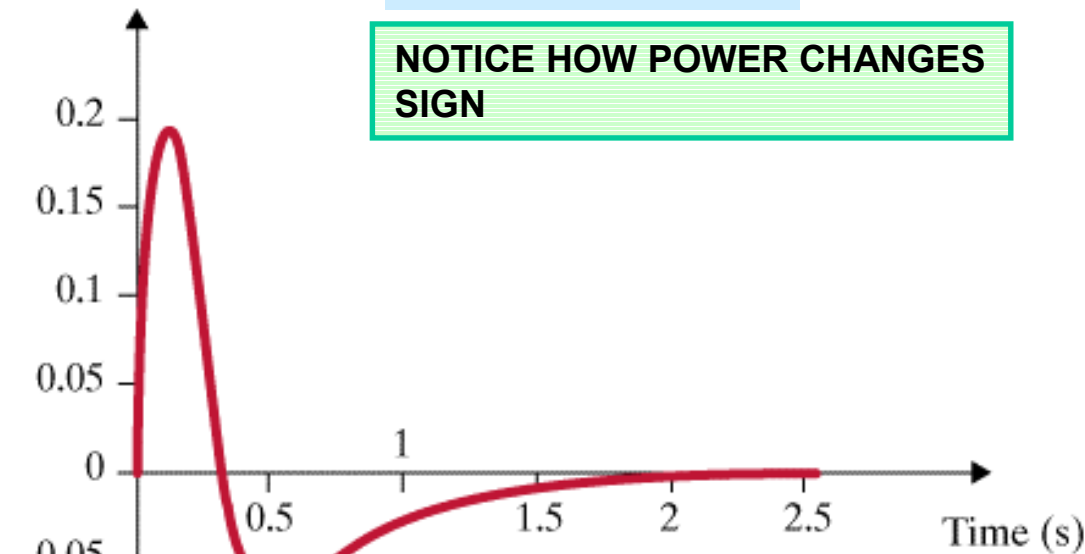
POWER  $p(t) = v(t)i(t)$

$$p(t) = 5t(1 - 3t)e^{-6t} \mu\text{W} \quad t \geq 0$$
$$= 0 \quad t < 0$$

ENERGY  $w(t) = \frac{1}{2} Li^2(t)$

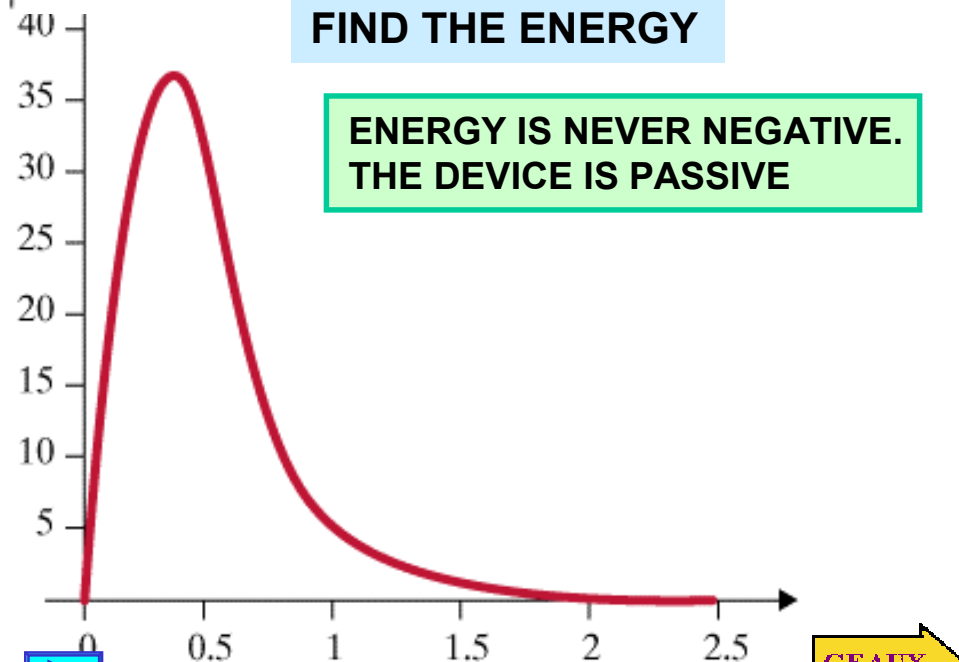
$$w(t) = 2.5t^2 e^{-6t} \mu\text{J} \quad t \geq 0$$
$$= 0 \quad t < 0$$

Power ( $\mu\text{W}$ )



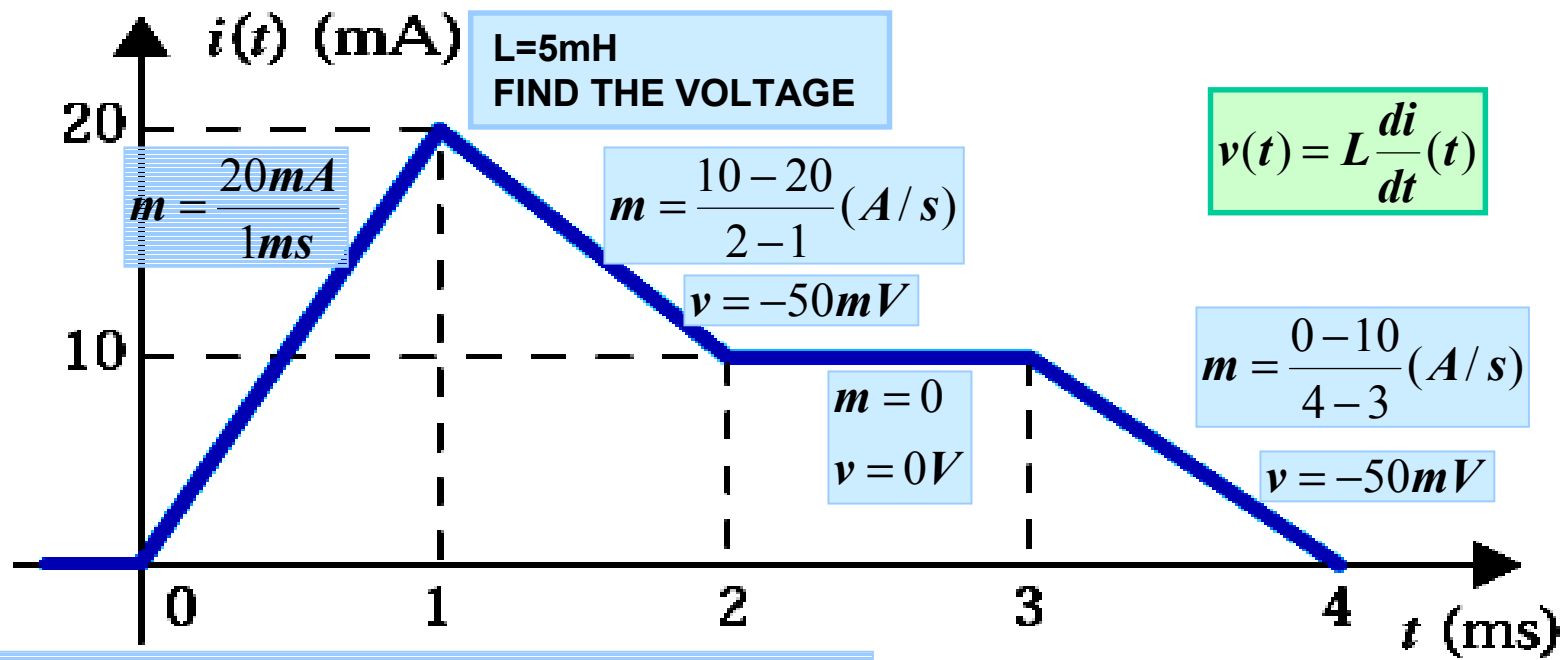
**FIND THE ENERGY**

ENERGY IS NEVER NEGATIVE.  
THE DEVICE IS PASSIVE

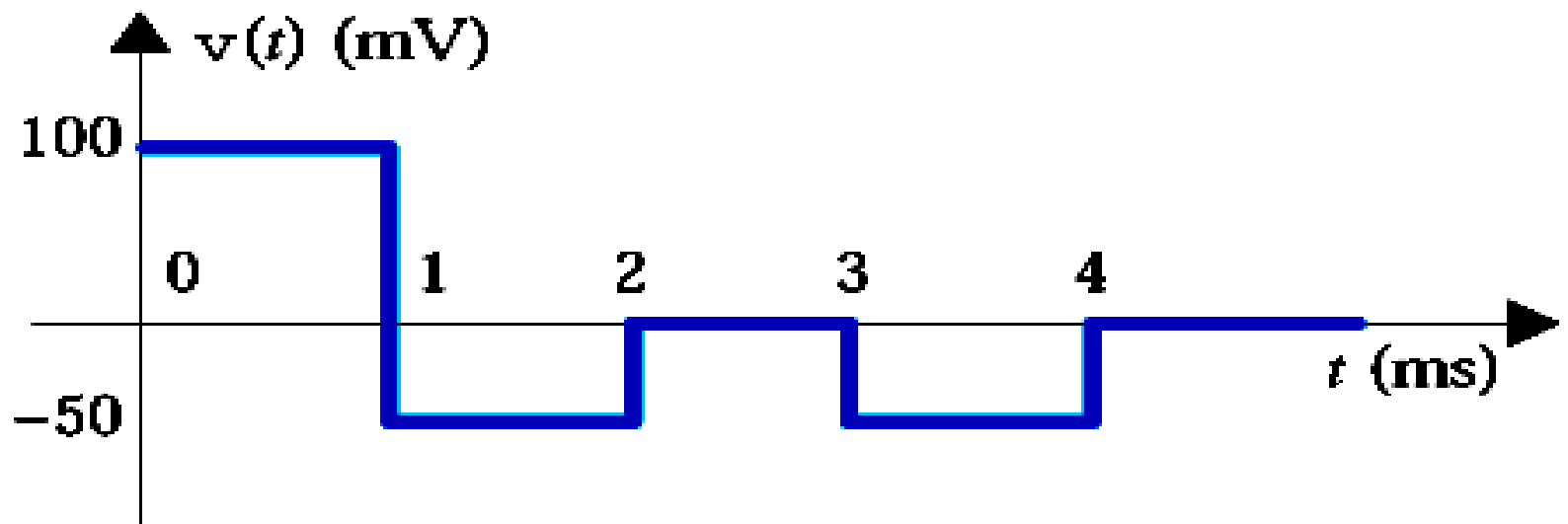


GEAUX

**LEARNING EXTENSION**



$v = 5 \times 10^{-3} (H) \times 20 (A/s); 0 \leq t < 1\text{ms} = 100\text{mV}$



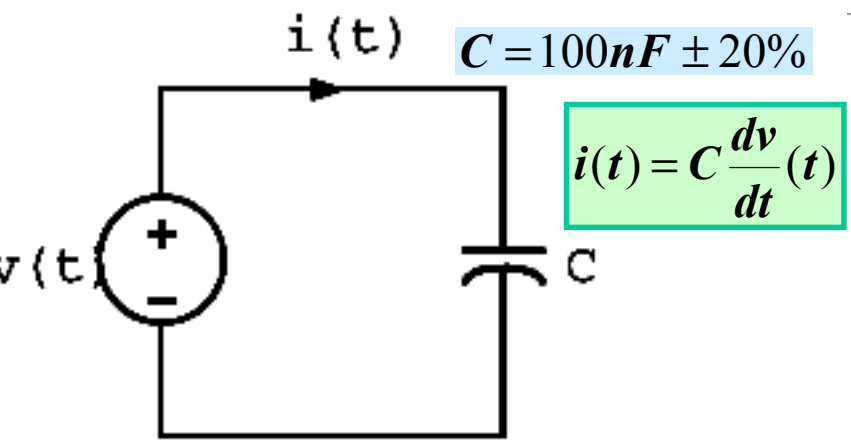
# CAPACITOR SPECIFICATIONS

CAPACITANCE RANGE  $pF \approx C \approx 50mF$   
 IN STANDARD VALUES

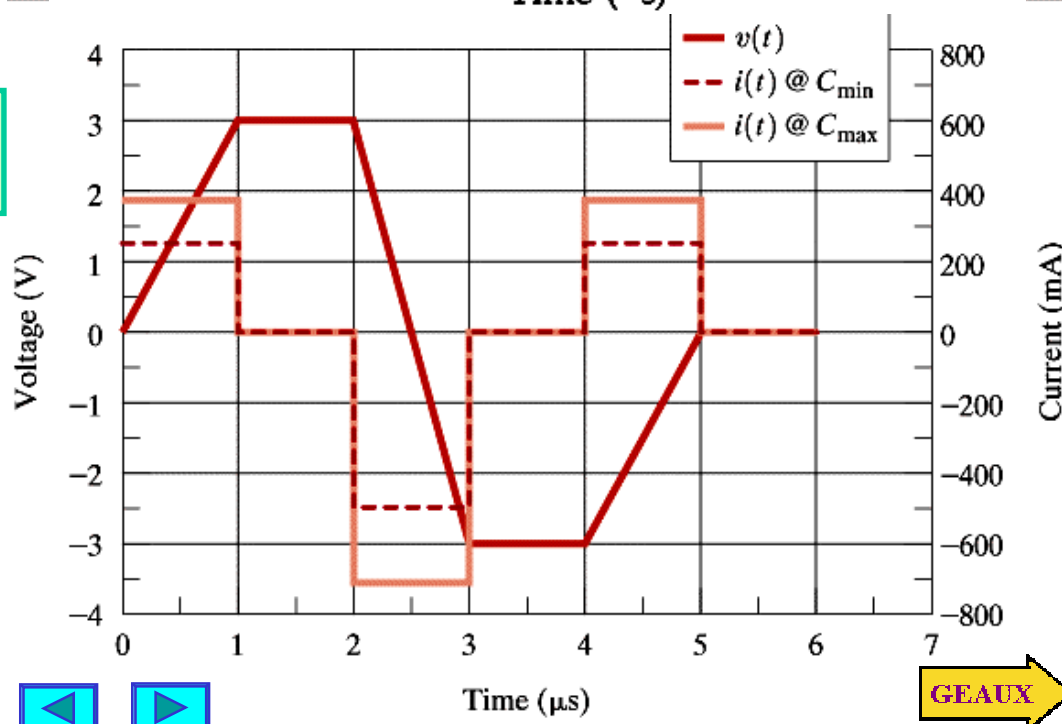
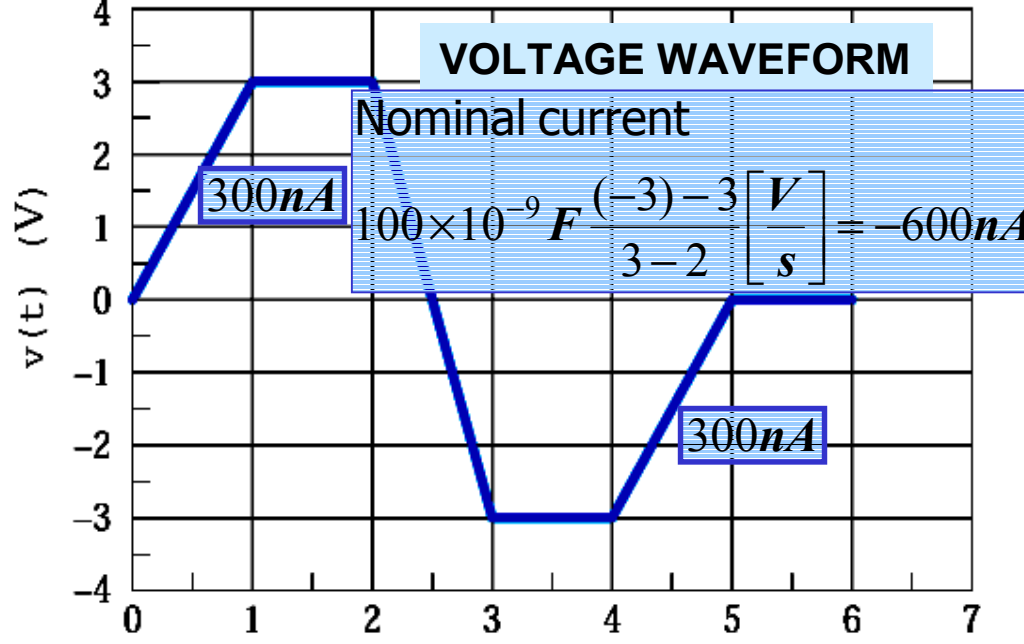
STANDARD CAPACITOR RATINGS  
 6.3V – 500V

STANDARD TOLERANCE  
 $\pm 5\%$ ,  $\pm 10\%$ ,  $\pm 20\%$

## LEARNING EXAMPLE



GIVEN THE VOLTAGE WAVEFORM  
 DETERMINE THE VARIATIONS IN  
 CURRENT



## INDUCTOR SPECIFICATIONS

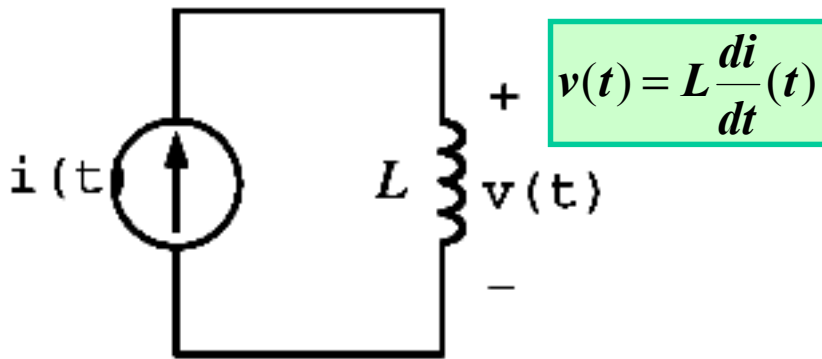
INDUCTANCE RANGES  $\approx 1\text{nH} \leq L \leq \approx 100\text{H}$   
 IN STANDARD VALUES

STANDARD INDUCTOR RATINGS  
 $\approx \text{mA} - \approx 1\text{A}$

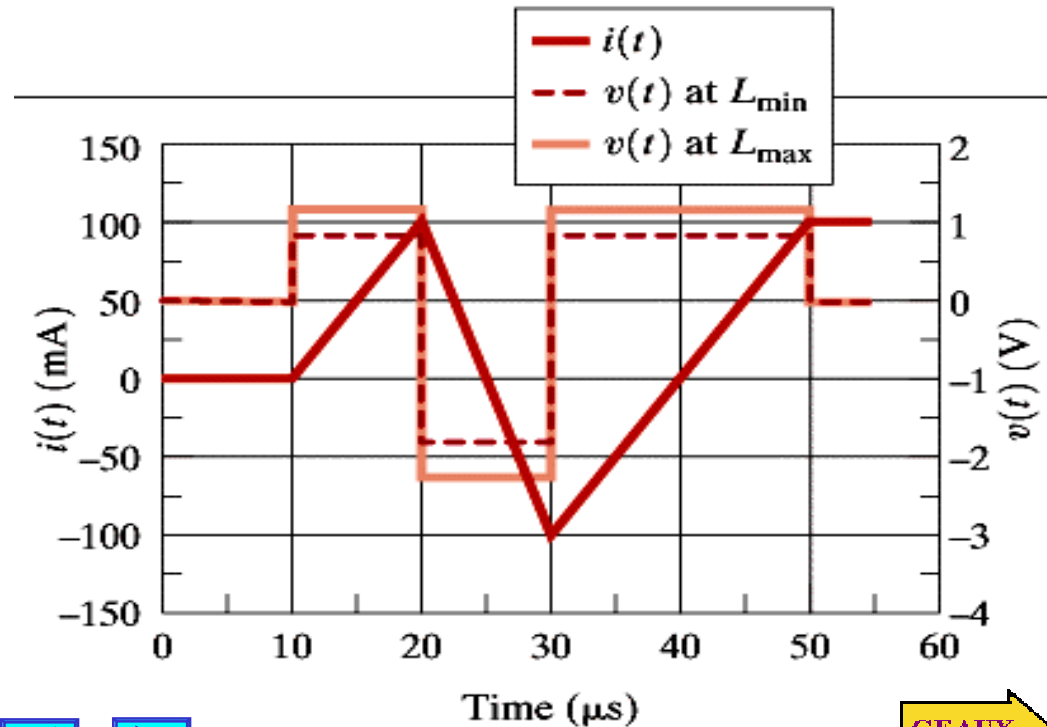
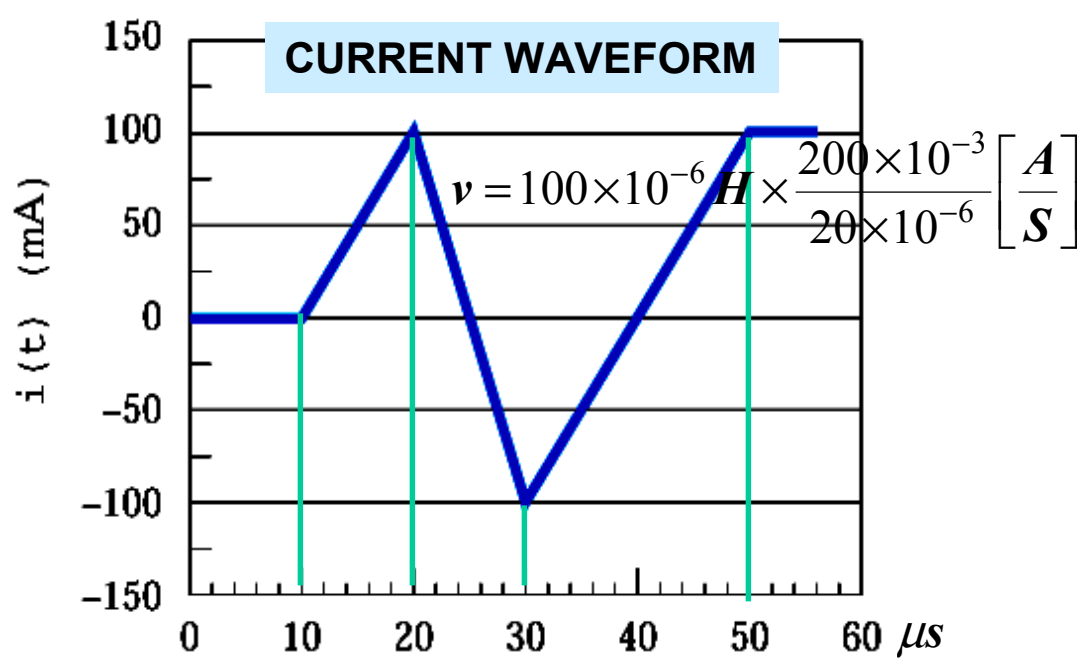
STANDARD TOLERANCE  
 $\pm 5\%, \pm 10\%$

## LEARNING EXAMPLE

$$L = 100\mu\text{H} \pm 10\%$$



GIVEN THE CURRENT WAVEFORM  
 DETERMINE THE VARIATIONS IN  
 VOLTAGE



## The Dual Relationship for Capacitors and Inductors

### Capacitor

$$i(t) = C \frac{dv(t)}{dt}$$

$$v(t) = \frac{1}{C} \int_{t_0}^t i(x) dx + v(t_0)$$

$$p(t) = Cv(t) \frac{dv(t)}{dt}$$

$$w(t) = \frac{1}{2} Cv(t)^2$$

$C \rightarrow L$

$v \rightarrow i$

$i \rightarrow v$

### Inductor

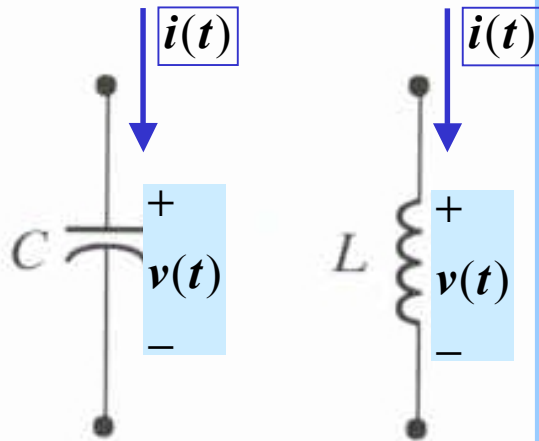
$$v(t) = L \frac{di(t)}{dt}$$

$$i(t) = \frac{1}{L} \int_{t_0}^t v(x) dx + i(t_0)$$

$$p(t) = Li(t) \frac{di(t)}{dt}$$

$$w(t) = \frac{1}{2} Li^2(t)$$

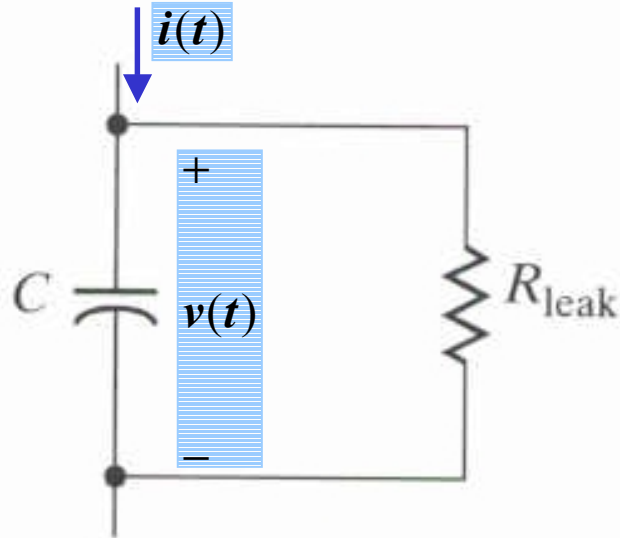
# IDEAL AND PRACTICAL ELEMENTS



IDEAL ELEMENTS

$$i(t) = C \frac{dv}{dt}(t)$$

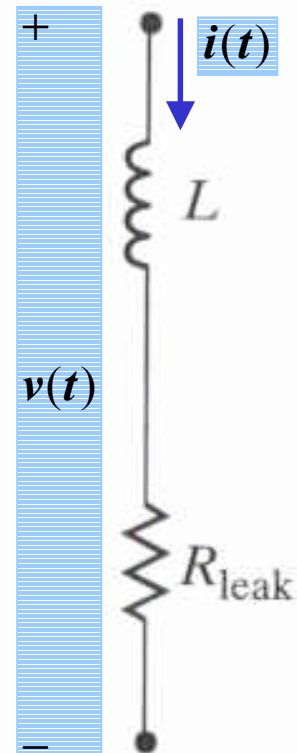
$$v(t) = L \frac{di}{dt}(t)$$



CAPACITOR/INDUCTOR MODELS INCLUDING LEAKAGE RESISTANCE

$$i(t) = \frac{v(t)}{R_{leak}} + C \frac{dv}{dt}(t)$$

MODEL FOR "LEAKY" CAPACITOR

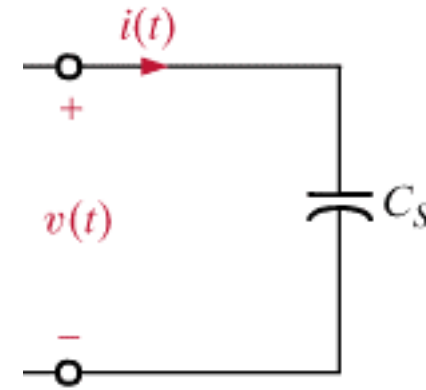
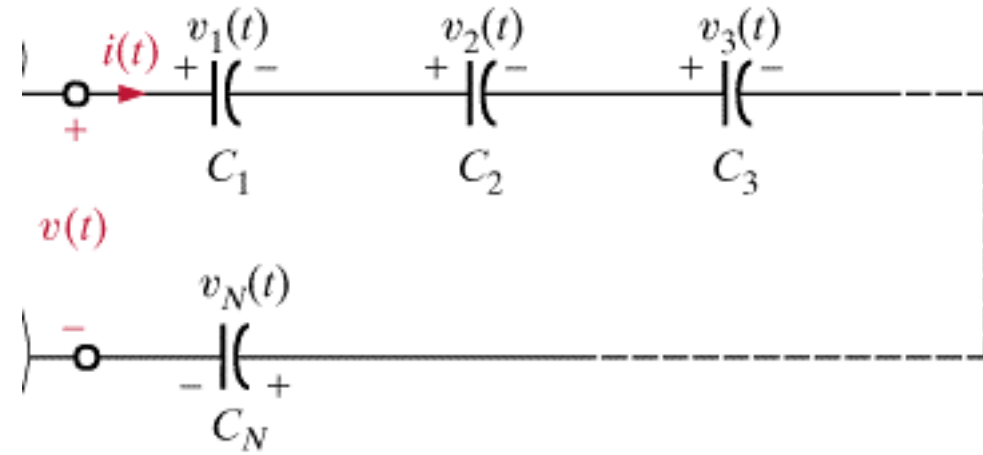


$$v(t) = R_{leak} i(t) + L \frac{di}{dt}(t)$$

MODEL FOR "LEAKY" INDUCTORS



# SERIES CAPACITORS



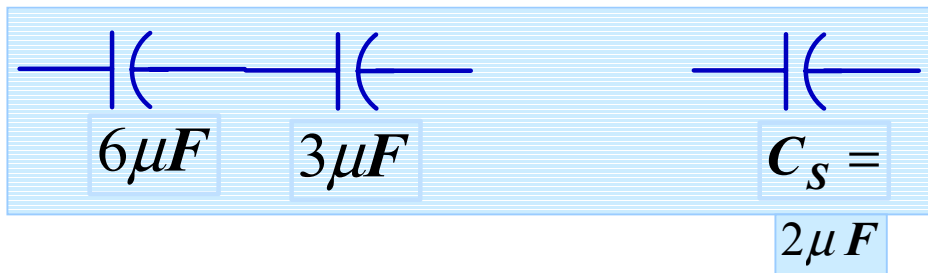
$$v(t) = v_1(t) + v_2(t) + v_3(t) + \dots + v_N(t)$$

$$C_s = \frac{C_1 C_2}{C_1 + C_2}$$

Series Combination of two capacitors

$$v_i(t) = \frac{1}{C_i} \int_{t_0}^t i(t) dt + v_i(t_0)$$

$$v(t) = \left( \sum_{i=1}^N \frac{1}{C_i} \right) \int_{t_0}^t i(t) dt + \sum_{i=1}^N v_i(t_0)$$



$$\frac{1}{C_s} = \sum_{i=1}^N \frac{1}{C_i} = \frac{1}{C_1} + \frac{1}{C_2} + \dots + \frac{1}{C_N}$$

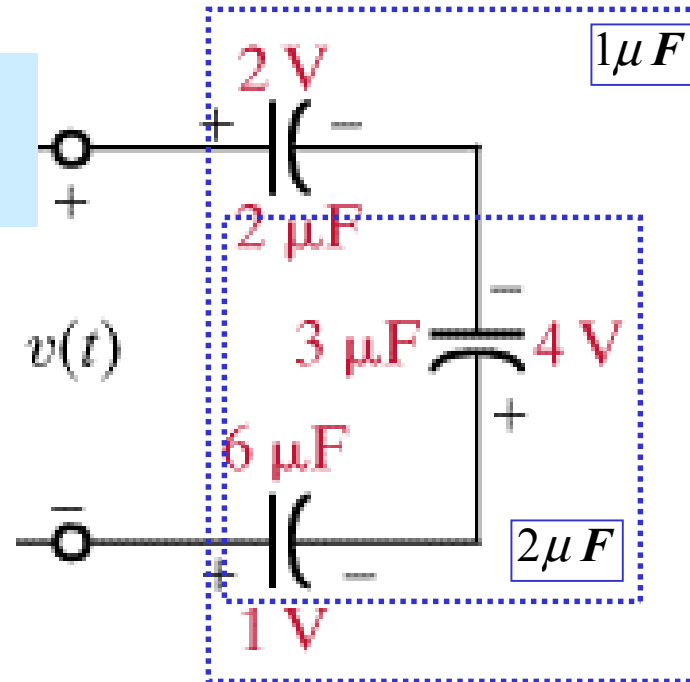
$$v(t_0) = \sum_{i=1}^N v_i(t_0)$$

NOTICE SIMILARITY WITH RESISTORS IN PARALLEL



## LEARNING EXAMPLE

DETERMINE EQUIVALENT CAPACITANCE AND THE INITIAL VOLTAGE



$$\frac{1}{C_s} = \frac{1}{2} + \frac{1}{3} + \frac{1}{6} = \frac{3+2+1}{6}$$

OR WE CAN REDUCE TWO AT A TIME

$$v(t_0) = -3V = +2V - 4V - 1V$$

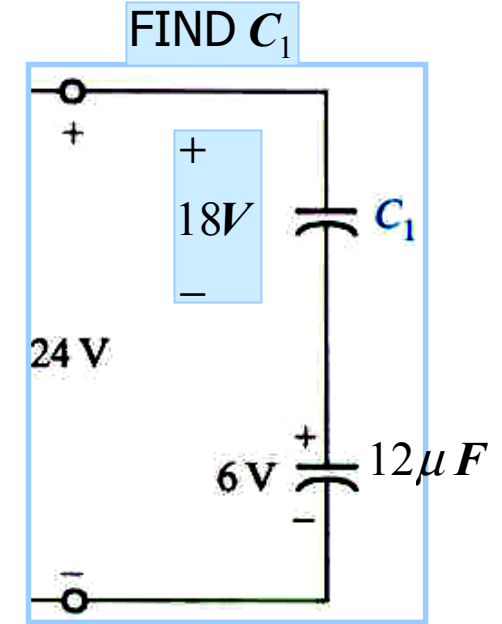
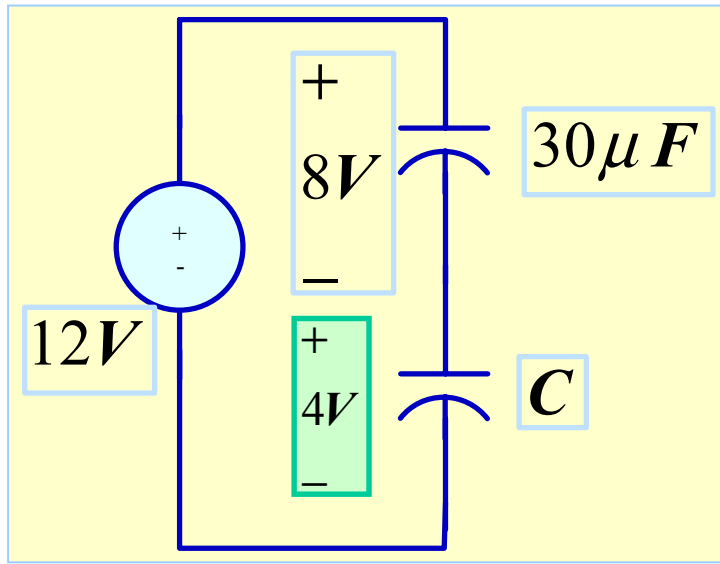
ALGEBRAIC SUM OF INITIAL VOLTAGES

POLARITY IS DICTATED BY THE REFERENCE DIRECTION FOR THE VOLTAGE



**LEARNING EXAMPLE**

Two uncharged capacitors are connected as shown.  
Find the unknown capacitance



**SAME CURRENT. CONNECTED FOR THE SAME TIME PERIOD**

**SAME CHARGE ON BOTH CAPACITORS**

$$Q = (30\mu F)(8V) = 240\mu C$$

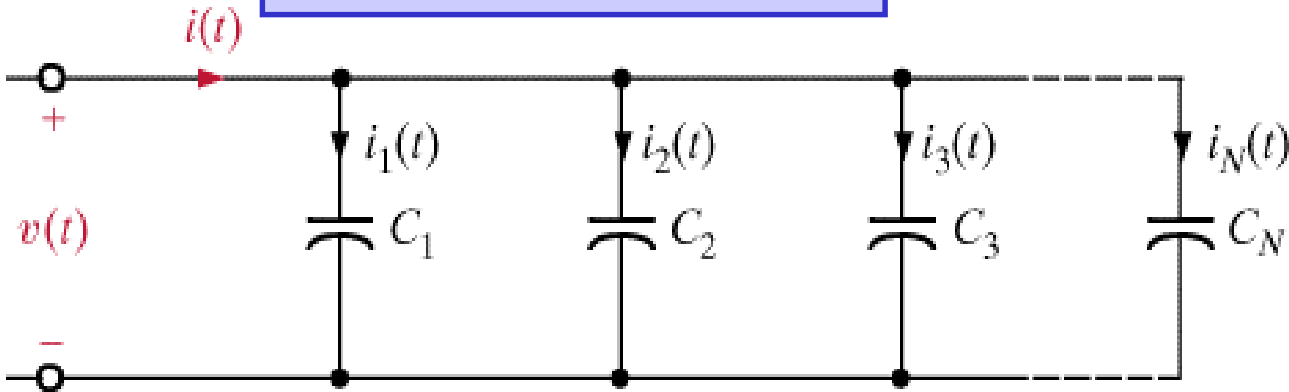
$$C = \frac{Q}{V} = \frac{240\mu C}{4V} = 60\mu F$$

$$Q = CV \Rightarrow Q = (12\mu F)(6V) = 72\mu C$$

$$C_1 = \frac{72\mu C}{18V} = 4\mu F$$



# PARALLEL CAPACITORS



$$i(t) = i_1(t) + i_2(t) + i_3(t) + \dots + i_N(t)$$

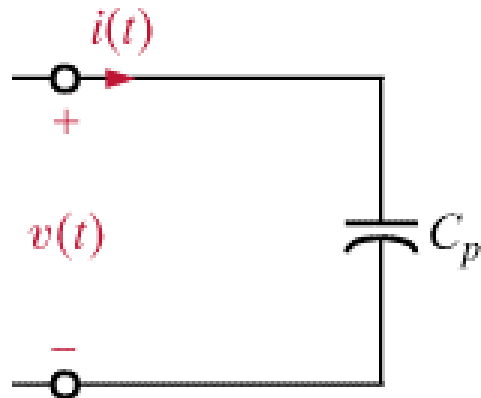
$$i_k(t) = C_k \frac{dv}{dt}(t)$$

$$= C_1 \frac{dv(t)}{dt} + C_2 \frac{dv(t)}{dt} + C_3 \frac{dv(t)}{dt} + \dots + C_N \frac{dv(t)}{dt}$$

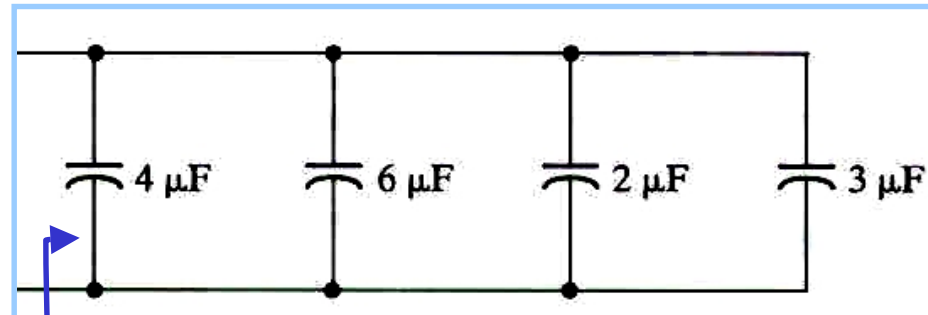
$i(t)$

$$= \left( \sum_{i=1}^N C_i \right) \frac{dv(t)}{dt}$$

$$C_p = C_1 + C_2 + C_3 + \dots + C_N$$



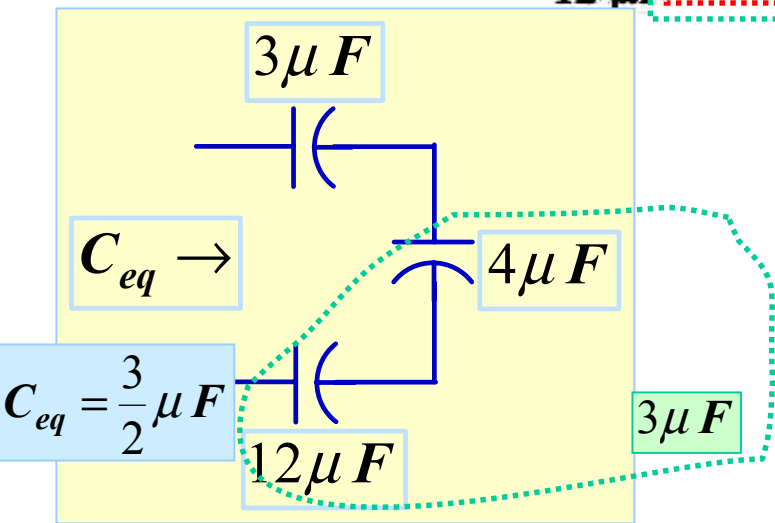
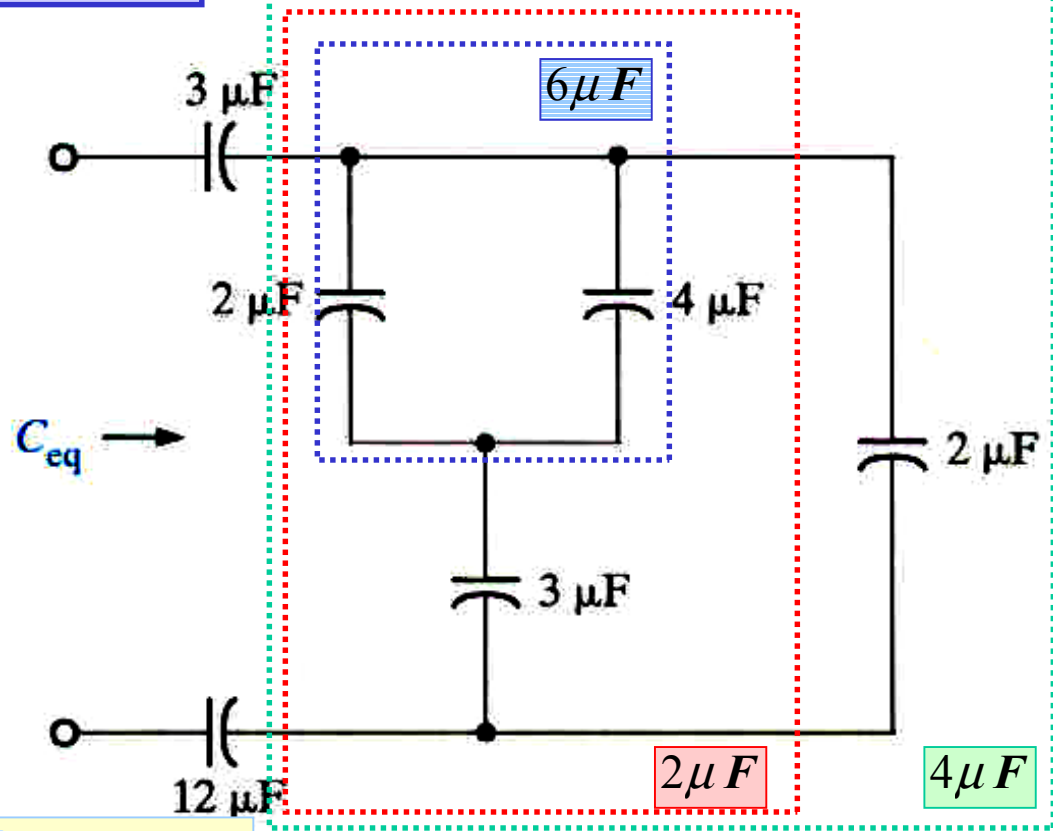
## LEARNING EXAMPLE



$$C_p = 4 + 6 + 2 + 3 = 15 \mu\text{F}$$



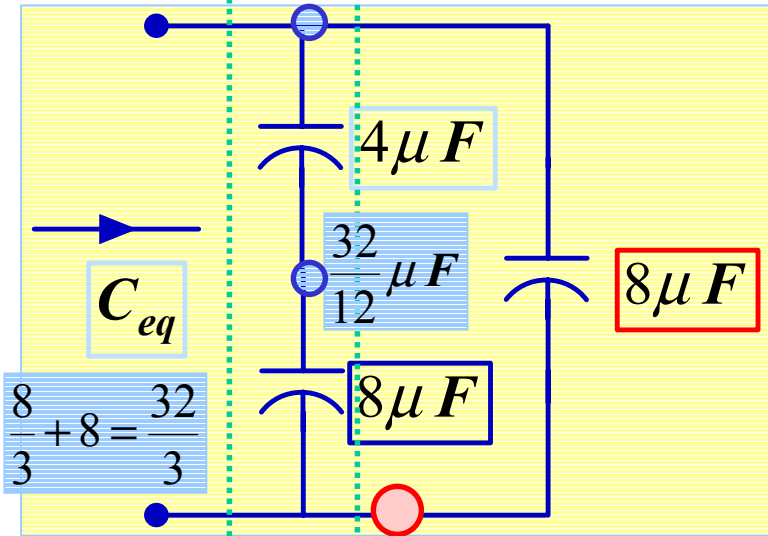
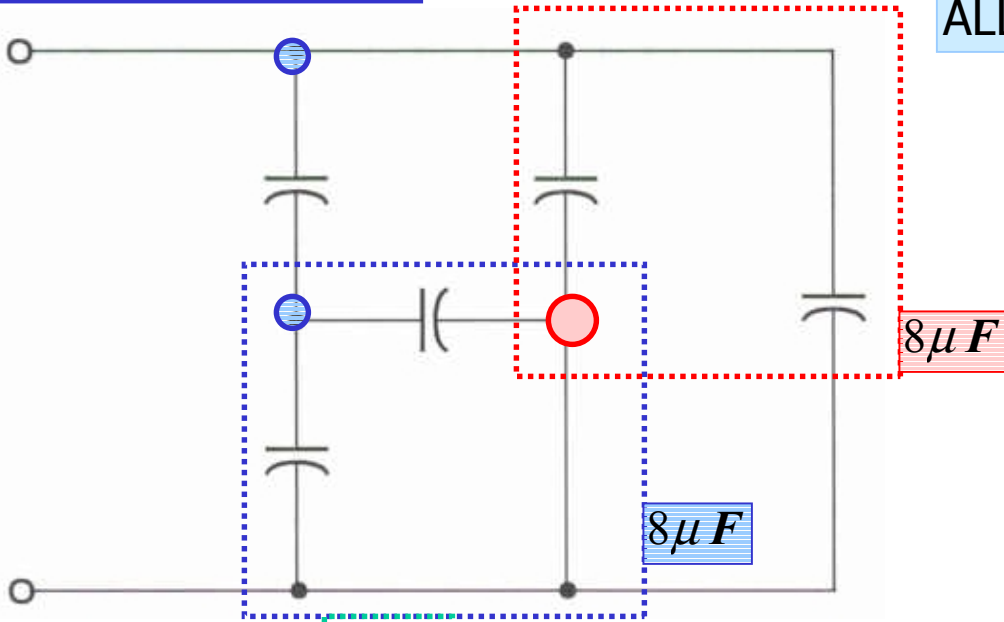
# LEARNING EXTENSION



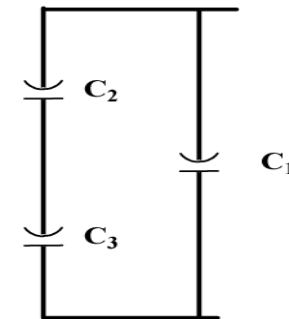
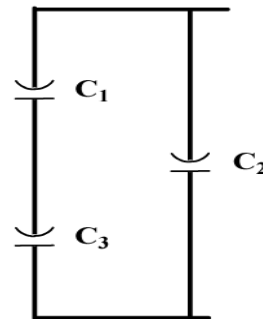
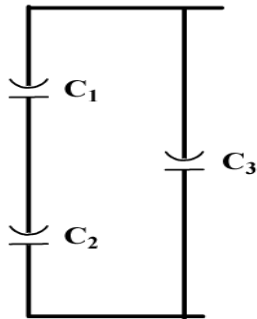
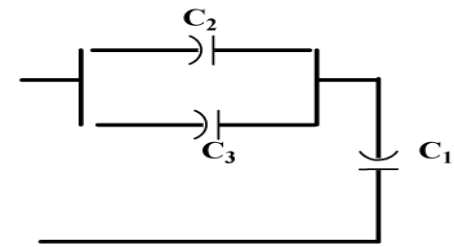
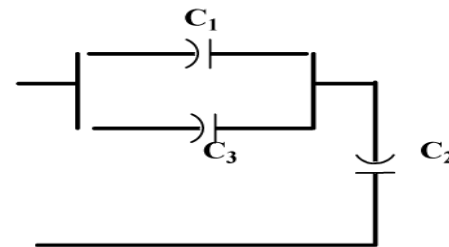
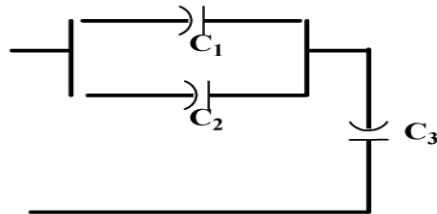
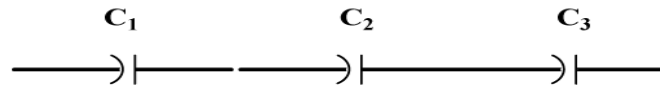
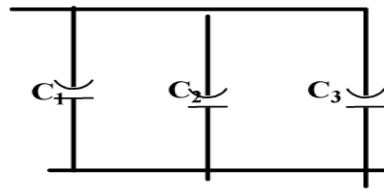
**SAMPLE PROBLEM**

**FIND EQUIVALENT CAPACITANCE**

ALL CAPACITORS ARE  $4\mu F$



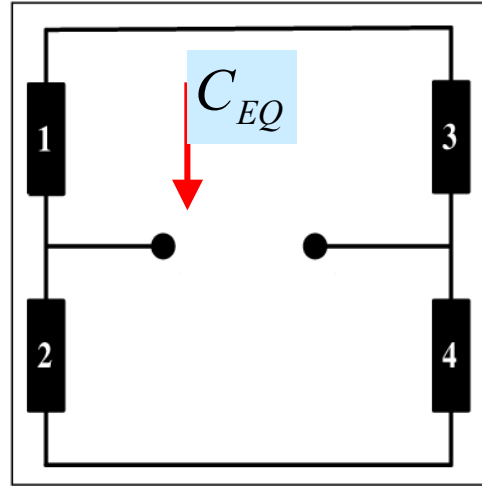
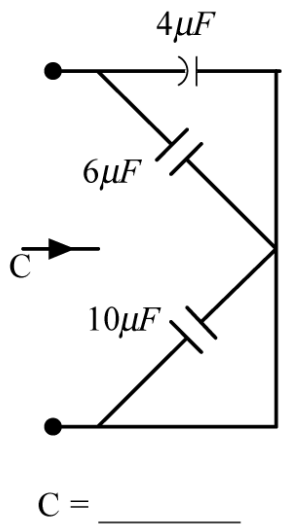
# SAMPLE PROBLEM



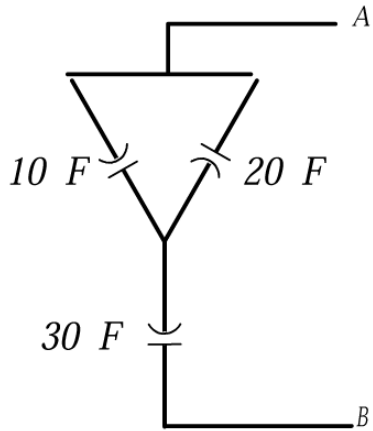
**IF ALL CAPACITORS HAVE THE SAME CAPACITANCE VALUE  $C$   
DETERMINE THE VARIOUS EQUIVALENT CAPACITANCES**



# Examples of interconnections



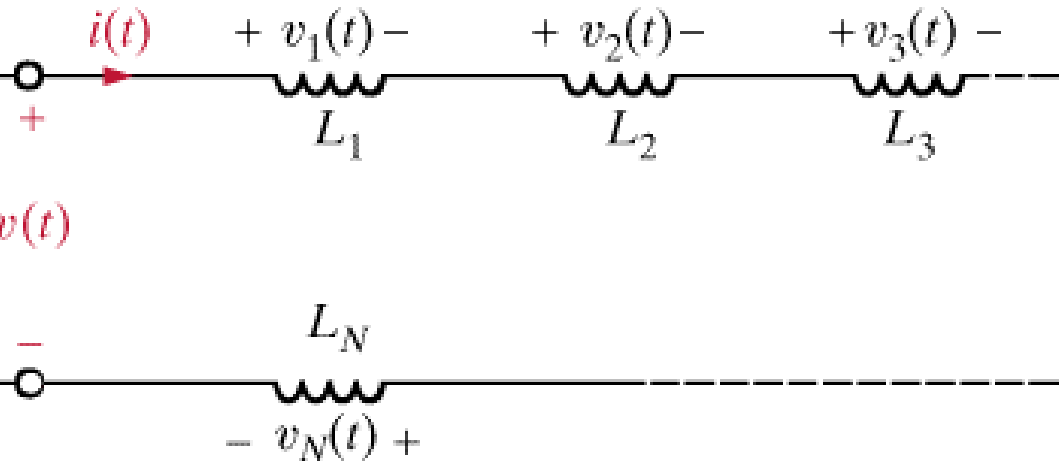
All capacitors are equal with  $C=8$  microFarads



$C_{AB} = \underline{\hspace{2cm}}$



# SERIES INDUCTORS

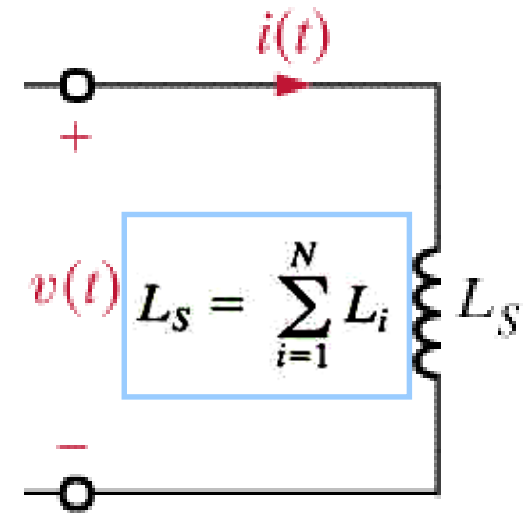


$$v(t) = v_1(t) + v_2(t) + v_3(t) + \dots + v_N(t)$$

$$v_k(t) = L_k \frac{di}{dt}(t)$$

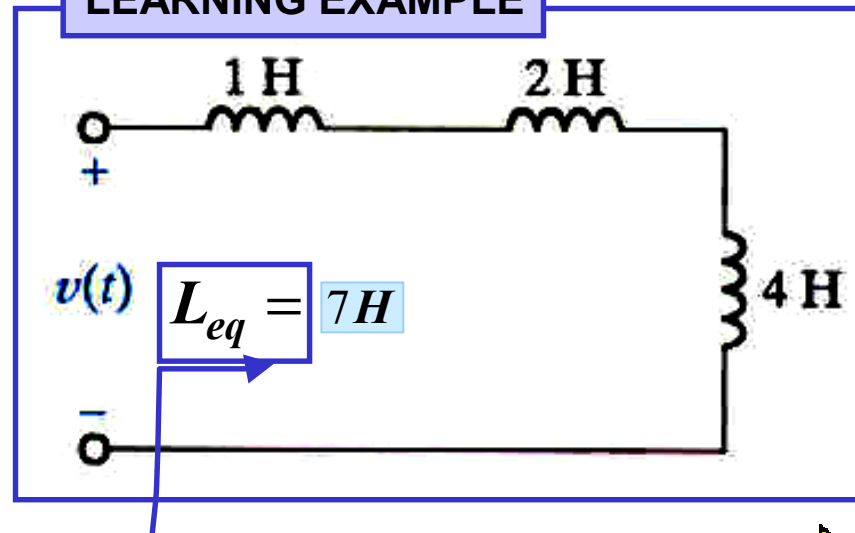
$$\begin{aligned} v(t) &= L_1 \frac{di(t)}{dt} + L_2 \frac{di(t)}{dt} + L_3 \frac{di(t)}{dt} + \dots + L_N \frac{di(t)}{dt} \\ &= \left( \sum_{i=1}^N L_i \right) \frac{di(t)}{dt} \end{aligned}$$

$$L_S = \sum_{i=1}^N L_i$$



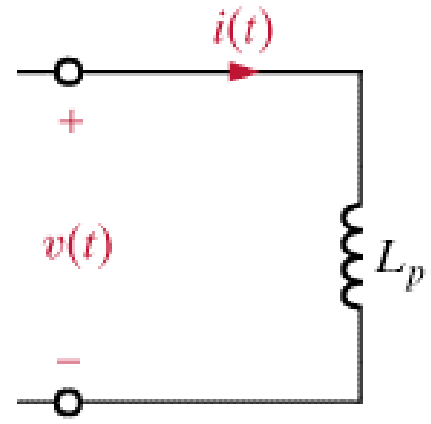
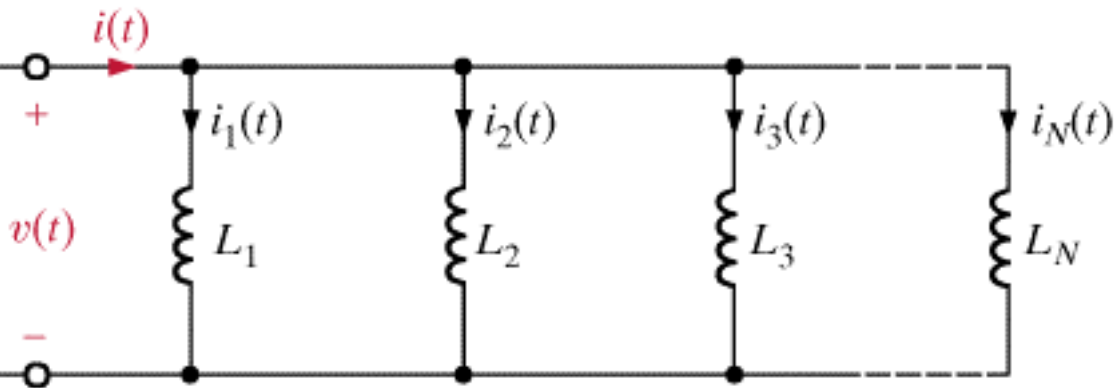
$$v(t) = L_S \frac{di}{dt}(t)$$

# LEARNING EXAMPLE





# PARALLEL INDUCTORS



$$i(t) = i_1(t) + i_2(t) + i_3(t) + \dots + i_N(t)$$

$$i_j(t) = \frac{1}{L_j} \int_{t_0}^t v(x) dx + i_j(t_0)$$

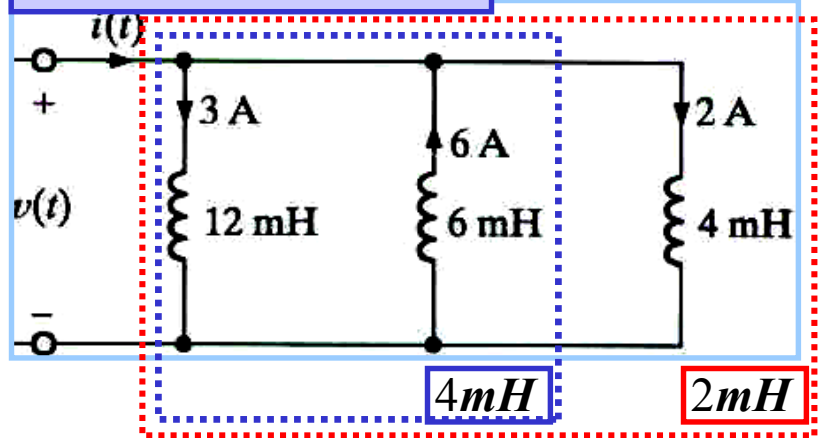
$$i(t) = \frac{1}{L_p} \int_{t_0}^t v(x) dx + i(t_0)$$

$$i(t) = \left( \sum_{j=1}^N \frac{1}{L_j} \right) \int_{t_0}^t v(x) dx + \sum_{j=1}^N i_j(t_0)$$

$$\frac{1}{L_p} = \frac{1}{L_1} + \frac{1}{L_2} + \frac{1}{L_3} + \dots + \frac{1}{L_N}$$

$$i(t_0) = \sum_{j=1}^N i_j(t_0)$$

## LEARNING EXAMPLE

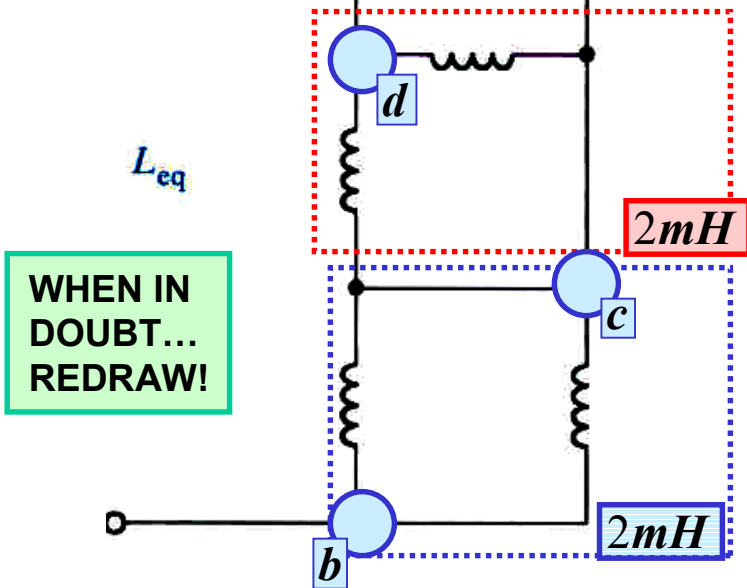


$$i(t_0) = 3A - 6A + 2A = -1A$$

**INDUCTORS COMBINE LIKE RESISTORS  
CAPACITORS COMBINE LIKE CONDUCTANCES**

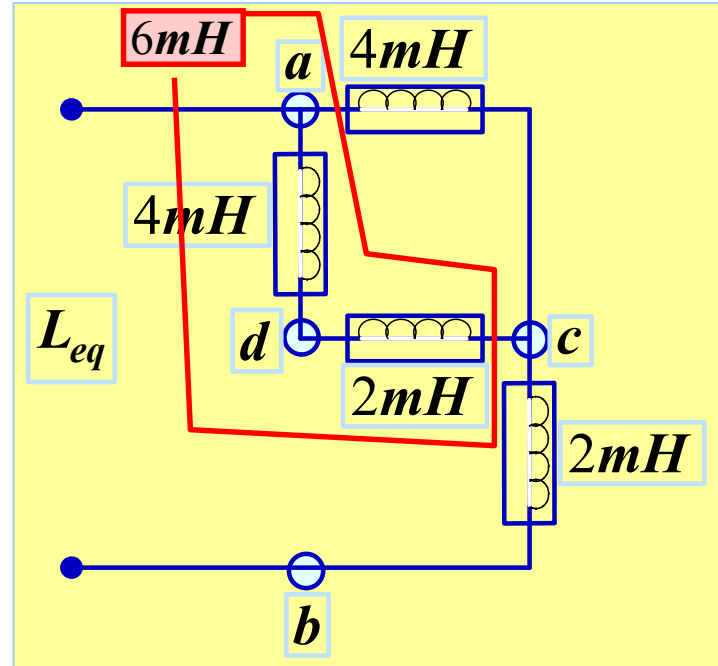
# LEARNING EXTENSION

ALL INDUCTORS ARE 4mH



WHEN IN DOUBT... REDRAW!

# CONNECT COMPONENTS BETWEEN NODES



$$L_{eq} = (6mH \parallel 4mH) + 2mH = 4.4mH$$

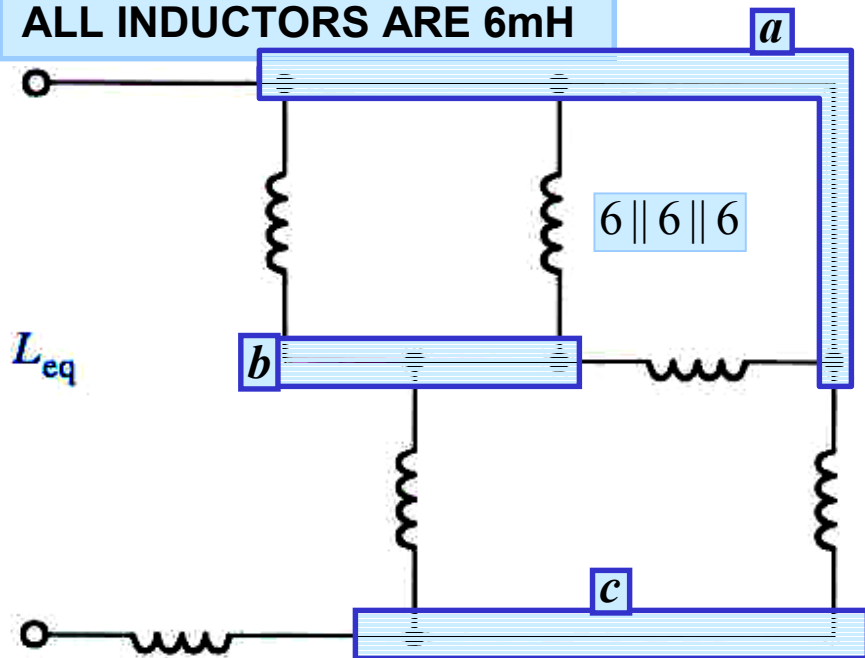
IDENTIFY ALL NODES

PLACE NODES IN CHOSEN LOCATIONS



# LEARNING EXTENSION

ALL INDUCTORS ARE 6mH

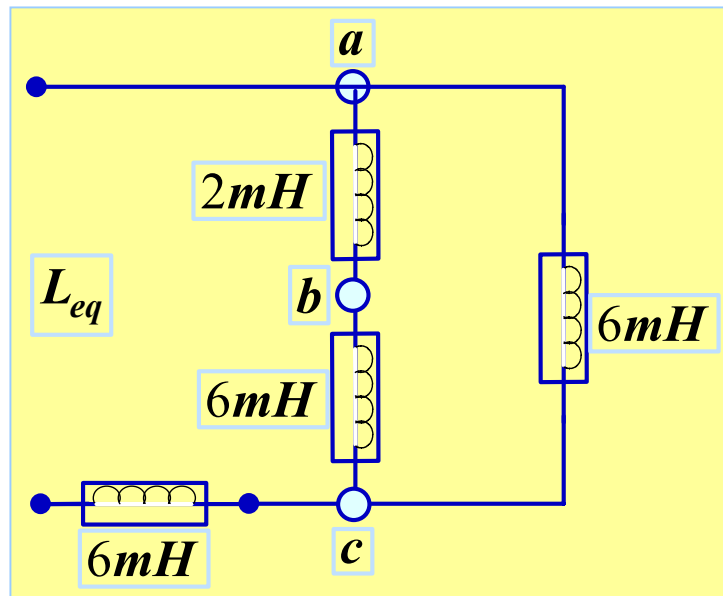


$L_{eq}$

NODES CAN HAVE COMPLICATED SHAPES. KEEP IN MIND DIFFERENCE BETWEEN PHYSICAL LAYOUT AND ELECTRICAL CONNECTIONS



SELECTED LAYOUT



$$L_{eq} = 6 + [(6 + 2) \parallel 6] = 6 + \frac{48}{14} = 6 \frac{24}{7} \text{ mH}$$

$$L_{eq} = \frac{66}{7} \text{ mH}$$



# SUMMARY

- The important (dual) relationships for capacitors and inductors are as follows:

$$q = Cv$$

$$i(t) = C \frac{dv(t)}{dt}$$

$$v(t) = L \frac{di(t)}{dt}$$

$$v(t) = \frac{1}{C} \int_{-\infty}^t i(x) dx$$

$$i(t) = \frac{1}{L} \int_{-\infty}^t v(x) dx$$

$$p(t) = Cv(t) \frac{dv(t)}{dt}$$

$$p(t) = Li(t) \frac{di(t)}{dt}$$

$$W_C(t) = 1/2Cv^2(t)$$

$$W_L(t) = 1/2Li^2(t)$$

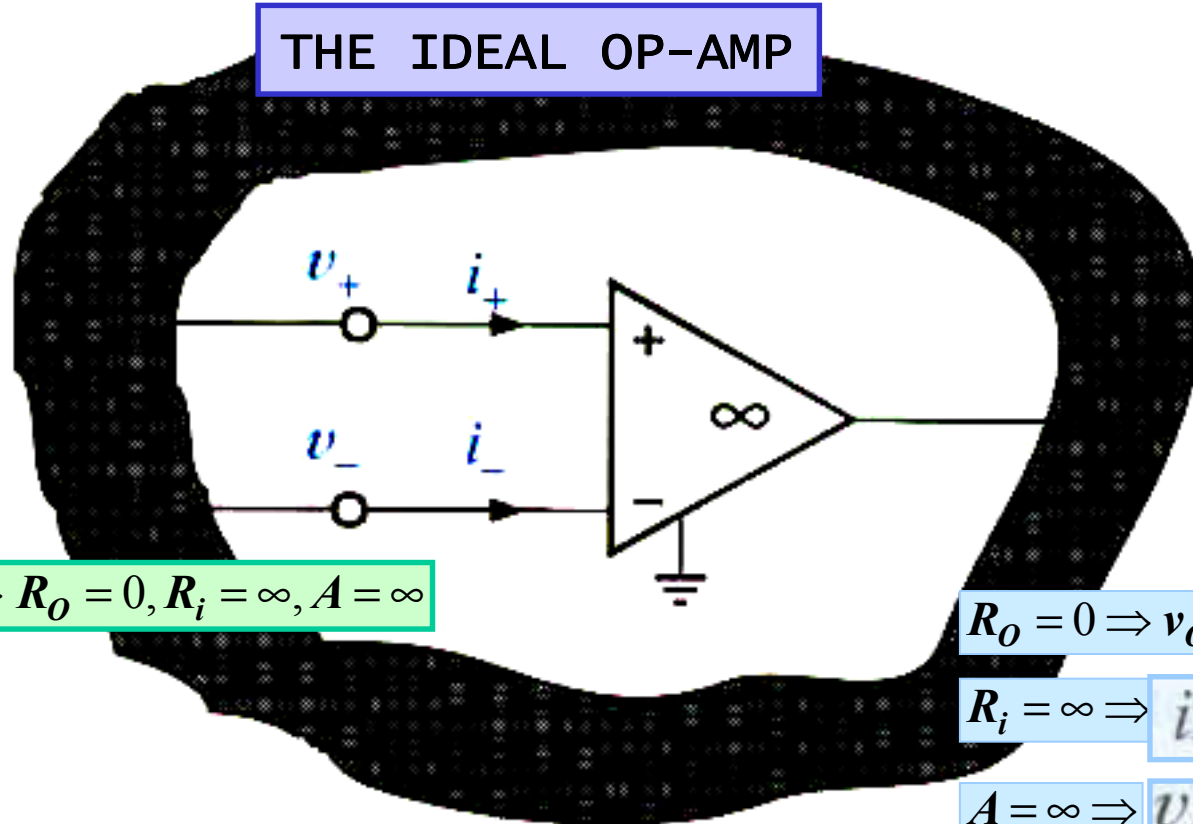
- The passive sign convention is used with capacitors and inductors.
- In dc steady state a capacitor looks like an open circuit and an inductor looks like a short circuit.
- Leakage resistance is present in practical capacitors and inductors.
- When capacitors are interconnected, their equivalent capacitance is determined as follows: Capacitors in series combine like resistors in parallel and capacitors in parallel combine like resistors in series.
- When inductors are interconnected, their equivalent inductance is determined as follows: Inductors in series combine like resistors in series and inductors in parallel combine like resistors in parallel.
- RC operational amplifier circuits can be used to differentiate or integrate an electrical signal.



# RC OPERATIONAL AMPLIFIER CIRCUITS

INTRODUCES TWO VERY IMPORTANT PRACTICAL CIRCUITS  
BASED ON OPERATIONAL AMPLIFIERS

## THE IDEAL OP-AMP



IDEAL  $\Rightarrow R_o = 0, R_i = \infty, A = \infty$

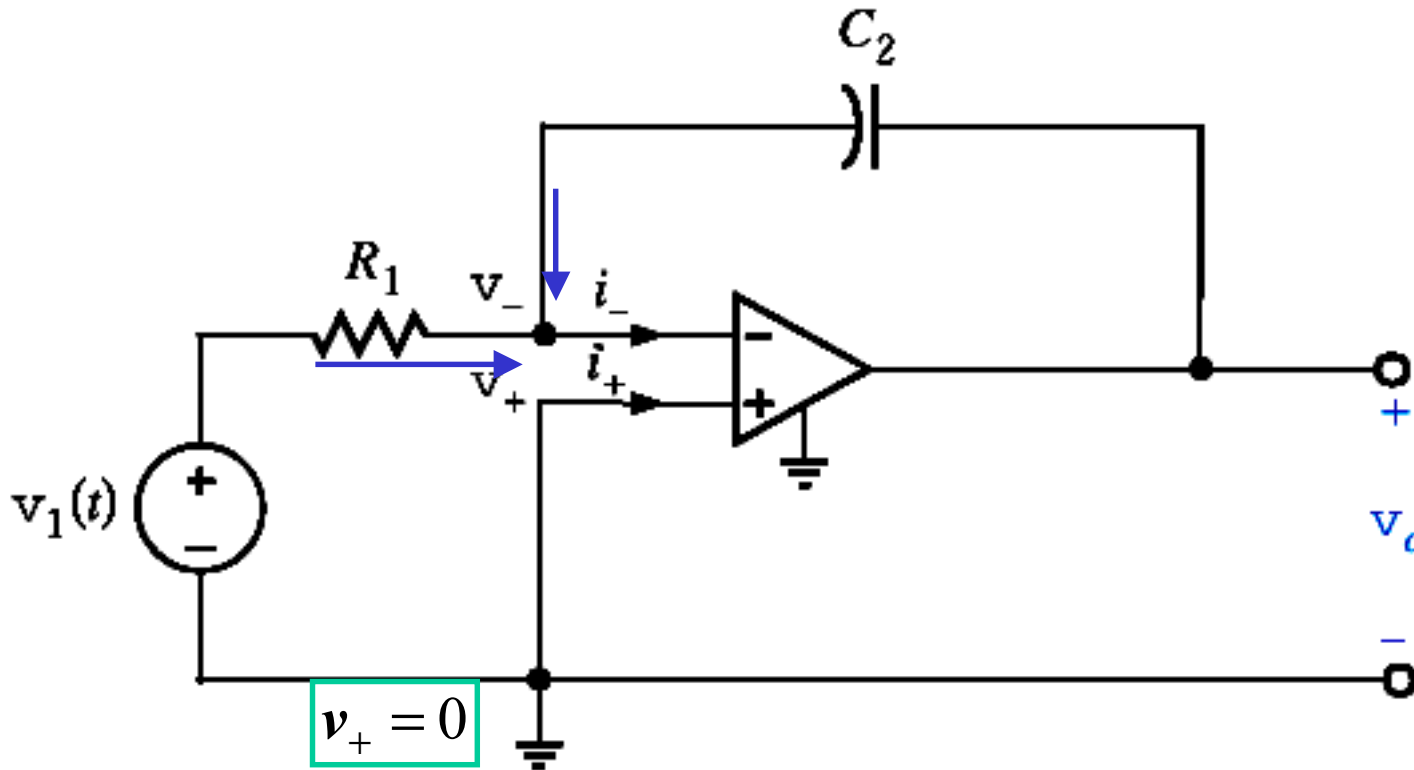
$$R_o = 0 \Rightarrow v_o = A(v_+ - v_-)$$

$$R_i = \infty \Rightarrow i_+ = i_- = 0$$

$$A = \infty \Rightarrow v_+ = v_-$$



# RC OPERATIONAL AMPLIFIER CIRCUITS -THE INTEGRATOR

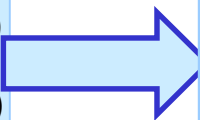


$$\frac{v_1 - v_-}{R_1} + C_2 \frac{d}{dt} (v_o - v_-) = i_-$$

## IDEAL OP-AMP ASSUMPTIONS

$$v_- = v_+ \quad (A = \infty)$$

$$i_- = 0 \quad (R_i = \infty)$$



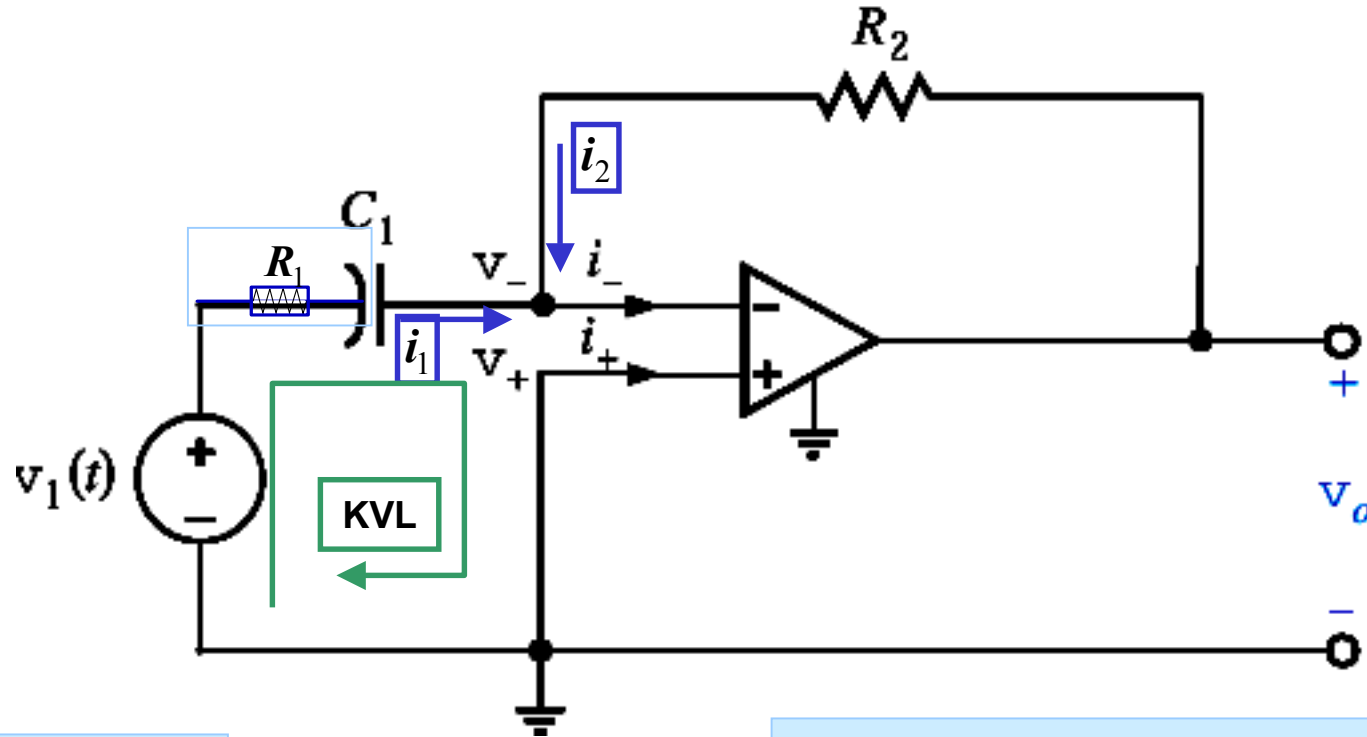
$$\frac{v_1}{R_1} = -C_2 \frac{dv_o}{dt}$$

$$v_o(t) = \frac{-1}{R_1 C_2} \int_{-\infty}^t v_1(x) dx$$

$$= \frac{-1}{R_1 C_2} \int_0^t v_1(x) dx + v_o(0)$$



# RC OPERATIONAL AMPLIFIER CIRCUITS - THE DIFFERENTIATOR



$$v_+ = 0$$

$$\text{KCL@ } v_- : i_1 + i_2 = i_-$$

## IDEAL OP-AMP ASSUMPTIONS

$$v_- = v_+ \quad (A = \infty)$$

$$i_- = 0 \quad (R_i = \infty)$$

$$i_1 + \frac{v_o}{R_2} = 0$$

replace  $i_1$  in terms of  $v_o$  ( $i_1 = -\frac{v_o}{R_2}$ )

$$R_1 C_1 \frac{dv_o}{dt} + v_o = -R_2 C_1 \frac{dv_1(t)}{dt}$$

$$v_1(t) = R_1 i_1 + \frac{1}{C_1} \int_{-\infty}^t i_1(x) dx$$

DIFFERENTIATE

$$R_1 C_1 \frac{di_1}{dt} + i_1 = C_1 \frac{dv_1(t)}{dt}$$

IF R1 COULD BE SET TO ZERO WE WOULD HAVE AN IDEAL DIFFERENTIATOR. IN PRACTICE AN IDEAL DIFFERENTIATOR AMPLIFIES ELECTRIC NOISE AND DOES NOT OPERATE. THE RESISTOR INTRODUCES A FILTERING ACTION. ITS VALUE IS KEPT AS SMALL AS POSSIBLE TO APPROXIMATE A DIFFERENTIATOR



## ABOUT ELECTRIC NOISE

ALL ELECTRICAL SIGNALS ARE CORRUPTED BY EXTERNAL, UNCONTROLLABLE AND OFTEN UNMEASURABLE, SIGNALS. THESE UNDESIRE SIGNALS ARE REFERRED TO AS NOISE

SIMPLE MODEL FOR A NOISY 60Hz SINUSOID CORRUPTED WITH ONE MICROVOLT OF 1GHz INTERFERENCE.

$$y(t) = \sin(120\pi t) + 10^{-6} \sin(2 \times 10^9 \pi t)$$

signal

noise

$$\frac{\text{noise amplitude}}{\text{signal amplitude}} = 10^{-6}$$

### THE DERIVATIVE

$$\frac{dy}{dt}(t) = 120\pi \cos(120\pi t) + 2000\pi \cos(2 \times 10^9 \pi t)$$

signal

noise

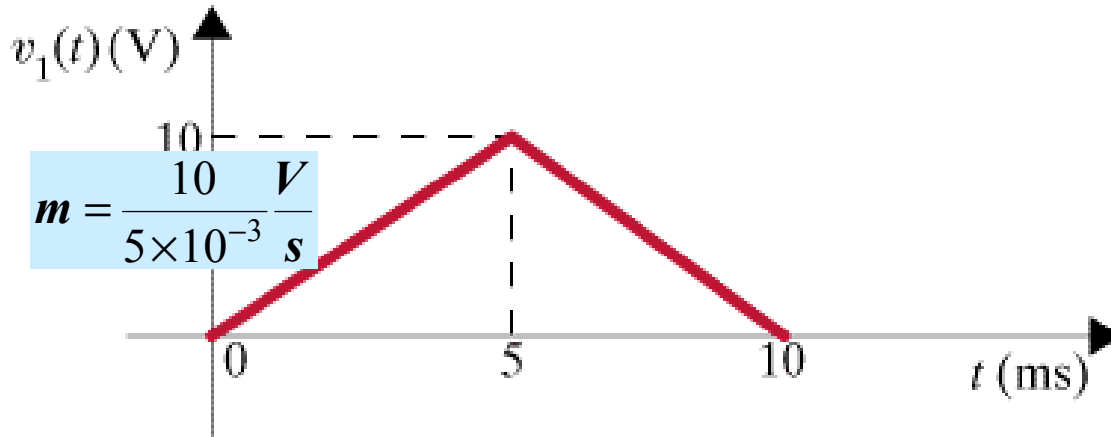
$$\frac{\text{noise amplitude}}{\text{signal amplitude}} = \frac{2000}{120} = 16.67$$





## LEARNING EXTENSION

INPUT TO IDEAL DIFFERENTIATOR WITH  $R_2 = 1k\Omega$ ,  $C_1 = 2\mu F$



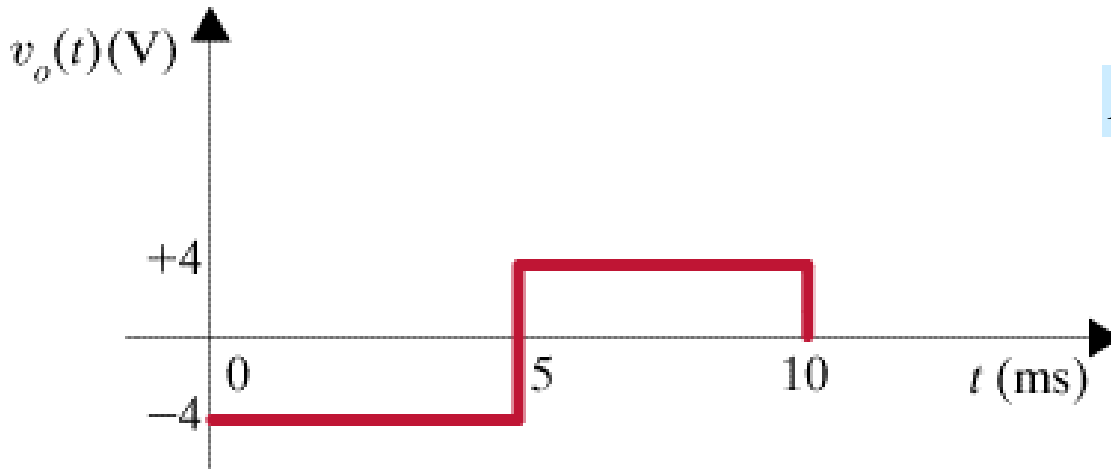
IDEAL DIFFERENTIATOR

$$v_o = -R_2 C_1 \frac{dv_1}{dt}(t)$$

DIMENSIONAL ANALYSIS

$$\Omega = \frac{V}{A} = \frac{V}{\frac{Q}{s}} = \frac{V \times s}{Q}$$

$$F = \frac{Q}{V} \Rightarrow \Omega \times F = s$$



$$R_2 C_1 = 1 \times 10^3 \Omega \times 2 \times 10^{-6} F = 2 \times 10^{-3} s$$



**LEARNING EXTENSION**

INPUT TO AN INTEGRATOR WITH  $R_1 = 5k\Omega, C_2 = 0.2\mu F$

CAPACITOR IS INITIALLY DISCHARGED

$R_1 C_2 = 10^{-3} s$

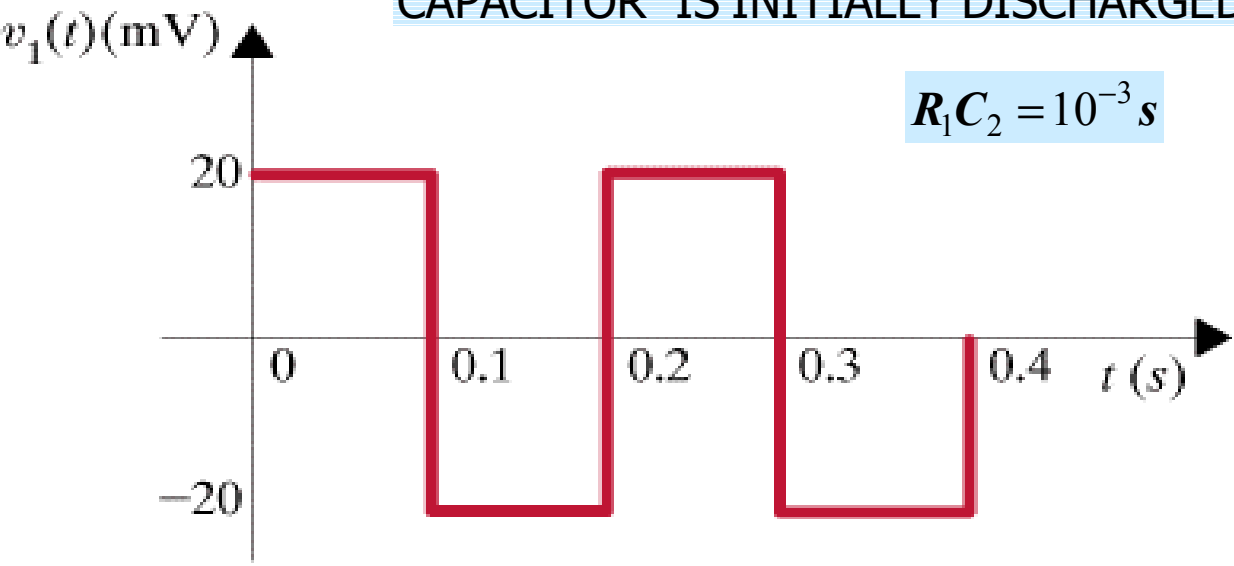
**INTEGRATOR**

$$v_o(t) = v_o(0) - \frac{1}{R_1 C_2} \int_0^t v_i(x) dx$$

**DIMENSIONAL ANALYSIS**

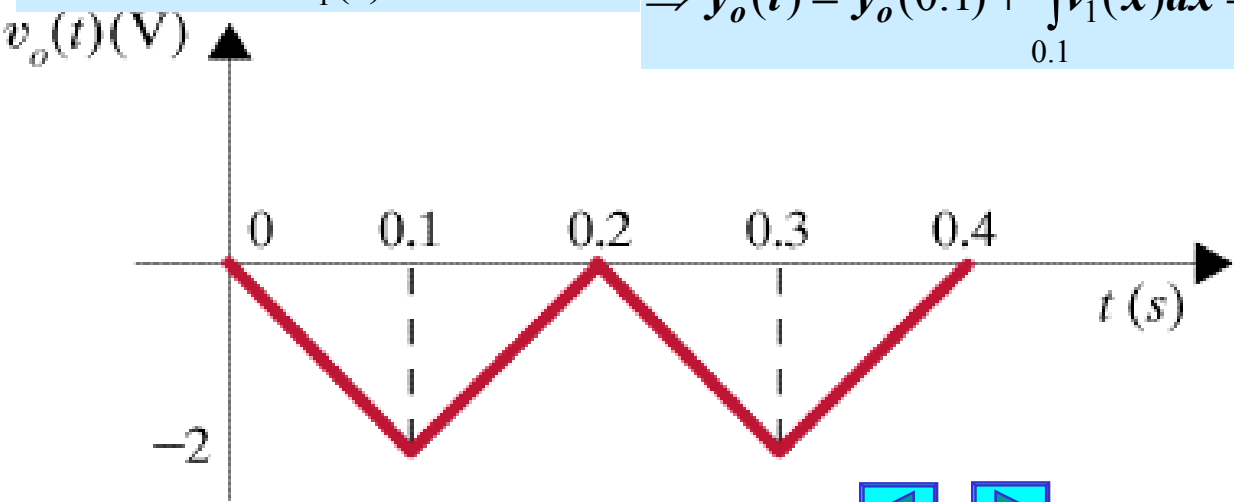
$$\Omega = \frac{V}{A} = \frac{V}{Q/s} = \frac{V \times s}{Q}$$

$$F = \frac{Q}{V} \Rightarrow \Omega \times F = s$$



$0 < t < 0.1s : v_1(t) = 20 \times 10^{-3} \Rightarrow y_o(t) = \int_0^t v_1(x) dx = 20 \times 10^{-3} t (V \times s) \Rightarrow y(0.1) = 2 \times 10^{-3} (V \times s)$

$0.1 < t < 0.2s : v_1(t) = -20 \times 10^{-3} \Rightarrow y_o(t) = y_o(0.1) + \int_{0.1}^t v_1(x) dx = 2 \times 10^{-3} - 20 \times 10^{-3} (t - 0.1) (V \times s)$

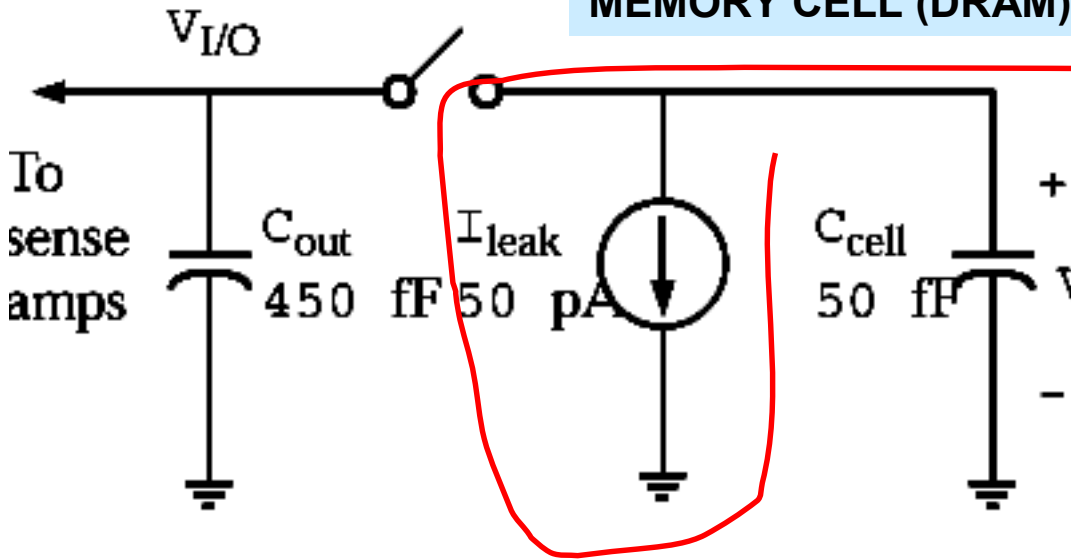


$$v_o(t) = \frac{1}{R_1 C_2} y_o(t)$$



# LEARNING EXAMPLE

## SIMPLE CIRCUIT MODEL FOR DYNAMIC RANDOM ACCESS MEMORY CELL (DRAM)



REPRESENTS CHARGE LEAKAGE FROM CELL CAPACITOR

NOTICE THE VALUES OF THE CAPACITANCES

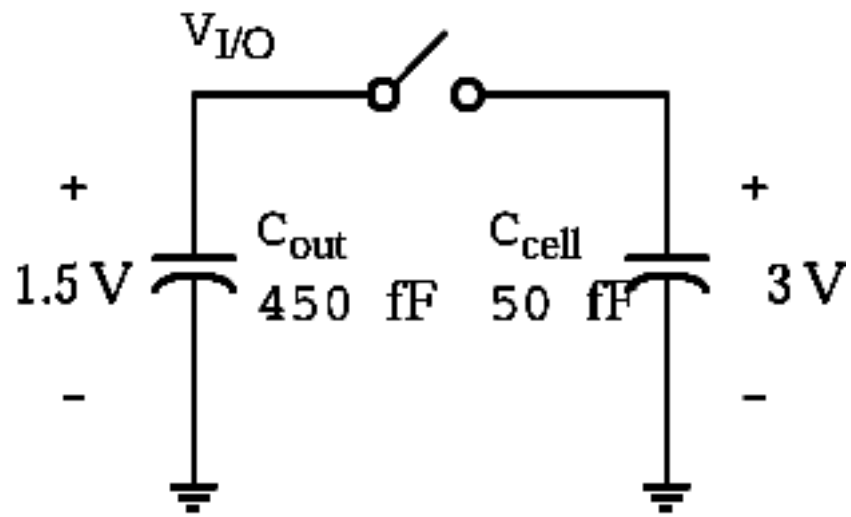
$V_{cell} > 1.5V$  FOR CORRECT STORAGE OF A LOGIC ONE

$$v_C = v_C(0) + \frac{1}{C} \int_0^t i_C(x) dx$$

$$V_{cell} = 3 - \frac{I_{leak}}{C_{cell}} t \geq 1.5 \Rightarrow \frac{I_{leak}}{C_{cell}} t \leq 1.5V$$

$$t_H = \frac{1.5(V) \times 50 \times 10^{-15}(F)}{50 \times 10^{-12} A} = 1.5 \times 10^{-3} s$$

THE CELL MUST BE "REFRESHED" AT A FREQUENCY HIGHER THAN  $1/t_H$



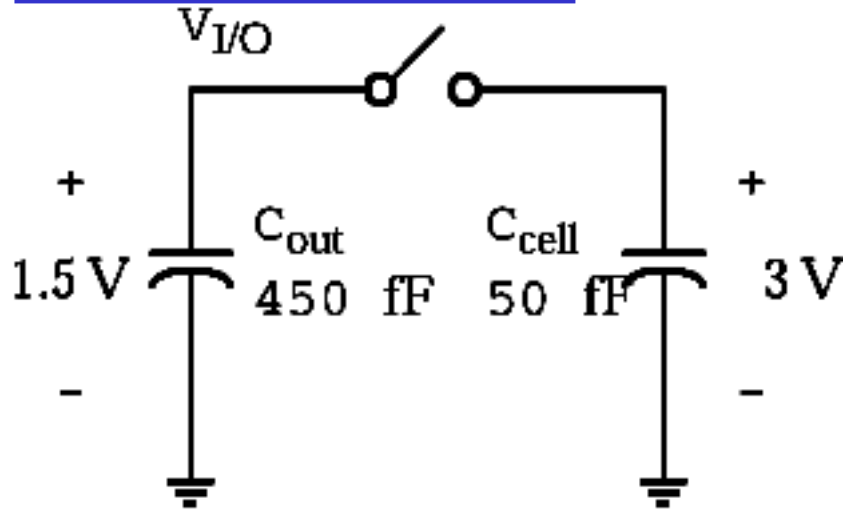
CELL AT THE BEGINNING OF A MEMORY READ OPERATION

SWITCHED CAPACITOR CIRCUIT

THE ANALYSIS OF THE READ OPERATION GIVES FURTHER INSIGHT ON THE REQUIREMENTS



## CELL READ OPERATION



IF SWITCH IS CLOSED BOTH CAPACITORS MUST HAVE THE SAME VOLTAGE

ASSUMING NO LOSS OF CHARGE THEN THE CHARGE BEFORE CLOSING MUST BE EQUAL TO CHARGE AFTER CLOSING

$$Q_{before} = 1.5V \times 450 \times 10^{-15} F + 3V \times 50 \times 10^{-15} F$$

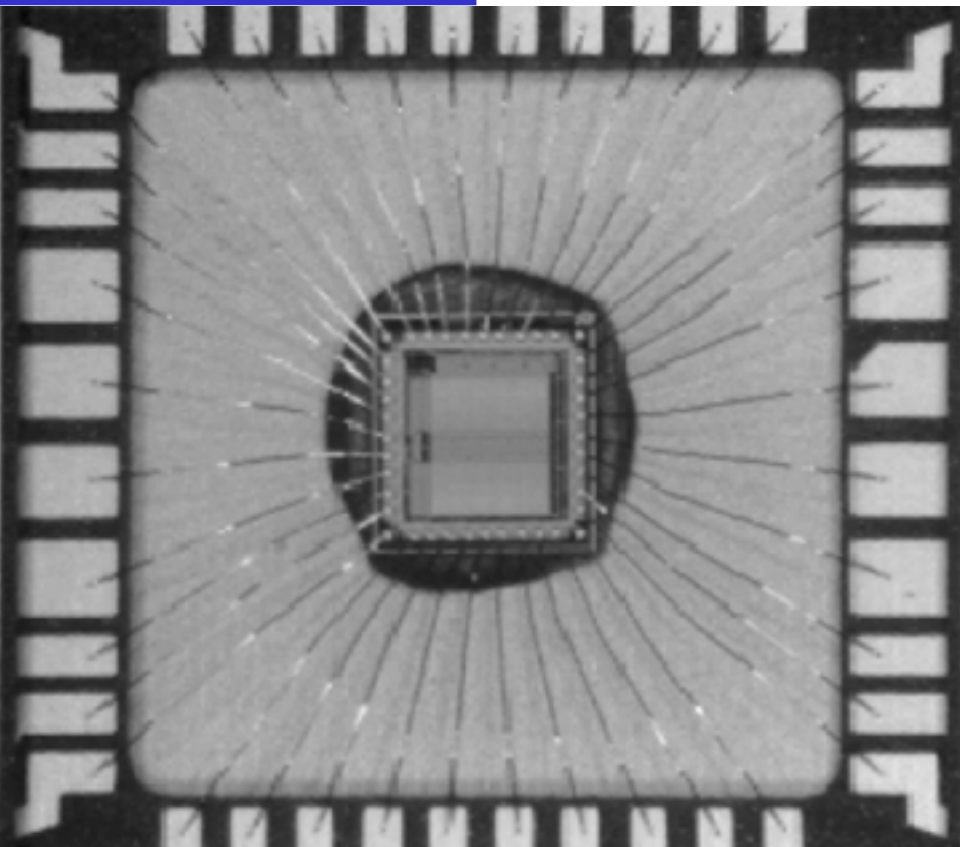
$$Q_{after} = V_{after} \times (500 \times 10^{-15} F)$$

$$V_{after} = 1.65V$$

Even at full charge the voltage variation is small.  
SENSOR amplifiers are required

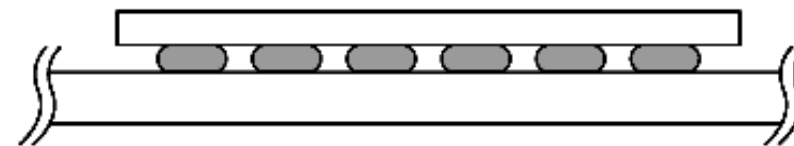
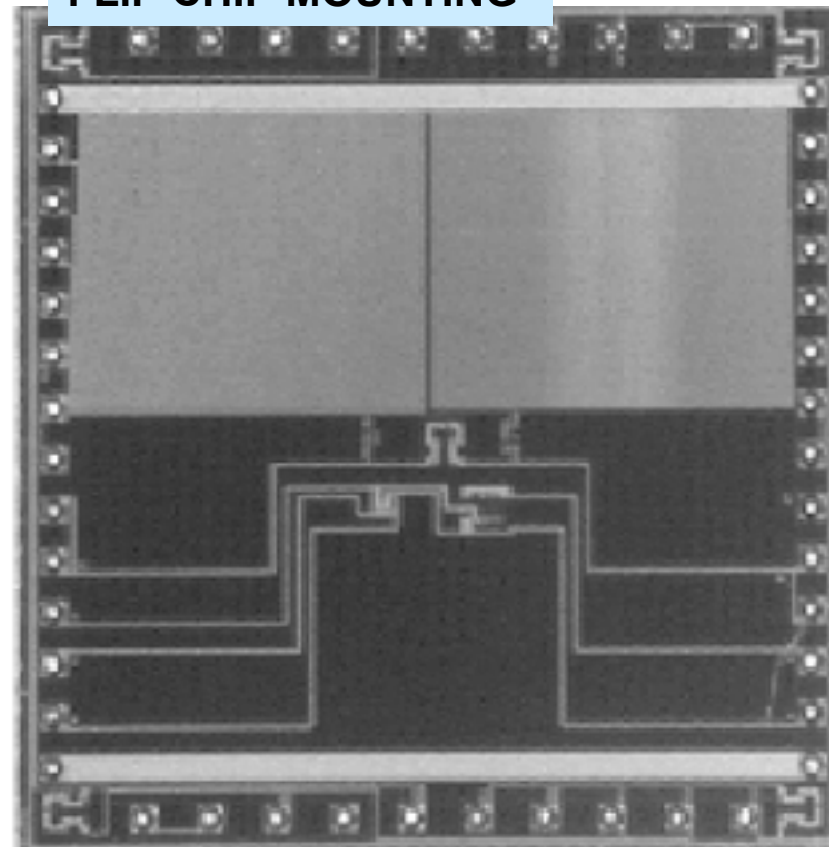
After a READ operation the cell must be refreshed





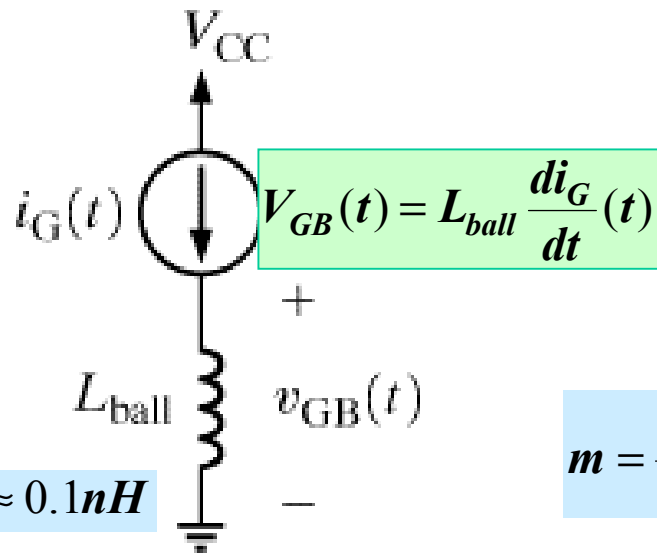
IC WITH WIREBONDS TO THE OUTSIDE

GOAL: REDUCE INDUCTANCE IN THE WIRING AND REDUCE THE "GROUND BOUNCE" EFFECT



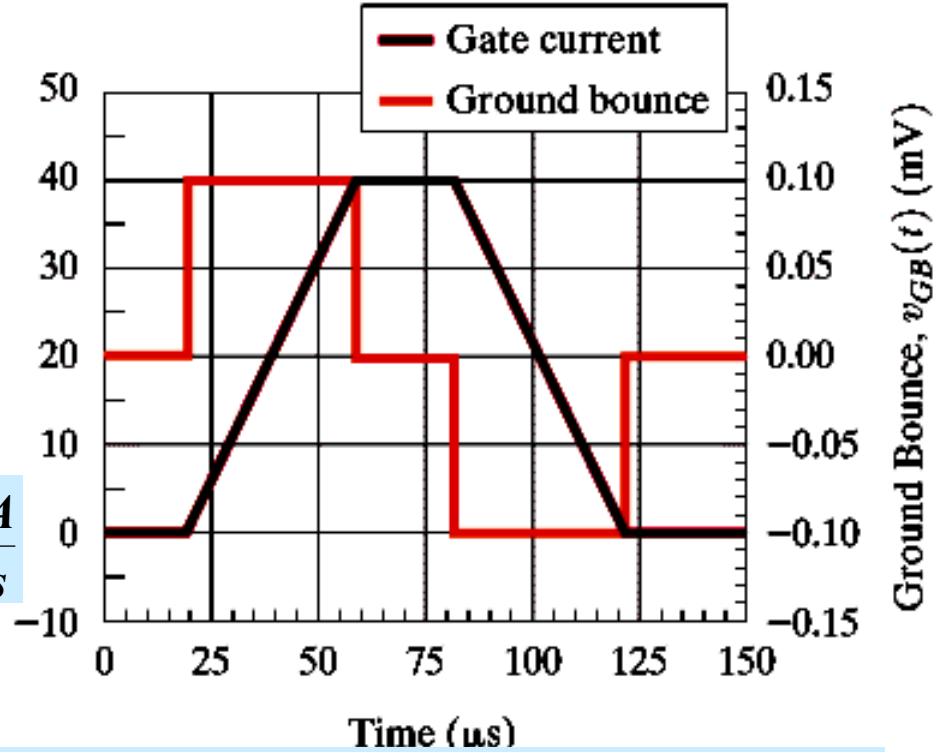
A SIMPLE MODEL CAN BE USED TO DESCRIBE GROUND BOUNCE

# MODELING THE GROUND BOUNCE EFFECT



$L_{ball} \approx 0.1nH$

$m = \frac{40 \times 10^{-3} A}{40 \times 10^{-9} s}$



IF ALL GATES IN A CHIP ARE CONNECTED TO A SINGLE GROUND THE CURRENT CAN BE QUITE HIGH AND THE BOUNCE MAY BECOME UNACCEPTABLE

USE SEVERAL GROUND CONNECTIONS (BALLS) AND ALLOCATE A FRACTION OF THE GATES TO EACH BALL

